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# 8554B SPECTRUM ANALYZER RF SECTION 



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#### Abstract

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## OPERATION AND SERVICE MANUAL

## 8554B SPECTRUM ANALYZER RF SECTION

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This manual applies directly to instruments with serial numbers prefixed 1245A.

For additional important information about serial numbers, see INSTRUMENTS COVERED BY MANUAL in Section I.
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Figure 1-1. Model 8554B Spectrum Anulyzer RF Section and Accessories Supplied

# SECTION I GENERAL INFORMATION 

## 1-1. INTRODUCTION

1-2. This manual contains the operating and service information for the Hewlett-Packard Model 8554B Spectrum Analyzer RF Section. The Spectrum Analyzer is shown in Figure 1-1 with all of its externally supplied accessories.
$1-3$. This section of the manual describes the instruments documented by this manual and covers instrument description, options, accessories, specifications and other basic information. The other sections provide the following information:

Section II, Installation: provides information about initial inspection, preparation for use, and storage and shipment.

Section III, Operation: provides information about panel features, and provides operating checks, instructions, and maintenance information.

Section IV, Performance Tests: provides the information required to verify that the instrument is performing as specified in Table 1-1.

Section V, Adjustments: provides the information required to properly adjust and align the instrument.

Section VI, Replaceable Parts: provides ordering information for all replaceable parts and assemblies.

Section VII, Manual Changes: this section is reserved to provide manual change information in future revisions of this manual.

Section VIII, Service: provides the information required to repair the instrument.

1-4. Packaged with this manual is an Operating Information Supplement. This is simply a copy of the first three sections of this manual. This supplement should stay with the instrument for use by the operator. Additional copies can be ordered through your nearest Hewlett-Packard Sales and Service Office; the part number is listed on the title page of this manual and on the rear
cover of the supplement.
1-5. Also listed on the title page of this manual is a "Microfiche" part number. This number can be used to order $4 \times 6$ inch microfilm transparencies of the manual. Each microfiche contains up to 60 photo duplicates of the manual's pages. The microfiche package also includes the latest Manual Changes supplement as well as all pertinent Service Notes.

1-6. Instrument specifications are listed in Table 1-1. These specifications are the performance standards or limits against which the instrument can be tested. Table 1-2 lists some supplemental performance characteristics. Supplemental characteristics are not specifications but are typical characteristics included as additional information for the user.

## 1-7. INSTRUMENTS COVERED BY MANUAL

1-8. This instrument has a two-part serial number. The first four digits and the letter comprise the serial number prefix. The last five digits form the sequential suffix that is unique to each instrument. The contents of this manual apply directly to instruments having the same serial number prefix(es) as listed under SERIAL NUMBERS on the title page.

1-9. An instrument manufactured after the printing of this manual may have a serial prefix that is not listed on the title page. This unlisted serial prefix indicates that the instrument is different from those documented in this manual. The manual for this instrument is supplied with a yellow Manual Changes supplement that contains "change information" that documents the differences.

1-10. In addition to change information, the supplement may contain information for correcting errors in the manual. To keep this manual as current and accurate as possible, Hewlett-Packard recommends that you periodically request the latest Manual Changes supplement. The supplement for this manual is keyed to this manual's print date and part number, both of which appear
on the title page. Complimentary copies of the supplement are available from Hewlett-Packard.

1-11. For information concerning a serial number prefix not listed on the title page or in the Manual Changes supplement, contact your nearest HewlettPackard office.

## 1-12. DESCRIPTION

1-13. The HP Model 8554B Spectrum analyzer RF Section covers the frequency range from 100 kHz to 1250 MHz . When it is combined with an IF Section and a Display Section it functions as the tuning section of an RF spectrum analyzer.

1-14. The analyzer electronically scans input signals and displays their frequency and amplitude on a CRT. The horizontal, $x$-axis, is calibrated in units of frequency and the vertical, y-axis, is calibrated in absolute units of voltage ( $\mu \mathrm{V}$ or mV ) or power $(\mathrm{dBm})$. Absolute or relative measurements of both amplitude and frequency can be easily made.

1-15. The horizontal (frequency) axis can be swept three different ways:
a. The center of the CRT is set to a frequency determined by the dial and the analyzer is swept symmetrically about that frequency.
b. The analyzer is not swept but is used as a fixed frequency receiver. Signal amplitude can be read on the CRT and signal modulation can be viewed as with an oscilloscope.
c. The analyzer is swept from 0 Hz to 1250 MHz.

1-16. Typically, the Spectrum Analyzer is used to measure the frequency and amplitude of the various components of a complex electrical signal, as well as frequency response, harmonic and intermodulation distortion, gain, attenuation, modulation index, spectral purity, noise density, and other parameters. These measurements may be made on amplifiers, oscillators, mixers, modulators, etc., to evaluate their performance with respect to their design specifications.

## 1-17. EQUIPMENT REQUIRED BUT NOT SUPPLIED

1-18. The 8554 B RF Section must be mated with an IF Section, such as the 8552A or the 8552B,
and a Display Section, such as the 140 T or the 141 T , before the units can perform as a spectrum analyzer.

## 1-19. IF Sections

1-20. The 8552A IF Section features calibrated bandwidths, log and linear amplitude calibration, and calibrated scan times. The 8552B IF Section has all of the features of the 8552 A and, in addition, manual scan, narrower bandwidth shape factors, 10 Hz video filter and an expanded $\log$ scale ( 2 dB per division).

## 1-21. Display Sections

$1-22$. The 140 S and 140 T Display Sections are equipped with a fixed persistence, non-storage CRT; the 141S and 141T Display Sections are equipped with a variable persistence, storage CRT. The 143S Display Section has a large screen ( 8 x 10 inch) CRT. Overlays are available for the standard 140A and 141A Oscilloscope Mainframes to provide $\log$ and linear graticule scales.

## 1-23. COMPATIBILITY

1-24. The HP 8554B RF Section is fully compatible with all HP 140S/T, 141S/T, and 143S Display Sections. The HP 8554B is also compatible with all HP $140 \mathrm{~A} / \mathrm{B}$ and $141 \mathrm{~A} / \mathrm{B}$ Oscilloscope Mainframes.

1-25. The HP 8554B RF Section is fully compatible with all current $8552 \mathrm{~A} / \mathrm{B}$ IF Sections. HP 8552 A's with serial number prefixes $809,821,825$, and 837 should be modified to ensure compatibility with the 8554 B . This modification can be made with a kit, HP Part No. 08552-6048, which is available on request at no cost.

## 1-26. ACCESSORIES SUPPLIED

1-27. The HP 8554 B is supplied with two HP 11593A $50 \Omega$ Coaxial Terminations (BNC). These connect to the LO outputs on the front panel.

## 1-28. OPERATING ACCESSORIES

1-29. The instruments listed below can be used to expand the analyzer's measurement capability. The brief descriptions list some of the features and applications of each instrument. For more information, contact your local Hewlett-Packard Sales and Service Office.

1-30. Tracking Generator. The HP 8444A Tracking Generator is a companion instrument to the 8554B/8552 Spectrum Analyzer. The tracking generator provides a CW signal that precisely tracks the analyzer's tuning frequency. The signal's amplitude is calibrated and can be adjusted from 0 dBm to -10 dBm . This signal can be used, with a frequency counter, to make precise frequency measurements (see HP 8444A manual). It can also be used to make frequency response and reflection coefficient measurements.

1-31. Preamplifiers. The HP 8447 -series amplifiers cover various portions of the analyzer's frequency range and can be used to improve the sensitivity of the analyzer:
a. The HP 8447 A ; $0.1-400 \mathrm{MHz}, 20 \mathrm{~dB}$ gain, $<5 \mathrm{~dB}$ noise figure.
b. HP 8447B; $0.4-1.3 \mathrm{GHz},>20 \mathrm{~dB}$ gain, $<6 \mathrm{~dB}$ noise figure ( $<5 \mathrm{~dB}$ below 1.0 GHz ).
c. HP $8447 \mathrm{D} ; 100 \mathrm{kHz}-1.3 \mathrm{GHz}, 26 \mathrm{~dB}$ gain, $<8.5 \mathrm{~dB}$ noise figure.

1-32. Oscilloscope Cameras. The 196B, 197A and 123A Oscilloscope Cameras attach directly to the analyzer's CRT bezel and can be used to permanently record any signal displayed on the CRT (see paragraph 3-43).

## 1-33. TEST EQUIPMENT REQUIRED

$1-34$. Tables $1-3$ and 1-4 list the test equipment and test equipment accessories to test, adjust and repair the 8554 B .

## 1-35. WARRANTY

1-36. The HP 8554B Spectrum Analyzer RF Section is warranted and certified as indicated on the inner front cover. For further information, contact your nearest Hewlett-Packard Sales and Service Office. Addresses are provided at the back of this manual.

Table 1-1. Specifications

## SPECIFICATIONS <br> 8554B/8552A/8552B

## FREQUENCY

Range: 100 kHz to 1250 MHz .
Scan Width (on 10 -division CRT horizontal axis):
Per Division: 15 calibrated scan widths from 100 $\mathrm{MHz} /$ div to $2 \mathrm{kHz} /$ div in a $1,2,5$ sequence.
Preset: $0 \cdot 1250 \mathrm{MHz}$, automatically selects 300 kHz bandwidth IF Filter.
Zero:Analyzer is fixed tuned receiver.

## Accuracy:

Center Frequency Accuracy: The dial indicates the display center frequency within 10 MHz .
Scan Width Accuracy: Frequency error between two points on the display is less than $10 \%$ of the indicated separation.

## Resolution:

Bandwidth: IF bandwidths of 0.1 to 300 kHz provided in a 1,3 sequence.
Bandwidth Accuracy:Individual IF bandwidths' 3 dB points calibrated to $\pm 20 \%$ ( 10 kHz bandwidth $\pm 5 \%$ ).
Bandwidth Selectivity: $60 \mathrm{~dB} / 3 \mathrm{~dB}$ IF bandwidth ratio $<20: 1$ for $I F$ bandwidths from 10 kHz to $300 \mathrm{kHz} .60 \mathrm{~dB} / 3 \mathrm{~dB}$ bandwidth ratio $<11: 1$ for IF bandwidths 100 Hz to $3 \mathrm{kHz} .{ }^{1}$

Stability:
Residual FM:
Stabilized: $<100 \mathrm{~Hz}$ peak-to-peak. Unstabilized: $<10 \mathrm{kHz}$ peak-to-peak.
Noise Sidebands: More than 70 dB below CW signal, 50 kHz or more away from signal, with 1 kHz IF bandwidth.

## AMPLITUDE

Absolute Amplitude Calibration Range:
Log: From -122 to $+10 \mathrm{dBm}, 10 \mathrm{~dB} / \mathrm{div}$ on a 70 dB display, or $2 \mathrm{~dB} /$ div on a 16 dB display. ${ }^{2}$
Linear: From $0.1 \mu \mathrm{~V} /$ div to $100 \mathrm{mV} /$ div in a 1,2 sequence on an 8 -division display.

## Dynamic Range:

Average Noise Level: $<-102 \mathrm{dBm}$ with 10 kHz IF Bandwidth.
Spurious Responses: All image and out-of-band mixing responses, harmonic and intermodulation distortion products are more than $65 \mathrm{~dB}^{3}$ below a -40 dBm signal at the input mixer.*
Residual Responses (no signal present at input): With input attenuation at $0 \mathrm{~dB}:<-100 \mathrm{dBm}$.

[^0]
## Accuracy:

Log
Linear
Frequency Response
(Flatness):
100 kHz to $1250 \mathrm{MHz}: \quad \pm 1 \mathrm{~dB} \quad \pm 12 \%$
Switching Between
Bandwidths (at $20^{\circ} \mathrm{C}$ ): $\pm 0.5 \mathrm{~dB} \quad \pm 5.8 \%$
Amplitude Display: $\quad \pm 0.25 \mathrm{~dB} / \mathrm{dB}$ but $\quad 2.8 \%$ of not more than full 8 div $\pm 1.5 \mathrm{~dB}$ over the deflection range

## RF INPUT SPECIFICATIONS

Input Impedance: $50 \Omega$ nominal. Reflection coefficient $<0.30$ ( 1.85 SWR ), input attenuator $\geqslant 10 \mathrm{~dB}$.
Maximum Input Level: Peak or average power +13 dBm (1.4 Vac peak), $\pm 50 \mathrm{Vdc}$ incident on input mixer and +33 dBm incident on input attenuator (in the 50 dB position for $<1 \mathrm{~dB}$ gain compression).

## GENERAL

## Calibrator Output:

Amplitude: $-30 \mathrm{dBm}, \pm 0.3 \mathrm{~dB}$.
Frequency: $30 \mathrm{MHz}, \pm 3 \mathrm{kHz}{ }^{4}$
Scan Time: 16 internal scan rates from $0.1 \mathrm{~ms} /$ div to 10 sec/div in a $1,2,5$ sequence and manual scan. ${ }^{5}$

## Scan Time Accuracy:

$0.1 \mathrm{~ms} /$ div to $20 \mathrm{~ms} /$ div: $\pm 10 \%$.
$50 \mathrm{~ms} /$ div to $10 \mathrm{~s} /$ div: $\pm 20 \%$.
Power Requirements: $100,120,220,240 \mathrm{~V}+5 \%-10 \%$, 50 to 60 Hz , normally less than 225 watts (includes plug-ins used).
Dimensions: With Model 140 T or 141T Display Section: $9-1 / 16 \mathrm{in}$. high (including height of feet) $\times 16-3 / 4 \mathrm{in}$. wide $\times 18-3 / 8 \mathrm{in}$. deep ( $229 \times 425 \times 467 \mathrm{~mm}$ ).

## Weight:

Model 8554B RF Section: Net, $10 \mathrm{lb}, 4 \mathrm{oz}(4.7 \mathrm{~kg})$. Model 8552A or 8552B IF Section: Net, $9 \mathrm{lb}(4.1 \mathrm{~kg})$. Model 140 T Display Section: Net, $37 \mathrm{lb}(16.8 \mathrm{~kg})$. Model 141T Display Section: Net, $40 \mathrm{lb}(18 \mathrm{~kg})$.

[^1]These Supplemental Performance Characteristics expand the 8554B/8552B specifications, describe the instruments' unique features and characteristics, and provide other information useful in applying the instrument.

## FREQUENCY CHARACTERISTICS

Frequency Range: For operation of the analyzer outside the 100 kHz to 1250 MHz range, see Figure B.

## Scan Width:

Preset 0-1250 MHz: Inverted marker identifies the frequency that becomes the center frequency for scan width per division and zero scan modes.
Zero: Analyzer becomes fixed-tuned receiver with frequency set by FREQUENCY and FINE TUNE controls and selectable bandwidths set by BANDWIDTH control. Amplitude variations are displayed versus time on the CRT.
Resolution: See Figure A for curves of typical Spectrum Analyzer resolution using different IF bandwidths.
Stability: First local oscillator can be automatically stabilized (phase-locked) to internal reference for scan widths of $200 \mathrm{kHz} /$ div or less. Signal display shift with stabilization $<50 \mathrm{kHz}$.
Long Term Drift: (At fixed center frequency, after 2-hour warm-up).
Stabilized: $\pm 10 \mathrm{kHz} / 10 \mathrm{~min}$.
Unstabilized: $\pm 50 \mathrm{kHz} / 10 \mathrm{~min}$.

## Temperature Drift:

Stabilized: $\left.100 \mathrm{kHz}\right|^{\circ} \mathrm{C}$.
Unstabilized: $200 \mathrm{kHz} /^{\circ} \mathrm{C}$.

## AMPLITUDE CHARACTERISTICS

The average noise level indicates the maximum sensitivity of the analyzer. For typical noise level versus input frequency curves from 100 kHz to 1250 MHz , see Figure B.

Dynamic Range: For dynamic range with other than -40 dBm input level, see Figure C.
Gain Compression: For $<-10 \mathrm{dBm}$ signal level to input mixer gain compression $<1 \mathrm{~dB}$.
Amplitude Accuracy:
Measurement Accuracy: Largely determined by frequency response ( $\pm 1 \mathrm{~dB}$ ) and display accuracy ( $\pm 1.5 \mathrm{~dB}$ ) for general use. This $\pm 2.5 \mathrm{~dB}$ can be improved using IF substitution techniques.
Frequency Response (flatness): For typical frequency response characteristics, see Figure B.
Log Reference Level: Controls provide continuous Log Reference Levels from +10 dBm to $-72 \mathrm{dBm}(-50$ dBm below 200 kHz ).
Log Reference Level Control: Provides 70 dB range ( 60 dB below 200 kHz ), in 10 dB steps. Accurate to $\pm 0.2 \mathrm{~dB}( \pm 2.3 \%$, Linear Sensitivity).
Log Reference Level Vernier: Provides continuous 12 dB range, accurate to $\pm 0.1 \mathrm{~dB}( \pm 1.2 \%)$ in 0 , -6 , and -12 dB positions, otherwise $\pm 0.25 \mathrm{~dB}$ $( \pm 2.8 \%)$.


Figure A. Typical Spectrum Analyzer Resolution


Figure B. Average Noise Level vs. Input Frequency


- 0 dB Input Attenuation \{Input Signal Level = RF Input Level - Input Alten,

Figure C. Typical Dynamic Range Curve

Amplitude Stability: $\pm 0.07 \mathrm{~dB} /{ }^{\circ} \mathrm{C}$ in $\mathrm{Log}, \pm 0.6 \% /{ }^{\circ} \mathrm{C}$ in Linear.
Display Uncalibrated Light: Warns if a combination of control settings (IF or video bandwidth; scan width or time) degrades the absolute calibration.
Video Filter: Average displayed noise; 10 kHz and 100 Hz bandwidths. 10 Hz bandwidth also available on 8552B IF Section.

## RF INPUT CHARACTERISTICS

Impedance: $50 \Omega$ nominal, Type $N$ connector; for $75 \Omega$ use matching transformer, such as HP 11694A.
Reflection Coefficient: When analyzer is tuned to input signal $\rho \leqslant 0.4$ (2.33 SWR), for input attenuation $=0$ $\mathrm{dB} ; \rho \leqslant 0.2$ ( 1.50 SWR ), for input attenuation $\geqslant 10$ dB.
Attenuator: 0 to 50 dB in 10 dB steps, coupled to $\log$ reference level indicator to automatically maintain absolute amplitude calibration. Attenuator accuracy $\pm 1.0 \mathrm{~dB}$ but not more than $\pm 0.6 \mathrm{~dB} /$ step.

## GENERAL CHARACTERISTICS

## Scan Mode:

Int: Analyzer repetively scanned by internally generated ramp; synchronization selected by scan trigger.
Single: Single scan with reset actuated by front panel pushbutton.
Ext: Scan determined by 0 to +8 volt external signal; scan input impedance $>10 \mathrm{k} \Omega$. Blanking: -1.5 V external blanking signal required.
Scan Trigger: For Internal scan mode, select between:
Auto: Scan free runs.
Line: Scan synchronized with power line frequency.
Ext: Scan synchronized with $>2$ volt ( 20 volt max.) trigger signal (polarity selected by internally located switch in Model 8552A or 8552B IF Section, normally negative).
Video: Scan internally synchronized to envelope of RF input signal (signal amplitude of 1.5 major divisions peak-to-peak required on display section CRT).
Auxiliary Outputs:
Vertical Output: Approximately 0 to -0.8 V for 8 division deflection on CRT display; approx. $100 \Omega$ output impedance.
Scan Output: Approximately -5 to +5 V for 10 division CRT deflection, $1 \mathrm{k} \Omega$ output impedance.
Pen Lift Output: 0 to $14 \mathrm{~V}(0 \mathrm{~V}$, pen down). Output available in INT and SINGLE SCAN modes and AUTO, LINE, and VIDEO scan trigger.
CRT Baseline Clipper: Front panel control adjusts blanking of CRT trace baseline to allow more detailed analysis of low repetition-rate signals and improved photographic records to be made.

EMI: Conducted and radiated interference is within requirements of MIL-1-16910C and MIL-I6181 D and methods CE03, and RE02 of MIL-STD-461 (except 35 to 40 kHz ) when 8554B and 8552 A or 8552 B are combined in a 140 T or 141 T Display Section.
Temperature Range: Operating, $0^{\circ}$ to $+55^{\circ} \mathrm{C}$; storage, $-40^{\circ}$ to $+75^{\circ} \mathrm{C}$.

## DISPLAY CHARACTERISTICS

## Model 141T Variable Persistence/Storage Display:

Plug-ins: Accepts Model 8550 -series Spectrum Analyzer plug-ins and Model 1400 -series Oscilloscope plug-ins.

## Cathode-ray Tube:

Type: Post-accelerator storage tube, 9000 volt accelerating potential; aluminized P31 phosphor; etched safety glass face plate reduces glare.
Graticule: $8 \times 10$ division (approximately $7.1 \times$ 8.9 cm ) parallax-free internal graticule; five subdivisions per major division on horizontal and vertical axes.

## Persistence:

Normal: Natural persistence of P31 phosphor (approximately 0.1 second).
Variable:
Normal Writing Rate Mode: Continuously variable from less than 0.2 second to more than one minute (typically to two or three minutes).
Maximum Writing Rate Mode: Typically from 0.2 second to 15 seconds.

Erase: Manual; erasure takes approximately 350 ms ; CRT ready to record immediately after erasure.
Storage Time: Normal writing rate; more than 2 hours at reduced brightness (typically 4 hours). More than one minute at maximum brightness.
Fast Writing Speed: More than 15 minutes (typically 30 minutes) at reduced brightness or more than 15 seconds at maximum brightness.
Functions Used with Oscilloscope Plug-ins Only: Intensity modulation, calibrator; beam finder.
Dimensions: 9-1/16 in. high (including height of feet) x $16-3 / 4$ in. wide $\times 18-3 / 4 \mathrm{in}$. deep ( $229 \times 425 \times 467$ mm ).
Weight: Model 141T Display Section: Net, $40 \mathrm{lb}(18 \mathrm{~kg})$.

## Model 140T Normal Persistence Display:

Plug-ins: Same as 141T.
Cathode-ray Tube:
Type: Post-accelerator, 7300 volt potential medium-short persistence (P7) phosphor; tinted and etched safety glass face plate reduces glare. (Normal persistence of P7 phosphor approximately 3 seconds).

Graticule: $8 \times 10$ division (approximately $7,6 \times$ $9,5 \mathrm{~cm}$ ) parallax-free internal graticule; five subdivisions per major division on horizontal and vertical.
Functions Used with Oscilloscope Plug-ins Only: Same at 141 T .
Dimensions: Same as 1417.
Weight: Model 140T Display Section: Net, 37 lb $(16.8 \mathrm{~kg})$.

Model 1435 Normal Persistence Large Screen Display:
Plug-ins: Same as 141T.
Cathode-ray Tube:
Type: Post accelerator, 20 kV accelerating po-
tential aluminized P31 phosphor. (Persistence approximately 0.1 second.)
Graticule: $8 \times 10$ divisions (approximately $8 \times$ 10 inch) parallax-free internal graticule, five subdivisions per major division on horizontal and vertical axes.
Functions Used with Time Domain Plug-ins Only: Same as 141T.
Dimensions: 21 in . high (including height of feet) x $16-3 / 4 \mathrm{in}$. wide $\times 18-3 / 8 \mathrm{in}$. deep ( $533 \times 425 \times$ 467 mm ).
Weight: Model 143S Display Section: Net, 62 lb ( $28,1 \mathrm{~kg}$ ).

Table 1-3. Test Equipment

| Item | Minimum Specifications | Suggested Model | Use* |
| :---: | :---: | :---: | :---: |
| Frequency Comb Generator | Frequency markers spaced $1,10,100 \mathrm{MHz}$ apart; usable to 1200 MHz <br> Frequency Accuracy: $\pm 0.01 \%$ <br> Output Amplitude: $>-40 \mathrm{dBm}$ | HP 8406A Comb Generator | P,A |
| HF Signal Generator | Frequency Range: 1-50 MHz <br> Output Amplitude: -20 dBm <br> Output Amplitude Accuracy: $\pm 1 \%$ <br> Frequency Accuracy: $\pm 1 \%$ <br> Output Impedance: 50 ohms | HP 606B HF Signal Generator | P,T |
| VHF Signal Generator | Frequency Range: $40-455 \mathrm{MHz}$ <br> Frequency Accuracy: $\pm 1 \%$ <br> Output Amplitude: $>-20 \mathrm{dBm}$ <br> Output Impedance: 50 ohms | HP 608F VHF Signal Generator | $\mathrm{P}, \mathrm{T}$ |
| Sweep Oscillator | Frequency Range: 1.110 MHz <br> Output Flatness: $\pm 0.25 \mathrm{~dB}$ over full band Output Impedance: 50 ohms Output Amplitude: at least 0 dBm | HP 8601A Generator/ Sweeper | P |
| Crystal Detector (2) | Frequency Range: $10-1300 \mathrm{MHz}$ <br> Sensitivity: $>0.4 \mathrm{mV} / \mu \mathrm{W}$ <br> Frequency Response: $\pm 0.2 \mathrm{~dB}$ <br> Polarity: Negative | HP 423A <br> Crystal Detector | P |
| UHF Signal Generator | Frequency Range: $450-1200 \mathrm{MHz}$ <br> Frequency Accuracy: 1\% <br> Output Amplitude: -20 dBm <br> Output Impedance: 50 ohms | HP 612A UHF Signal Generator | A, T |
| Audio Oscillator | Frequency Range: 10 kHz Output Amplitude: 2 Vrms Frequency Accuracy: $\pm 2 \%$ Output Impedance: 600 ohms | HP 200 CD Audio Oscillator | P |
| Sweep Oscillator | Frequency Range: $100-1250 \mathrm{MHz}$ Output Amplitude: +10 dBm Output Impedance: 50 ohms | HP 8690B/8699B Sweep Oscillator/ RF Unit | P |
| Frequency Counter | Frequency Range: $100 \mathrm{kHz}-3.0 \mathrm{GHz}$ <br> Accuracy: $\pm 0.001 \%$ <br> Sensitivity: 100 mVrms <br> Readout Digits: 7 digits | HP 5245L Frequency Counter with HP 5254C Plug-in | A,T |
| Tunable RF <br> Voltmeter | Bandwidth: 1 kHz <br> Frequency Range: 1-100 MHz <br> Sensitivity: $10 \mathrm{mV}-1 \mathrm{Vrms}$ <br> Input Impedance: $\geqslant 0.1$ megohms | HP 8405A Vector Voltmeter | T |
| *Use: Performance $=\mathbf{p} ;$ Adjustment $=$ A; Troubleshooting $=\mathrm{T}$ |  |  |  |

Test 1-3. Test Equipment (cont'd)

| Item | Minimum Specifications | Suggested Model | Use* |
| :---: | :---: | :---: | :---: |
| Digital Voltmeter | Voltage Accuracy: $\pm 0.2 \%$ <br> Range Selection: Manual or Automatic Voltage Range: 1-1000 Vdc full scale Input Impedance: 10 megohms Polarity: Automatic indication | HP 3440A Digital Voltmeter with HP 3443A Plug-in | T |
| Oscilloscope | Frequency Range: dc to 50 MHz Time Base: $1 \mu \mathrm{~s} / \mathrm{div}$ to $10 \mathrm{~ms} / \mathrm{div}$ Time Base Accuracy: $\pm 3 \%$ <br> Dual Channel, Alternate Operation ac or dc Coupling External Sweep Mode <br> Voltage Accuracy: $\pm 3 \%$ <br> Sensitivity: $0.005 \mathrm{~V} / \mathrm{div}$ | HP 180A with HP 1801A Vertical <br> Amplifier and HP 1821A Horizontal Amplifier HP 10004 10:1 Divider Probes (2) | P,A,T |
| Spectrum Analyzer | Frequency Range: $0-100 \mathrm{MHz}$ Scan Width: 10 MHz | HP 8553B/8552A/ 141T Spectrum Analyzer | A |
| Volt-ohm-ammeter | Resistance Range: $1 \Omega$ to $100 \mathrm{M} \Omega$ <br> Accuracy: $\pm 10 \%$ of Reading | HP 412A | T |
| *Use: Performance $=$ P; Adjustment $=$ A; Troubleshooitng $=T$ |  |  |  |

Table 1-4. Test Accessories

| Item | Required Features | Suggested Model | Use* |
| :---: | :---: | :---: | :---: |
| 20 dB Attenuator | Frequency Range: $100 \cdot 1250 \mathrm{MHz}$ <br> Flatness: $\pm 0.2 \mathrm{~dB}$ | HP 8491A Op 020 | P |
| 50-0hm Tee | Type N female connectors on two ports, with the third port able to accept HP 8405A probe tips. | HP 11536A 50 ohm Tee | T |
| 10 dB Fixed Attenuator | Frequency Range: 100 - 1250 MHz Flatness: $\pm 0.2 \mathrm{~dB}$ | HP 8491A, Option 010 | P |
| Dual Directional Coupler | Frequency Range: $100-1250 \mathrm{MHz}$ Directivity: 36 dB | HP 778D Dual Directional Coupler | P |
| BNC Tee | Two BNC Female Connectors, One Male BNC Connector | UG-274A/U <br> HP 1250-0781 | P,T |
| Adapter (two) | BNC Female to Type N Male | $\begin{aligned} & \text { UG-201A/U } \\ & \text { HP 1250-0067 } \end{aligned}$ | P,A |
| Voltage Probe | Dual Banana Plug-to-Probe Tip and Clip (Ground) Lead | HP 10025A StraightThrough Voltage Probe | T |
| Cable Assembly (5) | Male BNC Connectors, 48 inches long | HP 10503A | P,A,T |
| Cable Assembly | BNC Male to Dual Banana Plug, 45 inches long | HP 11001A | P |
| Tuning Tool, Blade | Nonmetallic Shaft, 6 inches long | General Cement 5003 (HP 8730-0013) | A |
| Tuning Tool, Slot | Nonmetallic, 6-inch shaft | Gowanda PC9668 | A |
| Wrench | Openend, 15/64-inch | HP 8710-0946 | A |
| Wrench | Openend, 5/16-inch | HP 8720-0030 | A |
| Wrench | No. 6, Allen Driver | HP 5020-0289 | A |
| Wrench | No. 10, Allen Driver | HP 5020-0291 | A |
| Wrench | Nut Driver, $5 / 16$-inch | HP 8720-0003 | A |
| Screwdrivers | Phillips \# 1 <br> Phillips \# 2 <br> Pozidrive \# 1 (Small) Stanley \# 5531 <br> Pozidrive \# 2 (Medium) Stanley \# 5332 | HP 8710-0899 HP 8710-0900 | A,T |
| Tuning Tool, Slot | Nonmetallic, 2.5 -inch shaft | HP 8710-0095 | A |
| Cover Assembly | Modified display section cover, see Paragraph 5-12 | Modified HP 5060-0470 | A |
| Soldering Iron | 47-1/2 watt | Ungar \# 776 with \# 4037 Heating Unit | A,T |
| * Use: Performance = P; Adjustment $=$ A; Troubleshooting $=T$ |  |  |  |

Table 1-4. Test Accessories (cont'd)

| Item | Required Features | Suggested Model | Use* |
| :---: | :---: | :---: | :---: |
| Adapter | Type N Tee Male Connector to Type N Female Connector | UG-107B/U | P |
| Adapter | Type N Male Connector to BNC Male Connector | $\begin{aligned} & \text { UG-1034/U } \\ & \text { HP 1250-0082 } \end{aligned}$ | P |
| Cable Assembly | Type N Connectors, 48 inches long | HP 11501A | P |
| Service Kit | Contents: <br> Display Section to Spectrum Analyzer Extender <br> Cable Assembly (HP 11592-60015) <br> Tuning Section to IF Section Interconnection Cable Assembly (HP 11592-60016) <br> Subminiature Female to BNC Male Test Cable, 3 each, 36 inches long (HP 11592-60001) <br> Subminiature Male to Subminiature Female Test Cable, 2 each, 8 inches long (HP 11592-60003) <br> Subminiature Female to Subminiature Female Test Cable, 2 each, 8 inches long (HP 11592-60002) <br> Extender Board Assembly, 15 pins, 30 conductors, for plug-in circuit boards (HP 11592-60011) <br> Fastener Assembly (2 each: HP 11592-20001 and HP 1390-0170) <br> Subminiature Jack-to-Jack Adapter (HP 1250-0827) <br> Wrench, open end, 15/16 inch (HP 8710-0946) <br> BNC Jack-to-OSM Plug Adapter (HP 1250-1200) <br> OSM Plug-to-Plug Adapter (HP 1250-1158) <br> Cable Assembly R and P Connector <br> (HP 11592-60013) | HP 11592A | A,T |



## SECTION II INSTALLATION

## 2-1. INTRODUCTION

2-2. This section explains how to prepare the HP Model 8554B Spectrum Analyzer RF Section for use. It explains initial inspection, how to install the RF Section in a mainframe, and storage and shipment.

## 2-3. INITIAL INSPECTION

$2-4$. Inspect the shipping container for damage. If the shipping container or cushioning material is damaged it should be kept until the contents of the shipment have been checked for completeness and the instrument has been checked mechanically and electrically. The contents of the shipment should be as shown in Figure 1-1, and procedures for checking electrical performance are given in Section IV. If the contents are incomplete, if there is mechanical damage or defect, or if the instrument does not pass the electrical performance test, notify the nearest Hewlett-Packard office. If the shipping container is damaged, or the cushioning material shows signs of stress, notify the carrier as well as the Hewlett-Packard office. Keep the shipping materials for carrier's inspection. The HP office will arrange for repair or replacement without waiting for claim settlement.

## 2-5. PREPARATION FOR USE

## CAUTION

Before applying power, check the rear panel on the Display Section for proper voltage selection.

## 2-6. Shipping Configuration

2-7. Since the RF and IF Sections are received separately, the plug-ins must be mechanically fitted together, electrically connected and inserted in display section or oscilloscope mainframe of the 140 -series. For mechanical and electrical connections, refer to Figure 2-1 and paragraph 2-16.

## 2-8. Power Requirements

2-9. Consumed power varies with the plug-ins used but is normally less than 225 watts. Line power
enters the Display Section or Mainframe, where it is converted to dc voltages, and then is distributed to the RF and IF Sections via internal connectors. The Display Section Operating and Service Manual covers such topics as available power cables, line power selection, bench operation, and rack mounting.

## 2-10. Mating Connectors

2-11. Mating connectors used with the Model 8554 B should be either 50 ohm-type BNC male or Type N male connectors that are compatible with US MIL-C-39012.

## 2-12. Operating Environment

$2-13$. The Operating environment should be within the following limitations:
a. Temperature: $0^{\circ} \mathrm{C}$ to $+55^{\circ} \mathrm{C}$
b. Humidity: $<95 \%$ relative
c. Altitude: $<15,000$ feet.

2-14. The Spectrum Analyzer uses a forced-air cooling system to maintain required operating temperatures within the instrument. The air intake and filter are located on the rear of the Display Section: air is exhausted through the side panel perforations. When operating the instrument, choose a location which provides at least three inches of clearance around the rear and both sides. Refer to the Display Section manual for maintenance instructions for the cooling system.

## 2-15. Interconnections

2-16. The RF and IF Sections are shipped separately; the plug-ins must be mechanically fitted together, electrically connected, and then inserted in the Display Section or mainframe. To make these connections refer to Figure 2-1 and proceed as follows:
a. Set the IF Section on a level bench. Locate slot near right rear corner of RF Section; also locate metal tab on IF Section that engages with this slot.
b. Grasp the RF Section near middle of chassis and raise until it is a few inches above the

IF Section.
c. Tilt RF Section until front is about 2 inches higher than the rear.
d. Engage assemblies in such a way that metal tab on the rear of the IF Section slips through the slot on RF Section.
e. With the preceding mechanical interface completed, gently lower RF Section until electrical plug and receptacle meet.
f. Position RF Section as required to mate the plug and receptacle. When plug and receptacle are properly aligned, only a small downward pressure is required to obtain a snug fit.
g. After the RF and IF Sections are joined mechanically and electrically, the complete assembly is ready to be inserted into the Display Section.
h. Pick up the RF/IF Sections and center in opening of Display Section. Push forward until assembly fits snugly into Display Section.
i. Push in front panel latch to securely fasten assembly in place.

2-17. To separate the RF/IF Sections from Display Section and to separate the RF Section from the IF Section, proceed as follows:
a. Push front panel latch in direction of arrow until it releases.
b. Firmly grasp the middle of latch flange and pull RF/IF Sections straight out.
c. Locate black press-to-release level near left front side of RF Section. Press this lever and simultaneously exert an upward pulling force on front edge of RF Section.
d. When the two sections separate at the front, raise RF Section two or three inches and slide metal tab at rear of IF Section out of the slot in which it is engaged.

## 2-18. LO Terminations

2-19. Two HP 11593A $50 \Omega$ Terminations are supplied with each HP 8554B. They should be
connected to the FIRST LO OUTPUT and THIRD LO OUTPUT connectors on the front panel.

## 2-20. STORAGE AND SHIPMIENT

## 2-21. Environment

2-22. The instrument should be stored in a clean, dry environment. The following environmental limitations apply to both storage and shipment:
a. Temperature: $-40^{\circ} \mathrm{C}$ to $+75^{\circ} \mathrm{C}$
b. Humidity: $<95 \%$ relative
c. Altitude: $<25,000$ feet.

## 2-23. Packaging

2-24. Original Packaging. Containers and materials identical to those used in factory packaging are available through Hewlett-Packard offices. If the instrument is being returned to Hewlett-Packard for servicing, attach a tag indicating the type of service required, return address, model number, and full serial number. Also, mark the container FRAGILE to assure careful handling. In any correspondence, refer to the instrument by model number and full serial number.

2-25. Other Packaging. The following general instructions should be used for repackaging with commercially available materials:
a. Wrap the instrument in heavy paper or plastic. (If shipping to a Hewlett-Packard office or service center, attach a tag indicating the type of service required, return address, model number, and full serial number.)
b. Use a strong shipping container. A doublewall carton made of 350 -pound test material is adequate.
c. Use enough shock-absorbing material (3 to 4 -inch layer) around all sides of the instrument to provide a firm cushion and prevent movement inside the container. Protect the control panel with cardboard.
d. Seal the shipping container securely.
e. Mark the shipping container FRAGILE to assure careful handling.


Figure 2.1. RF Section and IF Section Interconnections

# SECTION III OPERATION 

## 3-1. INTRODUCTION

3-2. This section provides complete operating instructions for the HP 8554B Spectrum Analyzer RF Section. It also provides a brief description of IF Section and Display Section controls; for a detailed description of these sections, refer to their manuals.

## 3-3. PANEL FEATURES

3-4. Front panel controls, indicators and connectors are shown and briefly described in Figure 3-6. Rear panel controls and connectors are shown and functionally described in Figure 3-7.

## 3-5. OPERATOR'S CHECKS

3-6. Upon receipt of the analyzer, or when any plug-in is changed, perform the operational adjustments listed in Figure 3-7. This procedure corrects for minor differences between units and ensures that the RF Section, IF Section and Display Section are properly matched.

3-7. If, after performing the operational adjustments, it is desired to further verify that the analyzer is operating correctly, perform the front panel check procedures listed in Table 4-1. Table 3-1 lists a few potential difficulties and their correction. Table 3-2 provides fuse replacement information.

## 3-8. OPERATING CONSIDERATIONS

## 3-9. RF Input

3-10. Input impedance is nominally 50 ohms from 100 kHz to 1250 MHz . A dc blocking capacitor (located between the RF INPUT jack and the analyzer's input attenuator) permits $R F$ measurements in the presence of dc levels as high as 50 volts. An external pad must be used for circuits that require a de return in their output.

## 3-11. Amplitude Characteristics

3-12. Absolute Voltage and Power Readings. There are two basic display modes, LOG and LINEAR. In LINEAR mode the analyzer measures voltage in mV or $\mu \mathrm{V}$. In LOG mode the analyzer measures power in dBm (into 50 ohms ).
$3-13$. The basic log display scale factor is 10 $\mathrm{dB} / \mathrm{div}$. With the 8552A IF Section this is selected
by setting LOG/LINEAR to LOG. With the 8552B IF Section this is selected by setting LOG/LINEAR to 10 dB LOG.

3-14. The LOG/LINEAR switch on the 8552B also has a 2 dB LOG position. In this position, the $\log$ display scale factor is $2 \mathrm{~dB} / \mathrm{div}$. (The LOG REF graticule on the CRT represents the same power level in both positions).

3-15. The absolute power level represented by the LOG REF graticule on the CRT is set by the LOG REF LEVEL and INPUT ATTENUATION controls. When LOG/LINEAR is set to LOG, the black " + " lamp lights to indicate that the absolute power level represented by the LOG REF graticule is the algebraic sum of:
a. The black dBm reading (under the lit index lamp ) on the LOG REF LEVEL switch, and
b. The black reading on the LOG REF LEVEL vernier.

3-16. When INPUT ATTENUATION is changed, the lit index lamp automatically switches position to indicate the change in absolute level of the LOG REF graticule line.

3-17. The LINEAR SENSITIVITY and INPUT ATTENUATION controls establish the linear display scale factor (volts/div). When LOG/LINEAR is set to LINEAR, the blue " $X$ " lamp lights to indicate that the display scale factor is the product of:
a. The blue volts/div reading (under the lit index lamp) on the LINEAR SENSITIVITY switch, and
b. The blue reading on the LINEAR SENSITIVITY vernier.

3-18. When INPUT ATTENUATION is changed, the lit index lamp automatically switches position to indicate the change in scale factor.

3-19. Sensitivity. The analyzer's noise level varies with bandwidth; the narrower the bandwidth, the lower the noise level, and the higher the sensitivity. However, the bandwidth chosen is usually a compromise because narrow bandwidths require narrow scan widths or slow scan times. (The DISPLAY UNCAL lamp will be unlit when the scan

Table 3-1. Operational Difficulties and Their Correction

| Difficulty | Corrections |
| :--- | :--- |
| Complete Failure or no vertical <br> deflection | Check fuses. Refer to Table 3-2. |
| Amplitude Calibration | Recalibrate in accordance with paragraph 2-16. |
| Insufficient horizontal gain | Adjust HORIZONTAL GAIN. |
| Incorrect horizontal position | Adjust HORIZONTAL POSITION. |
| Incorrect vertical position | Adjust VERTICAL POSITION. |
| No Scan | Set SCAN MODE to INT and SCAN TRIGGER to |
| Display skewed. | LINE. Check fuses. |
| Erratic tuning for narrow (blue-coded) | Adjust trace align. |
| scan widths. | Tune TUNING STABILIZER to OFF or use FINE |
| DISPLAY UNCAL light on | TUNE only. |
|  | Increase SCAN TIME or reduce SCAN WIDTH or in- |

Table 3-2. Fuse Information

| Designation/ Rating | Location | Indication |
| :---: | :---: | :---: |
| F1 1 Amp F2 1 Amp | Rear panel of IF Section. | If F1 opens because of overload, instrument may function normally until excessive current opens F2. (Generally, this will occur at turn on or during a sharp line surge). If a short develops in or if excessive current is drawn form the -10 volt supply, both F1 and F2 will open. When both fuses are open or when only F2 is open, the front panel indications are as follows: <br> a. No display. <br> b. All front panel indicator lamps, except SCANNING will function. |
| F481 3 Amp | See inside bottom cover display section. | No display. All indicator lamps function, except SCANNING. (SCANNING cannot be turned on in any scan mode.) |
| F461 1/2 Amp | Same as above | Nothing works, except POWER ON lamp. |
| F441 3/4 Amp | Same as above | Same as above |
| F421 1/4 Amp | Same as above | Same as above |
| F401 4 Amp Slo-Blo | Same as above | Nothing works. |

width and scan time are satisfactory for the bandwidth chosen).
$3-20$. For a given sensitivity requirement, the widest possible bandwidth should be used. Effective sensitivity can be increased by using the video filter. The filter averages the noise, making it possible to see small signals that would otherwise be obscured by noise peaks (see Figure 3-1).

3-21. Loss or gain in the signal path will affect sensitivity. For maximum sensitivity, set INPUT ATTENUATION to 0 dB . Sensitivity can be increased by the use of appropriate low-noise preamplifiers (see OPERATING ACCESSORIES in Section I).

3-22. Dynamic Range. Dynamic range is defined as the difference between the minimum and maximum signal levels that can be simultaneously displayed on the analyzer's CRT; with the $8554 \mathrm{~B} / 8552 / 140$ analyzer, this range is 70 dB .

a. Video Filter OFF

b. Video Filter 100 Hz

Figure 3-1. Increasing Effective Sensitivity Using the Video Filter

However, in practice any analyzer's dynamic range has system noise (which depends upon bandwidth) as the lower limit, and the input signal level that causes spurious responses to appear above the noise level as the upper limit.

3-23. Since spurious responses are generated by overdriving the analyzer, the operator can use the INPUT ATTENUATION control to tell if the analyzer is overloaded. Simply increase the attenuation by 10 dB ; input signals will drop 10 dB while internal distortion responses will drop more than 10 dB . (See Figure 3-2). To ensure that harmonic and intermodulation responses are not visible on the display, keep level at input mixer* below that indicated in Figure 3-3. For example, if noise level is -105 dBm , the maximum input level for a spurious free display is -40 dBm (for a dynamic range of 65 dB ). If the noise level is -115 dBm , then the maximum input is -45 dBm and the dynamic range is 70 dB .
$3-24$. Although spurious responses may be generated for large input signals, the analyzer is still calibrated. Gain compression is less than 1 dB for a signal level to the input mixer* of -10 dBm . Therefore, with 20 dB of input attenuation, the analyzer remains absolutely calibrated for input levels as high as +10 dBm . Note, however, that the maximum input level under any circumstances is +13 dBm . A signal level greater than this may damage the input attenuators or mixers of the analyzer. External pads, or directional couplers can be used to reduce a high level signal to a lower level compatible with the analyzer.
$3-25$. Residual and leakage responses are signals that appear on the display when there is no input to the analyzer. These signals may be identified by simply disconnecting the input. When working in high RF environments, stray signals can also be picked up by leakage through the input cable. Signals such as these can generally be eliminated by using cables with better shielding.

3-26. Amplitude Accuracy. Amplitude accuracy depends upon the measurement technique used. The factors involved in amplitude accuracy measurements are listed below.

|  | LOG | LINEAR |
| :--- | :---: | :---: |
| Frequency Response <br> (Flatness) | $\pm 1 \mathrm{~dB}$ | $\pm 12 \%$ |
| INPUT ATTENUA- <br> TOR (Excluding <br> Flatness) |  |  |

* Signal level at input mixer $=($ signal level at INPUT $)-($ INPUT attenuation)

a. INPUT ATTENUATION $=0 \mathrm{~dB}$ LOG REF LEVEL $=0 \mathrm{dBm}$

b. INPUT ATTENUATION $=10 \mathrm{~dB}$ LOG REF LEVEL $=0 \mathrm{dBm}$

c. INPUT ATTENUATION $=20 \mathrm{~dB}$ LOG REF LEVEL $=0 \mathrm{dBm}$


Figure 3-3. Dynamic Range

[^2]|  | LOG | LINEAR |
| :---: | :---: | :---: |
| Calibrator | $\pm 0.3 \mathrm{~dB}$ | $\pm 3.5 \%$ |
| LOG REF LEVEL• <br> LINEAR <br> SENSITIVITY | $\pm 0.2 \mathrm{~dB}$ | $\pm 2.3 \%$ |
| LOG REF LEVEL* <br> Vernier | $\pm 0.1 \mathrm{~dB}$ | $\pm 1.2 \%$ |
| Switching between bandwidths ( $20 /^{\circ} \mathrm{C}$ ) | $\pm 0.5 \mathrm{~dB}$ | $\pm 5.8 \%$ |
| Amplitude Stability | $\pm 0.07 \mathrm{~dB} /{ }^{\circ} \mathrm{C}$ | $\pm 0.6 \% /{ }^{\circ} \mathrm{C}$ |
| Amplitude Display | $\pm 0.25 \mathrm{~dB} / \mathrm{dB}$ but not more than $\pm 1.5 \mathrm{~dB}$ over full 70 dB display range. | $\pm 2.8 \%$ of full 8 division deflection. |

3-27. For general use, the amplitude measurement accuracy is determined by the frequency response flatness and the amplitude display accuracy. It is evident from the above chart that better accuracy can be obtained by moving the signal to be measured to the same display level as the calibrator, thus eliminating the amplitude display error of $\pm$ 1.5 dB . The frequency response error of 1 dB can also be eliminated by calibrating the display with a laboratory standard at the frequency of interest. To ensure maximum accuracy, the analyzer should be calibrated in the bandwidth used to make the measurement; also, the DISPLAY UNCAL light must be off.

## 3-28. Frequency Characteristics

3-29. Stabilization. The analyzer is stabilized when the TUNING STABILIZER is set to ON (up), and the PER DIVISION scan is set to any one of the scan widths that are blue color-coded (2 through 200 kHz per division). Coarse tuning the analyzer in the stabilized mode will cause the analyzer to jump lock points. When the analyzer is stabilized use FINE TUNE only.

3-30. Frequency Scanning. Any segment of the frequency range can be presented on the CRT. Any of the three scanning modes can be selected from the front panel: full scan ( $0-1250 \mathrm{MHz}$ ), PER DIVISION scan (scan width selectable from 1000 MHz to 20 kHz ), and ZERO scan. Switching be-

[^3]tween scanning modes makes it possible to periodically monitor the entire spectrum while individual signals are examined in detail as is done in surveillance and circuit-stability analysis. In the following paragraphs, the three scanning modes are described in their usual order of use.

3-31. $0-1250 \mathrm{MHz}$ Scan: for a quick view of the entire frequency range, the preset, $0-1250 \mathrm{MHz}$ scan mode can be selected. A marker identifies the center frequency of the CRT display for the other scan modes (PER DIVISION and ZERO) and since the marker is inverted, it cannot be confused with input signals. The marker can be tuned to any frequency by setting the tuning dial. When the marker is centered on a particular signal, the signal amplitude will dip to a minimum.

3-32. SCAN WIDTH PER DIVISION: in the PER DIVISION scan mode, the scan width is selected by the front panel SCAN WIDTH control. Scan widths of 1000 MHz to $20 \mathrm{kHz}(100 \mathrm{MHz}$ to 2 kHz per division) are selectable by the PER DIVISION control. The ten-division scan is symmetrical about the center frequency selected by the FREQUENCY control. Since the scan widths are calibrated, frequency separation is read directly from the CRT. The spectrum may be scanned in small or large segments, depending upon the application.

3-33. ZERO Scan: in the ZERO scan mode, the analyzer is a fixed-tuned receiver and is manually tuned by the FREQUENCY control. Using this mode and the calibrated time base, the demodulated waveform of a signal can be examined in the time-domain.

3-34. Frequency Accuracy. The frequency accuracy of a Spectrum Analyzer is related to both the center frequency accuracy and the scan width accuracy. The absolute frequency of any displayed signal is a function of the center frequency accuracy. The precision of a frequency difference measurement is determined by the scan width accuracy.
$3-35$. Since the frequency scan is very linear, absolute frequency can be determined by comparing the unknown signal to a known reference frequency. (The HP 8406A Comb Generator provides a comb of reference signals spaced at 100, 10 and 1 MHz for reference purposes). The reference and unknown signals are simultaneously applied to the analyzer; both frequencies can then be read directly from the CRT. A typical example is shown in Figure 3-4; the reference frequency was derived from the HP 8406A. The unknown frequency was read from the analyzer dial as about 380 MHz . The comb reference is known to be $385 \mathrm{MHz} \pm 0.01 \%$ $( \pm 38.5 \mathrm{kHz})$ and the scan width is 50 kHz per
division (signal separation is $150 \mathrm{kHz} \pm 10 \%$, that is $150 \mathrm{kHz} \pm 15 \mathrm{kHz}$ ). Thus, the frequency of the unknown can be calculated to be 385.15 MHz $\pm 53.5 \mathrm{kHz}$. Greater detail on "Accuracy Frequency Measurements" is given in Application Note AN63D. There is also a procedure for making precision frequency measurements using the HP 8444 A Tracking Generator. (See the HP 8444A manual.)


F'igure 3-4. Using a Reference Signal to Measure An Unknown Frequency

3-36. Frequency Resolution. The resolution of the analyzer is a measure of its ability to separate two closely spaced signals. This characteristic is largely determined by the IF passband.

3-37. Two signals of equal amplitude can be resolved if their frequencies differ by more than the 3 dB bandwidth of the analyzer (Figure 3-5a). If the signals are unequal in amplitude, the frequency separation must be greater or the bandwidth must be narrower (Figure 3-5b). Figure 1-3 in Table 1-2 shows a typical resolution chart for the Spectrum Analyzer. For a given bandwidth setting the chart shows the frequency resolving capability of the analyzer for two signals. For instance, two signals differing by 40 kHz and 30 dB can be resolved using the 10 kHz bandwidth. Maximum resolution can be obtained with a narrower bandwidth. If in the preceding example, the bandwidth is reduced to 1 kHz , the frequency resolution is improved by a factor of eight - to 5 kHz . A more narrow bandwidth than that required to resolve the frequency of the viewed signals should not be used, since a longer scan time will be required to maintain a calibrated amplitude display. (The DISPLAY UNCAL light is off when the scan time is compatible with the scan width and bandwidth selected). Eight calibrated bandwidths from 100 Hz to 300 kHz permit selection of the optimum
bandwidth for maximum sensitivity, resolution, or a choice of scan width and scan time.

## 3-38. Variable Persistence and Storage Functions

3-39. With the 141T Display Section the operator can set trace persistence for a bright, steady trace that does not flicker, even on the slow sweeps required for narrow band analysis. The variable persistence also permits the display of low repetition rate pulses without flickering and, using the longest persistence, intermittent signals can be captured and displayed. The storage capability allows side-by-side comparison of changing signals.

a. Signals of Equal Amplitude: SCAN WIDTH PER DIVISION $=2 \mathrm{kHz}$; BANOWIOTH $=1 \mathrm{kHz}$

b. Signals Differing in Amplítude by 30 dB : SCAN WIDTH PER DIVISION $=2 \mathrm{kHz}$; BANDWIOTH $=0.3 \mathrm{kHz}$

Figure 3-5. Resolving Two Signals That Are Close Together

3-40. Persistence and Intensity. The persistence and intensity determine how long a written signal will be visible. Specifically, PERSISTENCE controls the rate at which a signal fades and INTENSITY controls the trace brightness as the signal is written. With a given PERSISTENCE setting, the actual time of trace visibility can be increased by greater INTENSITY. Since the PERSISTENCE control only sets the fade rate, it follows that a brighter trace will fade more slowly. Conversely, a display of low intensity will disappear more rapidly. The same principle applies to a stored display of high and low intensity.

## CAUTION

Excessive INTENSITY will damage the CRT storage mesh. The INTENSITY setting for any sweep speed should just eliminate trace blooming with minimum PERSISTENCE setting.

3-41. Storage. The storage controls select the storage mode in which the CRT functions. In ERASE, the WRITING SPEED function is disconnected and all written signals are removed from the CRT. The STORE selector disconnects the WRITING SPEED and ERASE functions and implements signal retention at reduced intensity. In
the STORE mode, PERSISTENCE and INTENSITY have no effect.

3-42. Writing Speed. In the FAST mode, the fade rate is decreased, the entire screen becomes illuminated more rapidly, and the display is obscured. The effective persistence and storage time are considerably reduced.

## 3-43. Photographic Techniques

3-44. Excellent signal photography is possible when the Spectrum Analyzer is used with an oscilloscope camera and when proper techniques are employed. Both the HP 196B and the 197A Oscilloscope Cameras attach directly to the anlayzer's CRT bezel without adapters. Both cameras also have an Ultra-Violet light source that causes a uniform glow of the CRT phosphor. This gives the finished photograph a gray background that contrasts sharply with the white trace and the black graticule lines. Ulta-Violet illumination is normally used only when the CRT is of the non-storage and fixed persistence type (140T Display Section). For a storage or variable persistence CRT (141T Display Section), a uniform gray background can be obtained by simply taking the photograph in STORE rather than in VIEW.

## FRONT PANEL FEATURES

(1) FREQUENCY: coarse tunes CENTER FREQUENCY in SCAN WIDTH PER DIVISION and ZERO scan modes. In 0 to 1250 MHz scan mode it tunes inverted marker on CRT to indicate CENTER FREQUENCY for other scan modes.
(2) FINE TUNE: fine tunes CENTER FREQUENCY.
(3) TUNING STABILIZER: when set to up position, analyzer is phase locked if SCAN WIDTH is also set to ZERO or PER DIVISION and SCAN WIDTH PER DIVISION is set to any blue numeral. Use FINE TUNE to tune frequency. Coarse tune control (FREQUENCY) will cause analyzer to jump lock points ( 1 MHz jumps).
4) BANDWIDTH: selects resolution bandwidth of Spectrum Analyzer.
(5) CENTER FREQUENCY: dial indicates center frequency to which analyzer is tuned by FREQUENCY in SCAN WIDTH PER DIVISION and ZERO scan
modes (FINE TUNE does not move dial pointer). Indicates frequency of inverted marker in 0 - 1250 MHz scan mode.

6 SCAN WIDTH: selects frequency scan mode. PER DIVISION mode scans the analyzer symetrically about the CENTER FREQUENCY with a scan width set by the PER DIVISION control. 0-1250 MHz mode scans analyzer from 0 to 1250 MHz ; inverted marker and dial identify CENTER FREQUENCY for other scan modes. in ZERO scan mode analyzer becomes a fixed frequency receiver at the CENTER FREQUENCY.

1 PER DIVISION: selects the CRT horizontal calibration (frequency scale) in SCAN WIDTH PER DIVISION mode.
8 AMPL CAL: sets overall gain of analyzer for absolute amplitude calibration.
(9) INPUT ATTENUATION: adjusts the input signal level to the input mixer to maximize dynamic range.

## FRONT PANEL FEATURES

Set for -40 dBm at mixer (level at mixer $=$ level INPUT ATTENUATION).

10 DISPLAY UNCAL: warning light indicates that the CRT display is uncalibrated due to incompatible settings of SCAN WIDTH PER DIVISION, SCAN TIME PER DIVISION, BANDWIDTH, and VIDEO FILTER.

11 RF INPUT: 50 ohm female Type N coaxial input connector.


To prevent mixer burnout, attenuator damage, or both, the RF INPUT level should never exceed $1 \mathrm{Vrms}(+13 \mathrm{dBm})$ or $\pm 50 \mathrm{Vdc}$.

12 FIRST LO OUTPUT: 2050 MHz to 3300 MHz from YIG oscillator; female BNC connector. Terminate in 50 ohm load when not in use.
(13) THIRD LO OUTPUT: 500 MHz from third LO, female BNC connector. Terminate in 50 ohm load when not in use.

14 INDICATOR LAMPS: When " + " is lit, absolute level (in dBm ) of LOG REF graticule on CRT is sum of LOG REF LEVEL controls. When " X " is lit, linear scale factor (in mV or $\mu \mathrm{V}$ ) on CRT is product of LINEAR SENSITIVITY controls.

15 CAL OUTPUT: $30 \mathrm{MHz},-30 \mathrm{dBm}$ signal for amplitude calibration of analyzer; signal also contains harmonics: $60 \mathrm{MHz}, 90 \mathrm{MHz}$, etc.

16 PEN LIFT OUTPUT, TRIG/BLANK INPUT: provides +14 V pen lift signal during retrace (for use with X-Y recorders); present with SCAN MODE set to SINGLE or INT and SCAN TRIGGER set to VIDEO, LINE, or AUTO. Serves as input for external blanking signal ( -1.5 V ) with SCAN MODE set to EXT. Serves as input for external trigger signal (see 22 ) with SCAN MODE set to INT and SCAN TRIGGER set to EXT.

11 VERTICAL OUTPUT: provides output proportional to vertical deflection on CRT. Approximately 100 mV per major division with 100 ohm output impedance ( $0 \mathrm{~V}=$ bottom, $800 \mathrm{mV}=$ top ).
(18) SCAN IN/OUT: provides output voltage proportional to CRT horizontal deflection. 0 volts equals center
screen with 1 volt per division ( -5 to +5 V full screen). Output voltage available with SCAN MODE set to SINGLE, INT, and MAN (on 8552B). With SCAN MODE set to EXT, connector used as input for 0 to +8 V external scan signal.

19 DISPLAY ADJUST: the controls adjust deflection circuit gain and offset levels to calibrate CRT.

20 LOG REF LEVEL - LINEAR SENSITIVITY: these controls set the absolute amplitude calibration of CRT display. In LOG mode, sum of two control settings gives absolute level (in dBm ) of LOG REF graticule on CRT. In LINEAR mode, product of two control settings gives CRT scale factor in volts per division (in $\mu \mathrm{V}$ or mV ).

21 LOG/LINEAR: selects display mode, LOG (logarithmic) or LINEAR. Also selects LOG scale factor with 8552 B , either $10 \mathrm{~dB} / \mathrm{div}$ or $2 \mathrm{~dB} / \mathrm{div}$ (with 8552 A , scale factor is $10 \mathrm{~dB} / \mathrm{div}$ ). In LINEAR mode, scale factor is selected by LINEAR SENSITIVITY controls.

22 SCAN TRIGGER: selects synchronizing trigger when in the INT SCAN MODE.

AUTO: scan free runs.
LINE: scan synchronized to power line frequency.
EXT : scan initiated by external positive or negative pulses (2-20V normally negative) applied to TRIG/ BLANK INPUT (polarity set by switch in IF Section).
VIDEO: scan internal synchronized to envelope of RF input signal. Signal amplitude of 1.5 divisions peak-to-peak (min.) required on display section CRT.

23 SCAN MODE: selects scan source.
INT.: analyzer repetitively scanned by internally generated ramp; synchronization selected by SCAN TRIGGER. SCANNING lamp indicates time during which analyzer is being scanned.

EXT.: scan determined by externally applied 0 to +8 V signal at SCAN IN/OUT.

MAN: scan determined by MANUAL SCAN control; scan continuously variable across CRT in either direction. (Not available with 8552 A ).

SINGLE: single scan initiated by front panel pushbutton. SCANNING lamp indicates time during which analyzer is being scanned.

## FRONT PANEL FEATURES

Initiates or resets scan when SINGLE SCAN MODE is selected.

25 SCAN TIME PER DIVISION: selects time required to scan one major division on CRT display. Control acts as time base for time domain operation in ZERO scan.

26 VIDEO FILTER: post detection low pass filter for effective averaging of distributed signals such as noise. Bandwidths of $10 \mathrm{kHz}, 100 \mathrm{~Hz}$, and 10 Hz selectable; nominal bandwidth 400 kHz in OFF position. ( 10 Hz position not available with 8552 A ).

21 BASE LINE CLIPPER: allows blanking of the bright base line area of the CRT for better photography and improved display of transient phenomena.

28 MANUAL SCAN: controls spectrum analyzer horizontal scan in the MAN SCAN MODE. (Not available on 8552 A ).

29 CAL 10 V and 1 V : 10 V or 1 V square wave used to calibrate time domain plug-ins only.

30 FOCUS: focuses CRT spot for best definition.
31 BEAM FINDER: returns CRT trace to the center of the screen regardless of deflection potentials with time domain plug-ins only.

32 NON STORAGE, CONV: defeats the storage and variable persistence features of the CRT. Persistence is that of the standard P31 phosphor.

33 INTENSITY: adjusts the intensity of the trace on the CRT.

Excessive INTENSITY will damage the CRT
storage mesh. Whenever trace blooming occurs, turn INTENSITY down.
(34) ERASE: erases the CRT in the WRITING SPEED FAST or STD mode of operation. CRT ready to record immediately after erasure.

35 PERSISTENCE: adjusts the trace fade rate from 0.1 sec to more than 2 minutes in the WRITING SPEED FAST or STD modes of operation.

36 WRITING SPEED FAST, STD: these controls select the writing speed of the CRT in the PERSISTENCE mode of operation. The WRITING SPEED STD mode is almost always selected for spectrum analysis applications.

37 STORE TIME: controls the storage time and relative brightness of the display in the STORE mode of operation. Storage time more than 2 minutes at maximum brightness, more than 2 hours at minimum brightness.

38 STORE: stores the display on the CRT for extended viewing or photography. The CRT does not write in the STORE mode.

39 POWER: controls power to the mainframe and to both plug-ins.

40 ASTIG: adjusts the shape of the CRT spot.

41 TRACE ALIGN: used to adjust the CRT trace to align with the horizontal graticule lines.

42 CRT Graticule with LOG and LIN scales. LOG REF is the level used to reference the amplitude of displayed signals in the LOG display mode. LINEAR display amplitude is referenced from the baseline.

FRONT PANEL FEATURES


Figure 3-6. Front Panel Features (4 of 4)

## OPERATIONAL ADJUSTMENTS

## (1) input power

a. Set to correspond with available input voltage. (The instrument, as supplied, is fused for 120 volt, $50 / 60 \mathrm{~Hz}$ operation; if some other line voltage is used, refer to the Display Section service manual for power selection and fuse replacement procedures).
b. Connect line power cord to instrument jack and to a line power outlet.

## 2 ITENSITY MODULATION

Set INT/EXT switch to INT. (Set to EXT only if CRT Z axis is to be externally modulated - normally only used with 1400 series oscilloscope plug-ins).

## (3) FOCUS AND ASTIGMATISM

a. Make the following instrument control settings:
FREQUENCY . . . . . . . . . . . . . . . . . . . 30 MHz
FINE TUNE . . . . . . . Centered (1/1/2 turns from stop)
BANDWIDTH ........................ 100 kHz
SCAN WIDTH ................... PER DIVISION
PER DIVISION . . . . . . . . . . . . . . . . . . . . . 5 MHz
INPUT ATTENUATION .................. 10 dB
TUNING STABILIZER ................ On (up)
SCAN TIME PER DIVISION . . 10 MILLISECONDS
LOG REF LEVEL . . . . . . . . . . . . . . . . . -30 dBm
Vernier $\qquad$ ccw
LOG/LINEAR . . . . . . . . . . . . . . . . . . . 10 dB LOG
VIDEO FILTER . . . . . . . . . . . . . . . . . . . . . . . . OFF
SCAN MODE . . . . . . . . . . . . . . . . . . . . . . . . . INT
SCAN TRIGGER . . . . . . . . . . . . . . . . . . . . . AUTO
BASE LINE CLIPPER . . . . . . . . . . . . . . . . . . . . ccw
WRITING SPEED . . . . . . . . . . . . . . . . . . . . . . STD
PERSISTENCE . . . . . . . . . . . . . . . . . . . . . . MIN
INTENSITY . . . . . . . . . . . . . . . . . . . . . . 12 o'clock
POWER . . . . . . . . . . . . . . . . . . . . . . . . . . . . . ON
b. Adjust INTENSITY as needed. (Whenever blooming occurs on CRT, turn INTENSITY down). Set LOG REF LEVEL maximum counterclockwise. Using the VERTICAL POSITION control, bring the trace to the -40 dB graticule line.
c. Set SCAN TIME PER DIVISION to 10 SECONDS. Adjust FOCUS and ASTIG for the smallest round dot possible. Then set SCAN TIME PER DIVISION to 10 MILLISECONDS.

## (4) TRACE ALIGNMENT

Adjust TRACE ALIGN to set the trace parallel to the horizontal graticule lines.

## (5) HORIZONTAL POSITION AND GAIN

a. Alternately adjust HORIZONTAL GAIN and HORIZONTAL POSITION so that the trace just fills the horizontal graticule line.
b. Using the VERTICAL POSITION control, bring the trace to the bottom graticule line (ignore any slight misalignment of the trace).
c. Set LOG REF LEVEL to -30 dBm

## (6) VERTICAL POSITION AND GAIN

a. Connect CAL OUTPUT to RF INPUT.
b. Set SCAN WIDTH to $0-1250 \mathrm{MHz}$.
c. Tune FREQUENCY until inverted marker is below 30 MHz signal (see Display $a$ ).

## NOTE

Signal at far left graticule is LO feedthrough ( 0 Hz marker). First signal to right is 30 MHz , second signal to right is 60 MHz , etc. Use BASE LINE CLIPPER, if necessary, to prevent base line blooming.
d. Set SCAN WIDTH to PER DIVISION. Center 30 MHz signal with FREQUENCY (see Display b).
e. Reduce SCAN WIDTH PER DIVISION to 0.5 MHz ; keep signal centered with FREQUENCY (see Display c).
f. Set LOG REF LEVEL full counterclockwise. Adjust VERTICAL POSITION control to set trace to bottom graticule line, if necessary.
g. Adjust LOG REF LEVEL and Vernier to set signal peak to -60 dB graticule line.
h. Rotate LOG REF LEVEL 6 steps clockwise. Adjust VERTICAL GAIN until signal peak is at LOG REF graticule line.

## OPERATIONAL ADJUSTMENTS

i. Reduce SCAN WIDTH PER DIVISION to 50 kHz . Repeat steps $\mathrm{f}, \mathrm{g}$ and h until signal moves one major division for each step of LOG REF LEVEL control as shown on Display d.
(1) AMPLITUDE CALIBRATION
a. Set LOG REF LEVEL to -30 dBm and set


Display a. 30 MHz Calibrator Signal with SCAN WIDTH at 0-1250 MHz


Display b. 30 MHz Calibrator Signal with SCAN WIDTH PER DIVISION at 5 MHz

Vernier to 0 (full ccw). Adjust AMPL CAL until signal peak is at LOG REF graticule line.
b. Set LOG/LINEAR to LINEAR, and set LINEAR SENSITIVITY to $1 \mathrm{mV} / \mathrm{DIV}$ (Vernier should read 1 on blue scale - full ccw). Adjust AMPL CAL until signal peak is 7.1 divisions from bottom ( -30 dBm calibrator output is actually 7.07 mV across 50 ohms).


Display c. 30 MHz Calibrator Signal with SCAN WIDTH PER DIVISION at 0.5 MHz


Display d. Amplitude Calibration Steps in LOG Mode

OPERATIONAL ADJUSTMENTS


Figure 3-7. Operational Adjustments (3 of 3)

## SECTION IV PERFORMANCE TESTS

## 4-1. INTRODUCTION

4-2. This section provides instructions for performance testing the Model 8554B Spectrum Analyzer RF Section. The performance tests verify that the instrument meets the specifications listed in Table 1-1. Front panel checks for routine inspection are given in Table 4-1.
$4-3$. Perform the tests in procedural order with the test equipment called for, or with its equivalent. During the tests, all circuit boards, shields, covers and attaching hardware must be in place, and the RF and IF Sections must be installed in the Display Section. Allow the analyzer to warm up at least one hour before performing the tests.

## 4-4. EQUIPMENT REQUIRED

$4-5$. Test equipment and test equipment accessories for the performance tests (designated "P" in the "use" column) are specified in Tables 1-4 and 1-5. Equipment other than that listed may be used providing that it meets or exceeds the minimum specifications listed in the tables.

## 4-6. OPERATIONAL ADJUSTMENTS

4-7. Before proceeding to the performance tests, perform the operational adjustments specified in Figure 3-7 (in Section III). These adjustments correct for minor differences between units and ensure that the RF Section, IF Section and Display Section are properly calibrated.

## 4-8. FRONT PANEL CHECKS

$4-9$. The front panel checks provide a quick method for verifying that the RF Section is operating correctly. After performing the operational adjustments described in Figure 3-7, set the analyzer's controls as specified in Table 4-1 and perform the checks.

## 4-10. TEST SEQUENCE

4-11. The performance tests are suitable for incoming inspection, troubleshooting, and preventive maintenance. A test card for recording data is included at the back of this section.

## 4-12. Perform the tests in the following order:

a. Allow analyzer to warm up one hour.
b. Perform operational adjustments listed in Figure 3-7.
c. Perform front panel checks listed in Table 4-1.
d. Perform the performance tests in the order given.

## 4-13. PERFORMANCE TEST PROCEDURES

4-14. Each test is arranged so that the specification is written as is appears in Table 1-1. Next is a description of the test that includes any special instructions. Each test that requires test equipment has a test setup drawing and a list of required equipment.

Table 4-1. Front Panel Checks

| Function | Procedure | Result |
| :---: | :---: | :---: |
| Calibration | 1. Perform Operational Adjustments specified in Section III (Figure 3-7), then set analyzer as follows (CAL OUTPUT should be connected to RF INPUT): <br> FREQUENCY . . . 30 MHz <br> FINE TUNE . . . Centered <br> SCAN WIDTH <br> PER DIVISION <br> SCAN WIDTH PER <br> DIVISION . . . . . 5 MHz <br> INPUT ATTENUATION 20 dB TUNING STABILIZER . On SCAN TIME PER DIVISION <br> 5 MILLISECONDS <br> LOG/LINEAR . . . . LOG <br> LOG REF LEVEL -10 dBm <br> Vernier . . . . . . . . . ccw <br> VIDEO FILTER . . 10 kHz <br> SCAN MODE . . . . . . INT <br> SCAN TRIGGER . . AUTO <br> BASE LINE CLIPPER . ccw | 1. Analyzer calibrates normally. |
| Base Line Clipper | 2. Turn BASE LINE CLIPPER cw . | 2. At least the bottom two divisions should be blank. |
| Scan | 3. Return clipper to ccw. <br> 4. SCAN TIME PER DIVISION across its range | 4. Scan should occur in all positions. |
| Scan Width | 5. Set to 5 MILLISECONDS. <br> 6. Turn SCAN WIDTH PER DIVISION to 10 MHz . | 6. LO feedthrough, 30 MHz signal and second harmonic visible. |
|  | 7. Center CAL OUTPUT ( 30 MHz ) signal on display. <br> 8. Reduce SCAN WIDTH PER DIVISION to 200 kHz ; use FINE TUNE to center display. |  |
| Phase Lock | 9. Carefully turn FREQUENCY. | 9. Signal jumps to left or right edge of CRT ( $\pm 1 \mathrm{MHz}$ ). (This corresponds to the 1 MHz reference oscillator in the automatic phase control circuit). |
|  | 10. Turn TUNING STABILIZER to OFF; use FREQUENCY to center display | 10. Signal should jump $\leqslant 1 \mathrm{MHz}$ when TUNING STABILIZER is turned off. |

Table 4-1. Front Panel Checks (cont'd)

| Function | Procedure | Result |
| :---: | :--- | :--- |
|  | 11. Turn TUNING STABILIZER <br> on, use FINE TUNE to <br> center display. | 11. Signal should jump $\leqslant 1 \mathrm{MHz}$. |
| Bandwidth <br> and Display <br> Uncal Light | 12. Reduce BANDWIDTH and SCAN <br> TIME PER DIVISION using <br> FINE TUNE to center display. <br> 13. Return BANDWIDTH to 10 kHz and <br> SCAN TIME PER DIVISION to 5 | 12. Display should be stable and <br> viewable so long as DISPLAY <br> UNCAL is unlit. |
| MILLISECONDS. |  |  |

## PERFORMANCE TESTS

## 4-15. Input Impedance

SPECIFICATION: 50 Ohms nominal. Reflection coefficient $<0.3$ (1.85 SWR with INPUT ATTENUATION $\geqslant 10 \mathrm{~dB}$ ).

DESCRIPTION: The Spectrum Analyzer RF input impedance is verified by measuring the return loss in a 50 -ohm system as the RF input is swept by an external source from $100-1250 \mathrm{MHz}$. The analyzer is checked with input attenuation settings of 10 and 20 dB .


Figure 4-1. Return Loss Test Setup

## EQUIPMENT:

Sweep Oscillator ..... HP 8690B
Oscilloscope ..... HP 180A/1801A/1821A
Dual Directional Coupler ..... HP 778DCrystal Detector (2)HP 423A10 dB AttenuatorHP 8491 A OP 010

## PROCEDURE:

1. Connect the test setup as shown in Figure 4-1 and make the following control settings:

## Spectrum Analyzer:

INPUT ATTENUATION . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . 10 dB

## 8690B/8699B:

FUNCIION . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . START/STOP
SWEEP SELECTOR . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . AUTO
START/CW . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . 0.1 GHz
STOP/ $\Delta \mathrm{F}$ 1.25 GHz

ALC Depressed (on)
AMPLITUDE MOD All Released (off)

## 4-15. Input Impedance (cont'd)

$$
\text { SWEEP TIME (SEC) . . . . . . . . . . . . . . . . . . . . . . . . . . . } 1 \text { - } 0.1
$$

VERNIER . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . Full cw
RANGE . . . . . . . . . . . . . . . . . . . . . . . . . . . 0.1 - 2 GHz
POWER LEVEL . . . . . . . . . . . . . . . . . . . . . . . . . . . . . 10
180A/1801A/1821A:
DISPLAY
INPUT
POLARITY
VOLTS/DIV .005
2. Adjust the oscilloscope EXT SENS and HORIZONTAL POSITION to give a full ten division horizontal CRT deflection.
3. Reduce the sweeper POWER LEVEL until UNLEVELED indicator is not lit.
4. With the analyzer RF INPUT not connected, observe the 10 dB reference return loss represented by the CRT vertical deflection.
5. Remove the 10 dB pad, but reconnect the crystal detector to the directional coupler. Connect the directional coupler output to the analyzer RF INPUT.
6. Measure the return loss of the analyzer by observing the CRT display. The vertical deflection should be less than the reference level established in step 4.

Return loss: 10 dB $\qquad$
7. Repeat step 6 with analyzer INPUT ATTENUATION set to 20 dB through 50 dB .

Return loss: 10 dB $\qquad$

## PERFORMANCE TESTS

## 4-16. Average Noise Level

SPECIFICATION: $<-102 \mathrm{dBm}$ with 10 kHz IF bandwidth.
DESCRIPTION: Average noise level is checked by observing the average noise power level of the analyzer with the instrument vertically calibrated and no signal input. The test is made using a 10 kHz IF bandwidth.

## PROCEDURE:

1. Check the analyzer to make sure it is vertically calibrated. Refer to Figure 3-7.
2. Make the following analyzer control settings:

3. Observe the average noise power level on the CRT. It should be lower than -102 dBm as shown in Figure $4-2$ as FREQUENCY is tuned from 1250 MHz to 100 kHz . Make sure the LOG REF LEVEL Vernier is set at 0 during the measurement.


Figure 4-2. Average Noise Level

## 4-17. Frequency Response

## SPECIFICATION: $\pm 1 \mathrm{~dB} ; 100 \mathrm{kHz}$ to 1250 MHz .

DESCRIPTION: A very flat signal source is applied to the RF INPUT of the spectrum analyzer. As the source is slowly tuned across the spectrum analyzer's frequency range, the analyzer CRT display is observed in the LINEAR mode for amplitude flatness versus frequency. This test is performed in two segments; 100 MHz to 1250 MHz and 100 kHz to 100 MHz .


Figure 4-3. Frequency Response Test: 100. 1250 MHz

## EQUIPMENT:



## PROCEDURE:

1. To check the analyzer's frequency response from 100 MHz to 1250 MHz , connect the test setup shown in Figure 4-3. Make the following control settings:

## Spectrum Analyzer:

FREQUENCY ..... 0
BANDWIDTH ..... 300 kHz
SCAN WIDTH ..... $0-1250 \mathrm{MHz}$
INPUT ATTENUATION ..... 20 dB
LINEAR SENSITIVITY $1 \mathrm{mV} /$ DIV
8690B/8699B:FUNCTIONSTART/STOP
SWEEP SELECTOR ..... CW
START/CW1 GHzALCDepressed (On)
AMPLITUDE MOD All released (Off)

## PERFORMANCE TESTS

## 4-17. Frequency Response (cont'd)

2. Adjust the HP 8699B RF Unit frequency range to $0.1 \mathrm{GHz}-2 \mathrm{GHz}$ and adjust POWER LEVEL for a 7 mV ( 7 division) CRT display.
3. Tune the HP 8690 B START/CW from 0.1 GHz to 1.25 GHz and note the frequency at which the signal level is maximum. Readjust VERNIER for 8 mV at this frequency.
4. Slowly tune START/CW from 0.1 GHz to 1.25 GHz . Signal amplitude should be within:
6.3 $\qquad$ 8.0 div


Figure 4-4. Frequency Response Test: 100 kHz to 100 MHz
5. Note the signal amplitude at 0.1 GHz .
6.3 $\qquad$ 8.0 div
6. To check the frequency response from 100 kHz to 100 MHz , connect the test shown in Figure 4-4. Make the following control settings:

## Spectrum Analyzer:

FREQUENCY
50 MHz
BANDWIDTH
100 kHz
SCAN WIDTH
PER DIVISION
SCAN WIDTH PER DIVISION
10 MHz
INPUT ATTENUATION
10 dB
SCAN TIME PER DIVISION . . . . . . . . . . . . . . . . . . 5 MILLISECONDS
LOG/LINEAR
LINEAR
LINEAR SENSITIVITY $1 \mathrm{mV} / \mathrm{DIV}$
VIDEO FILTER
OFF
SCAN MODE INT
SCAN TRIGGER LINE

8601A
FREQUENCY
100 MHz
RANGE 110
SWEEP CW
1 kHz MOD OFF
OUTPUT LEVEL
$-30 \mathrm{dBm}$
7. Adjust the HP 8601A VERNIER for signal amplitude as measured in step 5.

## PERFORMANCE TESTS

## 4-17. Frequency Response (cont'd)

8. Slowly tune HP 8601A FREQUENCY from 100 MHz to 100 kHz (switch HP 8601A RANGE to 11 to check 1 MHz to 100 kHz ). Signal amplitude should be within:
$100 \mathrm{kHz}-100 \mathrm{MHz}$ :
6.3 $\qquad$ 8.0 div

## 4-18. Spurious Responses

SPECIFICATION: All image and out-of-band mixing responses, harmonic and intermodulation distortion products are more than 65 dB * below a -40 dBm signal at the input mixer.

DESCRIPTION: To verify spurious response level, the two-tone method of measuring intermodulation distortion is used. The outputs of two signal generators, tuned 50 kHz apart, are applied to the analyzer INPUT. Their levels are adjusted for a total power level of -40 dBm at the analyzer's input mixer. Second and third order intermodulation products are then measured on the CRT.


Figure 4-5. Spurious Responses Test Setup

## EQUIPMENT:

HF Signal Generator . . . . . . . . . . . . . . . . . . . . . . . . . . HP 606B
VHF Signal Generator
HP 608F

## PROCEDURE:

1. Connect the test setup shown in Figure 4-5 and set the analyzer as follows:

[^4]
## PERFORMANCE TESTS

4-18. Spurious Responses (cont'd)
INPUT ATTENUATION . . . . . . . . . . . . . . . . . . . . . . . . . . 0 dB
SCAN WIDTH . . . . . . . . . . . . . . . . . . . . . . . . . PER DIVISION
SCAN WIDTH PER DIVISION . . . . . . . . . . . . . . . . . . . . . 0.5 MHz
BASE LINE CLIPPER . . . . . . . . . . . . . . . . . . . . . . . . . . . ccw
SCAN TIME PER DIVISION . . . . . . . . . . . . . . . . . . . . 0.1 SECONDS
LOG/LINEAR . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . LOG
LOG REF LEVEL . . . . . . . . . . . . . . . . . . . . . . . . . -40 dBm
LOG REF LEVEL Vernier . . . . . . . . . . . . . . . . . . . . . . 0 dB (ccw)
VIDEO FILTER . . . . . . . . . . . . . . . . . . . . . . . . . . . OFF
SCAN MODE . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . INT
SCAN TRIGGER . . . . . . . . . . . . . . . . . . . . . . . . . . . . AUTO
2. Adjust the HP 608 F for a $10 \mathrm{MHz}, \mathrm{CW},-43 \mathrm{dBm}$ signal ( $\mathrm{f}_{2}$ ); adjust the HP 606 B for a $9.95 \mathrm{MHz}, \mathrm{CW}$, -43 dBm signal ( $\mathrm{f}_{1}$ ). Center the signals (they will appear as one) on the CRT.
3. Reduce SCAN WIDTH PER DIVISION to 50 kHz , keeping the signals centered with FINE TUNE. Adjust the signal generators so that the signals are separated by 50 kHz (one division) and are 3 dB below the LOG REF graticule line (see Figure 4-6).


Figure 4-6. Intermodulation Distortion Products
4. Check for third order intermodulation products at 10.050 MHz and at 9.900 MHz . They should be below $-65 \mathrm{~dB}(-105 \mathrm{dBm})$ on the CRT.

Third Order: $\qquad$ $-65 \mathrm{~dB}$
5. Tune the analyzer to 19.95 MHz and check for a second order intermodulation product (it will be between the signal generator second harmonics). It should be below $-65 \mathrm{~dB}(-105 \mathrm{dBm})$.

Second Order: $\qquad$ $-65 \mathrm{~dB}$

## NOTE

Signal generators exhibit harmonic distortion, typically about 35 dB below fundamental level. Harmonic distortion will occur at multiples of 9.950 and 10 MHz . Care must be taken not to confuse harmonic distortion produced by the source with intermodulation distortion produced by the input mixer.

## PERFORMANCE TESTS

## 4-19. Residual Responses

SPECIFICATION (no signal present at input): With input attenuation at $0 \mathrm{~dB}:<-100 \mathrm{dBm}$.
DESCRIPTION: Signals present on the display with no input are called residual responses. To measure residual responses, a reference is selected so that -100 dBm is easily determined. The display is searched for residual responses under the various test conditions called out.

## PROCEDURE:

1. Set the analyzer controls as follows:

| FREQUENCY | 250 MHz |
| :---: | :---: |
| FINE TUNE | Centered |
| BANDWIDTH | 10 kHz |
| INPUT ATTENUATION | 0 dB |
| SCAN WIDTH . . . . . PER | IVISION |
| SCAN WIDTH PER DIVISION | 50 MHz |
| BASE LINE CLIPPER | Max cew |
| SCAN TIME PER |  |
| DIVISION . . . . . 10 | SECONDS |
| LOG REF LEVEL | $-60 \mathrm{dBm}$ |
| LOG REF LEVEL Vernier |  |
| VIDEO FILTER | 100 Hz |
| SCAN MODE | INT |
|  |  |

## NOTE

Ignore Display Uncal light temporarily.


Figure 4-7. Residual Response Test
2. Terminate the RF INPUT jack in 50 ohms.
3. Observe the display as the analyzer scans from 0 to 500 MHz . The average noise level should be less than -100 dBm , and no residual responses should occur. Figure 4-7 represents a scan with no residual responses and with the average nose level indicated.

Residual Responses: $100 \mathrm{kHz}-500 \mathrm{MHz} \quad-100 \mathrm{dBm}$
4. If residual responses appear at an apparent level between -105 dBm and -100 dBm , center FREQUENCY about the residual and reduce SCAN WIDTH PER DIVISION to 10 MHz (DISPLAY UNCAL should become unlit). Again note the residual response level; it should remain below -100 dBm .
_-100 dBm
5. Repeat step 3 with original SCAN WIDTH PER DIVISION setting and step 4 with FREQUENCY at 750 MHz and 1000 MHz .

Residual Responses: $500-1000 \mathrm{MHz}$ $\qquad$ $-100 \mathrm{dBm}$
Residual Responses: $750-1250 \mathrm{MHz}$ $\qquad$ $-100 \mathrm{dBm}$

## PERFORMANCE TESTS

## 4-20. Noise Sidebands

SPECIFICATION: More than 70 dB below CW signal, 50 kHz or more away from signal for a 1 kHz IF Bandwidth.

DESCRIPTION: A stable -40 dBm CW signal is applied to the spectrum analyzer and displayed on the CRT. The amplitude of the noise associated sidebands and unwanted responses close to the signal are measured.


Figure 4-8. Noise Sideband Test

## EQUIPMENT:

Signal Generator . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . HP 608F

## PROCEDURE:

1. Connect the test setup shown in Figure 4-8 and make the following control settings:

## Spectrum Analyzer:

FREQUENCY . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . 400 MHz
BANDWIDTH 1 kHz
SCAN WIDTH PER DIVISION
SCAN WIDTH PER DIVISION 0.5 MHz

SCAN TIME PER DIVISION 10 MILLISECONDS
INPUT ATTENUATION 0 dB
LOG/LINEAR LOG
LOG REF LEVEL $-40 \mathrm{dBm}$
TUNING STABILIZER On
VIDEO FILTER OFF
SCAN MODE INT
SCAN TRIGGER . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . AUTO
608F:
MODULATION
CW
FREQUENCY RANGE . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . E
MEGACYCLES . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . 400
ATTENUATION . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . -40 dBm
2. Tune the analyzer to center the signal on the CRT display.

## PERFORMANCE TESTS

## 4-20. Noise Sidebands (cont'd)

3. Keeping the display centered, reduce SCAN WIDTH PER DIVISION to 20 kHz . Reduce SCAN TIME PER DIVISION to 0.2 SECONDS and set VIDEO FILTER to 100 Hz . Then adjust HP 608 F output vernier so that the signal peak is at the LOG REF graticule line.
4. Observe the noise level 2.5 divisions (i.e. 50 kHz ) or greater away from the signal. The average noise level should be at least seven divisions below the signal peak (i.e. below the -70 dB graticule line):

7 div

## 4-21. Scan Width Accuracy

SPECIFICATION: Frequency error between two points on the display is less than ten percent of the indicated separation.
DESCRIPTION: Wide scan widths are checked directly using a comb generator. Narrow scan widths are checked using a comb generator modulated by an audio oscillator. Comb generator frequency components line up opposite graticule lines, and the amount of error is measured at the +3 graticule line.


Figure 4-9. Scan Width Accuracy Tests $100 \mathrm{MHz} /$ Div

## EQUIPMENT:

Comb Generator
HP 8406A
Audio Oscillator
HP 200 CD

## PROCEDURE:

1. Connect the test setup in Figure 4-9 and make the following control settings:

## Spectrum Analyzer:

FREQUENCY . . . . . . . . 600 MHz
BANDWIDTH . . . . . . . . 300 kHz
SCAN WIDTH . . . . . PER DIVISION SCAN WIDTH PER DIVISION 100 MHz INPUT ATTENUATION . . . . 0 dB SCAN TIME PER DIVISION

| LOG/LINEAR | G |
| :---: | :---: |
| LOG REF LEVEL | -20 dBm |
| VIDEO FILTER | OFF |
| SCAN MODE | INT |
| SCAN TRIGGER | AUTO |



Figure 4-10. Scan Width Accuracy Measurement

## PERFORMANCE TESTS

## 4-21. Scan Width Accuracy (cont'd)

VIDEO FILTER . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . OFF
SCAN MODE . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . INT
SCAN TRIGGER . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . AUTO

## 8406A:

COMP FREQUENCY - MC
100 MC
INTERPOLATION AMP OFF
OUTPUT AMPLITUDE 3 o'clock
2. With control settings as in step 1 above, a comb signal occurs every 100 MHz on the display (see Figure $4-10$ ). Tune FREQUENCY and FINE TUNE to line up a comb signal with the far left graticule line.
3. Measure the amount of error in divisions that the comb signal deviates from the +3 graticule line. The comb signal should occur on the +3 line $\pm 0.8$ division.

$$
+2.2 \ldots+3.8 \mathrm{div}
$$

4. Repeat steps 2 and 3 with SCAN WIDTH PER DIVISION set to 10 MHz and a comb frequency of 10 MHz.
$\qquad$ +3.8 div
5. Repeat steps 2 and 3 with SCAN WIDTH PER DIVISION set to 1 MHz , BANDWIDTH at 10 kHz and a comb frequency of 1 MHz .
$+2.2$ $\qquad$ +3.8 div


Figure 4-11. Scan Width Accuracy Test: 50 kHz/Div
6. To test the 50 kHz SCAN WIDTH PER DIVISION setting, connect the test setup shown in Figure 4-11. Set controls as follows:

## Spectrum Analyzer:

```BANDWIDTH3 kHz
```

SCAN TIME PER DIVISION ..... 10 MILLISECONDS
SCAN WIDTH PER DIVISION ..... 50 kHz

## PERFORMANCE TESTS

4-21. Scan Width Accuracy (cont'd)
200CD:
RANGEX1K
FREQUENCY ..... 50 kHz
AMPLITUDE ..... 3 o'clock
8406A:
COMB FREQUENCY - MC ..... 10 MC
7. Maximize the comb signal amplitudes using the comb generator and audio oscillator output amplitude controls.
8. With controls set as in step 6 above, a comb signal occurs every 50 kHz on the display. Turn FINE TUNE to line up a comb signal with the far left graticule line.
9. Measure the amount of error, in divisions, that the comb signal deviates from the +3 graticule line. The comb signal should occur on the +3 line $\pm 0.8$ division.

$$
+2.2 \ldots \ldots+3.8 \text { div }
$$

10. To test the 20 kHz SCAN WIDTH PER DIVISION position, set the test equipment as follows:

## Spectrum Analyzer:

```
BANDWIDTH
3 kHz
```

SCAN TIME PER DIVISION
5 MILLISECONDS
SCAN WIDTH PER DIVISION 20 kHz

200CD:
RANGEX1K
FREQUENCY 20 kHz
AMPLITUDE 3 o'clock

## 8406A:

COMB FREQUENCY - MC 10 MC
11. With the control settings as in step 10 above, a comb signal occurs every 20 kHz on the display. Turn FINE TUNE to line up a comb signal with the far left graticule line.
12. Measure the amount of error, in divisions, that the comb signal deviates from the +3 graticule line. The comb signal should occur on the +3 line $\pm 0.8$ division.
$+2.2$ $\qquad$ +3.8 div

## PERFORMANCE TESTS

## 4-22. Center Frequency Accuracy

SPECIFICATION: The dial indicates the display center frequency within 10 MHz .
DESCRIPTION: Center frequency accuracy is verified by displaying test singals of known frequency accuracy. Test signals are the fundamental and harmonics of a 100 MHz comb generator.


Figure 4-12. Center Frequency Accuracy Test Setup

## EQUIPMENT:

Comb Generator
HP 8406A

## PROCEDURE:

1. Connect the equipment as shown in Figure 4-12. Make the following control settings:

## Spectrum Analyzer:

FREQUENCY . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . 100 MHz
BANDWIDTH . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . 30 kHz
SCAN WIDTH . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . PER DIVISION
SCAN WIDTH PER DIVISION . . . . . . . . . . . . . . . . . . . . . . . . . . . 10 MHz
INPUT ATTENUATION . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . 10 dB
SCAN TIME PER DIVISION . . . . . . . . . . . . . . . . . . . . . 10 MILLISECONDS
LOG/LINEAR LOG
LOG REF LEVEL . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . -30 dBm
VIDEO FILTER . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . OFF
SCAN MODE . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . INT
SCAN TRIGGER . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . LINE
8406A:

$$
\text { COMB FREQUENCY - MC . . . . . . . . . . . . . . . . . . . . . . . . . . . . . } 100 \text { MC }
$$

OUTPUT AMPLITUDE
2. Tune FREQUENCY to 100 MHz ; a comb signal should be displayed $\pm 1$ division of center graticule line.
3. Tune FREQUENCY to the remaining dial calibration points to verify accuracy.

| a. | 200 MHz | -1 | +1 div | g . | 800 MHz | -1 | +1 div |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| b. | 300 MHz | -1 | +1 div | h. | 900 MHz | -1 | +1 div |
| c. | 400 MHz | -1 | +1 div | i. | 1000 MHz | $-1$ | +1 div |
| d. | 500 MHz | -1 | +1 div | j. | 1100 MHz | $-1$ | +1 div |
| e. | 600 MHz | -1 | +1 div | k. | 1200 MHz | -1 | +1 div |
| f. | 700 MHz | -1 | +1 div |  |  |  |  |

## PERFORMANCE TESTS

## 4-23. Local Oscillator Stability and Residual Frequency Modulation

SPECIFICATION:
Stabilized: less than 100 Hz peak-to-peak. Unstabilized: less than 10 kHz peak-to-peak.
DESCRIPTION: The linear portion of the analyzer IF filter skirt is used to slope detect low-order residual FM. The analyzer is stabilized, and the detected FM is displayed in the time domain.


Figure 4-13. Stability Test for Local Oscillator

## EQUIPMENT:

Comb Generator
HP 8406A

## PROCEDLRE:

1. Connect the test setup as shown in Figure $4-13$ and make the following control settings:

## Spectrum Analyzer:

FREQUENCY . . . . . . . . . . . 100 MHz
BANDWIDTH ................. 1 kHz
SCAN WIDTH . . . . . . . PER DIVISION
SCAN WIDTH PER DIVISION . . 2 kHz
INPUT ATTENUATION ........ . . 0 dB
SCAN TIME PER DIVISION
................. . . 50 MILLISECONDS
LOG/LINEAR . . . . . . . . . . . . . .LINEAR
LINEAR SENSITIVITY . . . . . . . Step 2
TUNING STABILIZER . . . . . . . . . . ON
VIDEO FILTER . . . . . . . . . . . . . . OFF
SCAN MODE . . . . . . . . . . . . . . . . INT
SCAN TRIGGER ............. AUTO
8406A:
COMB FREQUENCY - MC . . . 100 MC
OUTPUT AMPLITUDE . . . . . 3 o'clock
2. Adjust LINEAR SENSITIVITY and its vernier for a full eight-division display.
3. Refer to Figure 4-14. Tune FREQUENCY so that the upward slope of the display intersects the CENTER FREQUENCY graticule line one division from the top.


Figure 4-14. Demodulation Sensitivity Measurement

## PERFORMANCE TESTS

## 4-23. Local Oscillator Stability and Residual Frequency Modulation (cont'd)

4. Note where the slope intersects the middle horizontal graticule line:

Horizontal Displacement: $\qquad$ divisions
5. Use the horizontal displacement to calculate demodulation sensitivity.
a. Convert the horizontal displacement (divisions) into hertz.

Example: ( 2 kHz SCAN WIDTH $\times(0.2 \mathrm{div})=400 \mathrm{~Hz}$
b. Calculate demodulation sensitivity by dividing the vertical displacement in divisions into the horizontal displacement in Hz :
Example: $\frac{400 \mathrm{~Hz}}{3 \text { divisions }}=133 \mathrm{~Hz} /$ div
6. Turn SCAN WIDTH to ZERO scan. Set FINE TUNE for a response level within the calibrated three division range (one division from the top to the center horizontal graticule line).
7. Measure the peak-to-peak deviation, and multiply it by the demodulation sensitivity obtained in step 5b above.

Example: 0.5 div pk-pk signal deviation $\mathrm{x} 133 \mathrm{~Hz} / \mathrm{div}=66.5 \mathrm{~Hz}$ Residual FM
_ 100 Hz peak-to-peak
8. To measure unstabilized residual FM, repeat the test with the following control settings:

9. Calculate demodulation sensitivity as in steps 2 through 5 .
10. Switch to ZERO scan, TUNING STABILIZER to OFF, and set FINE TUNE so that the display occurs in the calibration three-division range (one division from the top, to the center horizontal graticule line).
11. Measure the vertical displacement and multiply it by the demodulation sensitivity obtained in step 9 above.
$\qquad$ 10 kHz peak-to-peak

Table 4-2. Performance Test Record
Hewlett-Packard Model 8554B
Test Performed by $\qquad$
Spectrum Analyzer RF Section
Serial No. $\qquad$ Date

| Para. <br> No. | Test Description | Measurement Units | Min. | Actual | Max. |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 4-15. | Input Impedance Return Loss: 10 dB INPUT ATTENUATION 20-50 dB INPUT ATTENUATION | $\begin{aligned} & \mathrm{dB} \\ & \mathrm{~dB} \end{aligned}$ | $\begin{aligned} & 10 \\ & 10 \end{aligned}$ |  |  |
| 4-16. | Average Noise Level <br> At 10 kHz BANDWIDTH: $\begin{array}{r}-102 \mathrm{dBm} ; 100 \mathrm{kHz} \\ \text { to } 1250 \mathrm{MHz}\end{array}$ | dBm |  | - | -102 |
| 4-17. | Frequency Response <br> Flatness 100 kHz to $1250 \mathrm{MHz}: \pm 1 \mathrm{~dB}$ | Linear Div | 6.3 |  | 8.0 |
| 4-18. | Spurious Responses -40 dBm Input Signal Level: IM products down -65 dB Third Order Second Order | $\begin{aligned} & \mathrm{dB} \\ & \mathrm{~dB} \end{aligned}$ | $\begin{aligned} & -65 \\ & -65 \end{aligned}$ |  |  |
| 4-19. | Residual Responses $100 \mathrm{kHz}-1250 \mathrm{MHz}$; Residual Responses down -100 dBm | dB |  |  | -100 |
| 4-20. | Noise Sidebands <br> Noise Level 50 kHz away from signal: $>70 \mathrm{~dB}$ down | Log Divisions below carrier | 7 | - |  |
| 4-21. | Scan Width Accuracy <br> Frequency error between two points: $\pm 10 \%$ | Divisions at +3 graticule | +2.2 | - | +3.8 |
| 4-22. | $\begin{aligned} & \text { Center Frequency } \\ & \text { Accuracy at } 100 \mathrm{MHz}: \pm 10 \mathrm{MHz} \\ & \text { at } 200 \mathrm{MHz}: \pm 10 \mathrm{MHz} \\ & \text { at } 300 \mathrm{MHz}: \pm 10 \mathrm{MHz} \\ & \text { at } 400 \mathrm{MHz}: \pm 10 \mathrm{MHz} \\ & \text { at } 500 \mathrm{MHz}: \pm 10 \mathrm{MHz} \\ & \text { at } 600 \mathrm{MHz}: \pm 10 \mathrm{MHz} \\ & \text { at } 700 \mathrm{MHz}: \pm 10 \mathrm{MHz} \\ & \text { at } 800 \mathrm{MHz}: \pm 10 \mathrm{MHz} \\ & \text { at } 900 \mathrm{MHz}: \pm 10 \mathrm{MHz} \\ & \text { at } 1000 \mathrm{MHz}: \pm 10 \mathrm{MHz} \\ & \text { at } 1100 \mathrm{MHz}: \pm 10 \mathrm{MHz} \\ & \text { at } 1200 \mathrm{MHz}: \pm 10 \mathrm{MHz} \end{aligned}$ | Divisions <br> Divisions <br> Divisions <br> Divisions <br> Divisions <br> Divisions <br> Divisions <br> Divisions <br> Divisions <br> Divisions <br> Divisions <br> Divisions | $\begin{aligned} & -1 \\ & -1 \\ & -1 \\ & -1 \\ & -1 \\ & -1 \\ & -1 \\ & -1 \\ & -1 \\ & -1 \\ & -1 \\ & -1 \end{aligned}$ |  | $\begin{aligned} & +1 \\ & +1 \\ & +1 \\ & +1 \\ & +1 \\ & +1 \\ & +1 \\ & +1 \\ & +1 \\ & +1 \\ & +1 \\ & +1 \end{aligned}$ |

Table 4-2. Performance Test Record (cont'd)

| Para <br> No. | Test Description | Measurement <br> Units | Min. | Actual | Max. |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 4-23. | Local Oscillator Stability and Residual Fre- <br> quency Modulation <br> Stabilized: 100 Hz peak-to-peak <br> Unstabilized: 10 kHz peak-to-peak | $\mathrm{Hz} \mathrm{pk-pk}$ <br> $\mathrm{kHz} \mathrm{pk-pk}$ |  |  |  |

# SECTION V <br> ADJUSTMENTS 

## 5-1. INTRODUCTION

5-2. This section describes adjustments required to return the analyzer RF Section to peak operating condition when repairs are required. Included in this section are test setups, and check and adjustment procedures. Adjustment location photographs are contained in foldouts in Section VIII.
$5-3$. Record data, taken during adjustments, in the spaces provided. Comparison of initial data with data taken during periodic adjustments assists in preventive maintenance and troubleshooting.

## 5-4. EQUIPMENT REOUIRED

$5-5$. Tables $1-4$ and $1-5$ contain a tabular list of test equipment and test accessories required in the adjustment procedures. In addition, the tables contain the required minimum specifications and a suggested manufacturers model number.

5-6. In addition to the test equipment and test accessories in Tables 1-4 and 1-5, a Display Section and an IF Section are required. Perform the Display Section and IF Section adjustments prior to performing the RF Section adjustments.

## 5-7. Posidriv Screwdrivers

$5-8$. Many screws in the instrument appear to be Phillips, but are not. Table 1-5 gives the name and number of the Posidriv screwdrivers designed to fit these screws. To avoid damage to the screw slots, Posidriv screwdrivers should be used.

## 5-9. Blade Tuning Tools

$5-10$. For adjustments requiring a non-metallic metal-blade tuning tool, use the J.F.D. Model No. 5284 (HP 8710-1010). In situations not requiring non-metallic tuning tools, an ordinary small screwdriver or other suitable tool is sufficient. No matter what tool is used, never try to force any adjustment control in the analyzer. This is especially critical when tuning variable slug-tuned inductors, and variable capacitors.

## 5-11. HP 11592A Service Kit

$5-12$. The HP 11592A Service Kit is an accessory item available from Hewlett-Packard for use in maintaining both the $R F$ and IF Sections of the Spectrum Analyzer. Some adjustments can be made without this kit by removing the top covers from both the RF Section and the Display Section. However, this procedure exposes dangerous potentials in the Display Section chassis and should not be used unless absolutely necessary. All adjustments can and should be performed with the analyzer plug-ins installed on the extender cables provided in the service kit, or using a Display Section cover that has the area over the RF Section cut out. The kit can be obtained by contacting the nearest Hewlett-Packard Sales and Service Office. An extra cover can be obtained from HewlettPackard and modified.

5-13. Table 1-5, Test Equipment Accessories, contains a detailed description of the contents of the service kit, and any item in the kit may be ordered separately. In the case of the 11592-60015 Extender Cable Assembly, the wiring is especially critical and fabrication should not be attempted. However, other items in the kit may be built if desired.

## 5-14. Extender Cable Installations

5-15. Push the front panel latch in the direction indicated by the arrow until the latch disengages and pops out from the panel. Pull the plug-ins out of the instrument. Remove the top cover of the RF Section.
$5-16$. Place the plate end of the HP 11592-60015 Extender Cable Assembly in the Display Section and press firmly into place so that the plugs make contact. The plate and plugs cannot be installed upside down as the plate has two holes corresponding to the two guide rods in the mainframe.
$5 \cdot 17$. Connect the upper cable plug to the $R F$ Section and the lower cable plug to the IF Section. The plugs are keyed so that they will go on correctly and will not make contact upside down.

## 5-18. FACTORY SELECTED COMPONENTS

$5-19$. Table $5-2$ contains a list of factory selected components by reference designation, basis of
selection, and schematic diagram location. Factory selected components are designated by an asterisk $\left(^{*}\right)$ on the schematic diagrams in Section VIII.

## 5-20. RELATED ADJUSTMENTS

$5-21$. These adjustments should be performed when the troubleshooting information in Section VIII indicates that an adjustable circuit is not operating correctly. Perform the adjustments after repairing or replacing the circuit. The troubleshooting procedures specify the required adjustments.

5-22. Many adjustments are interrelated. The adjustments are listed below by related set, and if one adjustment in a set is made, the other adjustments in that set should be made:

Third Converter Circuits

1. 500 MHz Local Oscillator (Third LO) Paragraph 5-25.
2. 550 MHz Amplifier and Bandpass Filter Paragraph 5-26.

First and Second Converter Circuits

1. $\quad 1500 \mathrm{MHz}$ Local Oscillator (Second LO) Paragraph 5-27.
2. YIG Oscillator (First LO) - Paragraph 5-28.
3. 2050 MHz Bandpass Filter - Paragraph 5-29.
4. 1500 MHz Notch Filter and Low Pass Filter - Paragraph 5-30.

O-1250 MHz Scan and Marker Shift - Paragraph 5-31.

## Phase Lock Circuits

1. Sampler Balance and Sampler Bias Paragraph 5-32.
2. Full Scan Sampler Output - Paragraph 5-33.
3. Search Oscillator - Paragraph 5-34.
4. Lock Range - Paragraph 5-35.

50 MHz Amplifier Gain - Paragraph 5-36.

Analogic Checks - Paragraph 5-37.

## 5-23. CHECK AND ADJUSTMENT PROCEDURES

$5-24$. Table $5-1$ is a cross reference from adjustable component to adjustment procedure. Table 5-2 lists the factory selected components. Paragraphs 5-25 through 5-37 give the adjustment procedures and checks for the 8554B.

Table 5-1. Adjustable Components

| Adjustable Component | Adjustment Paragraph | Service Sheet | Description |
| :---: | :---: | :---: | :---: |
| A4R10 SEARCH LOOP GAIN ADJ | 5-34 | 7 | Adjusts response time of search loop. |
| A4R15 LOCK <br> RANGE ADJ | 5-35 | 7 | Adjusts phase lock circuits for minimum signal shift when lock is initiated. |
| A5R9 <br> MARKER SHIFT | 5-31 | 9 | Calibrates inverted marker in $0-1250 \mathrm{MHz}$ SCAN WIDTH mode. |
| A4R22 FULL SCAN CENTER ADJ | 5-31 | 9 | Calibrates center frequency in $0-1250 \mathrm{MHz}$ SCAN WIDTH mode. |
| $\begin{aligned} & \text { A6R7 } \\ & 2 \text { GHz ADJ } \end{aligned}$ | 5-28 | 4 | Calibrates CENTER FREQUENCY dial at low end. |
| $\begin{aligned} & \mathrm{A} 6 \mathrm{R} 2 \\ & 3 \mathrm{GHz} \text { ADJ } \end{aligned}$ | 5-28 | 4 | Calibrates CENTER FREQUENCY dial at high end. |
| $\begin{aligned} & \text { A6R18 } \\ & \text { SWEEP CAL } \end{aligned}$ | 5-28 | 4 | First LO scan width adjustment. |
| $\begin{aligned} & \text { A8 IF ADJ } \\ & 1,2,3 \end{aligned}$ | 5-29 | 3 | Adjusts gain and flatness of 2050 MHz band pass filter. |
| A8 LO ADJ | 5-27 | 3 | Adjusts second LO for 1500 MHz $\pm 100 \mathrm{kHz}$. |
| A9C7, C8, C9 | 5-26 | 5 | Adjusts bandwidth and flatness of 550 MHz filter. |
| A9A2R11 SCAN LINEARITY ADJ | 5-25 | 5 | Third LO fine scan linearity adjustment. |
| $\begin{aligned} & \text { A9A2R12 } \\ & \text { SCAN WIDTH } \\ & \text { ADJ } \end{aligned}$ | 5-25 | 5 | Third LO scan width adjustment. |
| A9A3C1 50 MHz <br> IF FILTER ADJ | 5-26 | 5 | Adjusts impedance of 550 MHz filter |
| $\begin{aligned} & \text { A9A4C4 } \\ & 500 \mathrm{MHz} \\ & \text { FREQ ADJ } \end{aligned}$ | 5-25 | 5 | Adjusts third LO center frequency for $500 \mathrm{MHz} \pm 150 \mathrm{kHz}$. |
| $\begin{aligned} & \text { A10A1R14 } \\ & \text { SAMPLER BIAS } \end{aligned}$ | 5-32 | 6 | Sets sampler for maximum efficiency. |
| A10A2R4 SAMPLER BAL ADJ | 5-32 | 6 | Sets sampler output for 0 Vdc with no RF input to sampler. |
| $\begin{aligned} & \text { A.12A1R5 } \\ & \text { GAIN ADJ } \end{aligned}$ | 5-36 | 5 | Adjusts gain of 50 MHz amplifier to achieve -2 to -4 dB overall conversion loss of 8554B. |

Table 5-2. Factory Selected Components

| Component | Service <br> Sheet | Range of <br> Values | Basis of Selection |
| :--- | :---: | :---: | :--- |
| A4R3 | 7 | 9 | 1 k to <br> $2.5 \mathrm{k} \Omega$ |
| A5R4 | See Paragraph 5-33. |  |  |
| Adjusts frequency at which gain compensation <br> starts (point at which A5CR5 start to conduct). <br> Value selected to give best frequency response <br> flatness at high end of band. |  |  |  |
| A5R8 | 9 | 3 k to $11 \mathrm{k} \Omega$ | Value selected to give best frequency response <br> flatness at high end of band. |
| A6R34 | 4 | 27 k to <br> $35 \mathrm{k} \Omega$. | Coarse sweep width adjust. <br> Centers range of A6R18 SWEEP <br> CAL ADJ. |
| A6R38 | 4 | 27 k to | Value selected to equal the <br> value of A6R34. |
| A8A3C1 | 3 | 0.2 to <br> 0.7 pF | Selected for maximum power out <br> of oscillator. |
| A9A2R17 | 5 | 5 | See Paragraph 5-25. |
| A12A1R6 | 5 | 100 to <br> $200 \Omega$. | Coarse gain adjust. <br> Centers the range of A12A1R5 <br> 50 MHz GAIN ADJ. |
| A9A4C3 | 5 | 2 to <br> 4 pF | See Paragraph 5-25. |

## ADJUSTMENTS

## 5-25. 500 MHz Local Oscillator (Third LO)

REFERENCE: Service Sheet 5.
DESCRIPTION: Third LO is adjusted for a center frequency of 500 MHz . Then a comb signal is centered on the CRT and used as a 500 MHz reference. The SCAN WIDTH ADJ is temporarily mis-adjusted to sweep the $\mathrm{LO} 4 \mathrm{MHz}(500 \mathrm{MHz} \pm 2 \mathrm{MHz}$; this puts five 1 MHz comb signals on the CRT. A9A4C3 and A9A2R17 are selected, and A9A2R11 is adjusted for even comb spacing, then SCAN WIDTH ADJ is adjusted for a correct LO sweep.
EQUIPMENT:
Frequency Counter . . . . . . . . . . . . . . . . . . . . . . . . . . . . HP 5245L/5254C
Comb Generator HP 8406A

Modified Display Section Cover
Non-metallic Tuning Tool

## WARNING

If the following procedure is attempted without the modified Display Section cover, dangerous potentials (up to 7000 Vdc ) will be exposed. Exercise extreme caution.

## PROCEDURE:

1. Install a cover over the Display Section with a cutout above the RF Section.

## ADJUSTMENTS

## 5-25. 500 MHz Local Oscillator (Third LO) (cont'd)



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Figure 5-1. 500 MHz Local Oscillator Adjustment Test Setup
2. Connect the frequency counter to the THIRD LO OUTPUT as shown in Figure 5-1.
3. Set the analyzer as follows:
FREQUENCY ..... 650 MHz
FINE TUNE Centered ( $11 / 2$ turns from stop)
BANDWIDTH ..... 30 kHz
SCAN WIDTH PER DIVISION
SCAN WIDTH PER DIVISION ..... 2 MHz
INPUT ATTENUATION ..... 0 dB
TUNING STABILIZER ..... OFF
SCAN TIME PER DIVISION ..... 10 MILLISECONDS
LOG/LINEAR ..... LOG
LOG REF LEVEL ..... $-40 \mathrm{dBm}$
VIDEO FILTER ..... OFF
SCAN MODE ..... INT
SCAN TRIGGER ..... AUTO
BASE LINE CLIPPER 12 o'clock
4. Adjust the frequency counter to measure 500 MHz . Adjust A9A4C4 ( 500 MHz FREQ ADJ) for 500 $\mathrm{MHz} \pm 150 \mathrm{kHz}$.
$\qquad$ 500.150 MHz
5. Connect the comb generator to RF INPUT (see Figure 5-6); keep the counter connected to the third LO. Adjust the comb generator for 10 MHz comb signals, visible on the analyzer CRT.

## 5-25. 500 MHz Local Oscillator (Third LO) (cont'd)

6. Center one of the comb signals on the CRT with the FREQUENCY control (leave FINE TUNE centered).
7. Set SCAN WIDTH PER DIVISION to 200 kHz , keeping the signal centered with FREQUENCY; do not use FINE TUNE. (This signal is at the center of the third LO frequency scan -500 MHz ).
8. Add 1 MHz comb signals using the INTERPOLATION AMPLITUDE -1 MHz control; leave them low in amplitude so that the 10 MHz comb signal can be easily discerned.
9. Adjust A9A2R12 (SCAN WIDTH ADJ) clockwise until five comb signals (one 10 MHz comb and four 1 MHz combs) are visible on the CRT. Keep the 10 MHz comb signal centered with FINE TUNE; leave FREQUENCY set as it was in step 7. The display should resemble Figure 5-2.
10. There should be a comb signal every 2.5 major divisions $\pm 0.25$ major divisions. If not within tolerance, adjust A9A2R11 (SCAN LINEARITY ADJ) and recenter 10 MHz comb with A9A4C4 (FREQUENCY ADJ). If linearity cannot be correctly adjusted, select values of A9A2R17 and A9A4C3 until it can (repeat steps 1 through 9 after each selection).

Comb Spacing: 2.25 $\qquad$ 2.75 DIV
11. Adjust A9A2R12 (SCAN WIDTH ADJ) until three comb signals are visible on the CRT (SCAN WIDTH PER DIVISION should be 200 kHz ). Keep the 10 MHz comb centered with FINE TUNE. There should be 1 MHz comb signals centered on the extreme left and right graticule lines and the 10 MHz comb should be centered.
12. Center FINE TUNE; frequency counter should read $500 \mathrm{MHz} \pm 150 \mathrm{kHz}$ in ZERO scan. If not, carefully adjust A9A2R11 (SCAN LINEARITY ADJ) until it does.


Figure 5-2. 500 MHz LO Linearity Display

## ADJUSTMENTS

## 5-26. 550 MHz Amplifier and Bandpass Filter

REFERENCE: Service Sheet 5.
DESCRIPTION: A variable 550 MHz signal is connected to the 550 MHz amplifier input and the 50 MHz converter output is observed for gain and bandpass shape. Prior to adjusting 550 MHz bandpass filter, perform 500 MHz LO Check, Paragraph 5-25.


Figure 5-3. 550 MHz Amplifier and Bandpass Filter Adjustment Test Setup

## EQUIPMENT:



## PROCEDURE:

1. Connect the equipment as shown in Figure 5-3 and make the following control settings:
5-26. 550 MHz Amplifier and Bandpass Filter (cont'd)
Spectrum Analyzer (under test):
FINE TUNE Centered
TUNING STABILIZER ..... OFF
SCAN WIDTH ..... ZERO
SCAN MODE ..... INT
SCAN TRIGGER ..... AUTO
Frequency Counter:
SAMPLE RATE 8 o'clock
SENSITIVITY ..... PLUG-IN
LEVEL PRESET
TIME BASE ..... 10 ms
FUNCTION FREQUENCY
8553B/8552A/141T Analyzer:
FREQUENCY ..... 50 MHz
SCAN WIDTH PER DIVISION ..... 1 MHz
BANDWIDTH ..... 100 kHz
INPUT ATTENUATION ..... 10 dB
BASELINE CLIPPER ..... 2 o'clock
SCAN TIME PER DIVISION ..... 2 MILLISECONDS
LOG/LINEARLINEAR
LINEAR SENSITIVITY ..... $2 \mathrm{mV} / \mathrm{DIV}$
VIDEO FILTER ..... OFF
SCAN MODE ..... INT
SCAN TRIGGER ..... AUTO
Signal Generator:
FREQUENCY ..... 550 MHz
OUTPUT ..... CW
OUTPUT LEVEL ..... $-30 \mathrm{dBm}$
2. Adjust 8554 B FINE TUNE control for a frequency indication of 500 MHz on the frequency counter.
3. Adjust $8553 \mathrm{~B} / 8552 \mathrm{~A} / 141 \mathrm{~T}$ Spectrum Analyzer LINEAR SENSITIVITY controls for a seven division vertical deflection.
4. Vary the signal generator $\pm 5 \mathrm{MHz}$ around a center frequency of 550 MHz .
5. Check 3 dB bandwidth (see Figure 5-4).

$$
8
$$

$\qquad$ 11 MHz
6. Select 0.5 MHz SCAN WIDTH PER DIVISION on 8553B RF Section.
7. Repeat step 4 and abserve CRT display for flatness over a 3 MHz range (see Figure 5-4).
$\qquad$ $\pm 0.2 \mathrm{~dB}$
8. If steps 5 and 7 are not within limits, repeat step 4 and adjust A9C7, A9C8 and A9C9 for correct bandpass. Adjust A9A3C1 for maximum signal level.
9. Repeat steps 4 through 8 as required.

## ADJUSTMENTS

## $5-26.550 \mathrm{MHz}$ Amplifier and Bandpass Filter (cont'd)



Figure 5-4. Bandpass Flatness Display

## 5-27. 1500 MHz Local Oscillator (Second LO)

REFERENCE: Service Sheet 3.
DESCRIPTION: The second ( 1500 MHz ) local oscillator is checked for a center frequency of 1500 MHz $\pm 100 \mathrm{kHz}$. The notch filter, connected to the 500 MHz output of the A8 First and Second Converter Assembly, is removed and the second LO feedthru signal is checked using a frequency colinter. Allow at least two hours for instrument to warm up and stabilize.


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Figure 5-5. 1500 MHz Local Oscillator Adjustment Test Setup

## ADJUSTMENTS

## 5-27. 1500 MHz Local Oscillator (Second LO) (cont'd)

EQUIPMENT:

| Frequency Counter | HP 5245L/5254C |
| :---: | :---: |
| Test Cable | HP 11592-60001 |
| Modified Display Section Cover |  |
| No. 10 Allen Driver |  |
| 5/16 inch open-end wrench |  |

## WARNING

If the following procedure is attempted without the modified Display Section cover, dangerous potentials (up to 7000 Vdc ) will be exposed. Exercise extreme caution.

## PROCEDURE:

1. Remove plug-ins form Display Section and remove top cover from 8554B RF Section. Install plug-ins in Display Section.
2. Install a cover on the Display Section with a cutout above the analyzer plug-ins.
3. Apply power to the analyzer and allow at least two hours for stabilization.
4. With test setup shown in Figure 5-5, make the following control settings:

## Frequency Counter:

SAMPLE RATE . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . 9 o'clock
SENSITIVITY . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . PLUG-IN
TIME BASE . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . 10 ms
FUNCTION . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . FREQUENCY
5. Tune the 5254 C frequency control for a maximum meter indication around 1.45 GHz .
6. Measure the second LO feedthru signal. If necessary adjust LO ADJ tune slug (both if unit has two) for a frequency of $1500 \mathrm{MHz} \pm 100 \mathrm{kHz}$.

## ADJUSTMENTS

## 5-28. YIG-Tuned Oscillator (First LO)

REFERENCE: Service Sheet 4.
DESCRIPTION: The YIG-tuned Oscillator (first LO) is checked and adjusted, if necessary, at the 0 and 1 GHz dial points. Tuning linearity over the full tuning range is checked in 100 MHz increments. The 1500 MHz Local Oscillator Check and Adjustment, Paragraph 5-27, should be performed before adjusting the YIG Oscillator.


Figure 5-6. YIG-Tuned Oscillator Adjustment Test Setup

## EQUIPMENT:

Frequency Comb Generator
HP 8406A
Modified Display Section Cover

## WARNING

If the following procedure is attempted without the modified Display Section Cover, dangerous potentials (up to 7000 Vdc ) will be exposed. Exercise extreme caution.

## PROCEDURE:

1. Install a cover over the Display Section with a cutout above the analyzer plug-ins.
2. Remove plug-ins from Display Section and remove top cover from 8554 B RF Section. Reinstall plug-ins in Display Section.
3. Apply power to analyzer and allow at least two hours for stabilization.

## ADJUSTMENTS

## 5-28. YIG-Tuned Oscillator (First LO) (cont'd)

## 4. Connect the test setup in Figure 5-6 and make the following control settings:

## Spectrum Analyzer:

FREQUENCY ..... 0 MHz
FINE TUNE Centered ( $11 / 2$ turns from stop)
TUNING STABILIZER ..... OFF
BANDWIDTH ..... 100 kHz
SCAN WIDTH ..... PER DIVISION
SCAN WIDTH PER DIVISION ..... 20 MHz
INPUT ATTENUATION ..... 0 dB
BASE LINE CLIPPER ..... 9 o'clock
LOG REF LEVEL ..... $-10 \mathrm{dBm}$
LOG/LINEAR ..... LOG
SCAN TIME PER DIVISION ..... 20 MILLISECONDS
VIDEO FILTER ..... OFF
SCAN MODE ..... INT
SCAN TRIGGER ..... LINE
Comb Generator:
COMB FREQUENCY - MC ..... 100 MC
OUTPUT AMPLITUDE ..... 3 o'clock
5. With FINE TUNE centered and FREQUENCY set to " 0 " on frequency dial, adjust A6R7 (2 GHz ADJ) to locate feedthru signal at CENTER FREQUENCY graticule line.
6. Rotate FREQUENCY for indicated dial frequency, of 1000 MHz . Center FINE TUNE and adjust A6R2 ( 3 GHz ADJ) to center the 1 GHz comb signal at the CENTER FREQUENCY graticule line.
7. Set SCAN WIDTH PER DIVISION to 2 MHz and repeat steps 5 and 6 until signal is within one division of CENTER FREQUENCY graticule at a frequency dial indication of 0 and 1000 MHz .
8. Rotate FREQUENCY from 0 to 1200 MHz in 100 MHz steps. Check signal, in relation to CENTER FREQUENCY, at each step. With dial pointer aligned with 100 MHz dial markers and FINE TUNE control centered; displayed signal should be within four divisions of CENTER FREQUENCY graticule.
$\qquad$ divisions
9. Set BANDWIDTH to 300 kHz , and set SCAN WIDTH PER DIVISION to 100 MHz . Adjust frequency controls on RF Section to center LO feedthru signal on left hand graticule line.
10. Observe 1000 MHz comb signal at the right hand graticule line.
11. If necessary adjust SWEEP CAL A6R18 to position marker signals on the vertical graticule lines. Some adjustment of FINE TUNE control may be necessary to position the comb markers when adjusting A6R18.
12. Set SCAN WIDTH PER DIVISION to 50 MHz . Adjust frequency controls to align left hand comb signal with -4 graticule line. Record spacing of right hand comb signal at or near the +4 graticule line.

3. If spacing of comb signals is not within limits, perform IF Section Horizontal Scan Checks and Adjustments in IF Section manual.

## ADJUSTMENTS

## 5-29. 2050 NHz Bandpass Filter

## REFERENCE: Service Sheet 3.

DESCRIPTION: Allow at least two hours for instrument to warm up and stabilize before adjusting 2050 MHz bandpass filters. With the second LO set to 1500 MHz and third LO at 500 MHz , the first LO is tuned to a center frequency of 2050 MHz . The first LO is swept over a 10 MHz range and the resultant feedthru signal at the output of the 50 MHz amplifier is displayed on the CRT of a separate analyzer. The three tunable cavities in the first and second converter are adjusted for amplitude and flatness over a 2 MHz bandpass.


Figure 5-7. 2050 MHz Bandpass Filter Adjustment Test Setup
EQUIPMENT:

| Frequency Counter | HP 5245L/5254C |
| :---: | :---: |
| Comb Generator | HP 8406A |
| Spectrum Analyzer | HP 8553B/8552A/141T |
| Test Cable (2) | HP 11592-60001 |
| Modified Display Section Cover |  |
| No. 10 Allen Driver |  |
| 5/16 inch Open End Wrench |  |

## WARNING

If the following procedure is attempted without the modified Display Section Cover, dangerous potentials (up to 7000 Vdc ) will be exposed. Exercise extreme caution.

## PROCEDURE:

1. Install a cover over the display section with a cutout above the analyzer plug-ins.

## NOTE

Allow at least two hours warmup time for instrument to stabilize prior to making frequency adjustments.

## ADJUSTMENTS

## 5-29. 2050 MHz Bandpass Filter (cont'd)

2. Perform 500 MHz LO Frequency Check and Adjustment, Paragraph 5-25. Leave FINE TUNE control centered.
3. Perform 1500 MHz LO Frequency Check and Adjustment, Paragraph 5-27.
4. Perform YIG Oscillator Frequency Check and Adjustment, Paragraph 5-28. Tune FREQUENCY control for YIG oscillator frequency of 2050 MHz (read on counter).
5. With the test setup connected as shown in Figure 5-7, make the following control settings:

## Spectrum Analyzer (under test):

FREQUENCY . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . See step 4

FINE TUNE . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . See step 2
TUNING STABILIZER . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . OFF
INPUT ATTENUATION . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . 10 dB
SCAN WIDTH PER DIVISION . . . . . . . . . . . . . . . . . . . . . . . . . . . 2 MHz
SCAN TIME PER DIVISION . . . . . . . . . . . . . . . . . . . . . . 2 MILLISECONDS
SCAN MODE . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . INT
SCAN TRIGGER . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . LINE
Comb Generator:
COMB FREQUENCY - MC . . . . . . . . . . . . . . . . . . . . . . . . . . . . . 10 MC
OUTPUT AMPLITUDE . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . 3 o'clock
Spectrum Analyzer:
FREQUENCY . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . 50 MHz
FINE TUNE Centered
BANDWIDTH 300 kHz
SCAN WIDTH PER DIVISION 1 MHz
INPUT ATTENUATION . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . 10 dB
BASE LINE CLIPPER . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . 9 o’clock
SCAN TIME PER DIVISION . . . . . . . . . . . . . . . . . . . . . . . . . 5 SECONDS
SCAN MODE . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . INT
SCAN TRIGGER . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . LINE
LOG/LINEAR LINEAR
LINEAR SENSITIVITY
$2 \mathrm{mV} / \mathrm{DIV}$
6. Center 50 MHz comb signal on CRT display of $8553 \mathrm{~B} / 8552 \mathrm{~A} / 141 \mathrm{~T}$ analyzer. Disconnect comb generator from analyzer.
7. Connect 50 MHz output signal from the 8554 B to RF INPUT on the 8553 B and adjust LINEAR SENSITIVITY controls for a 7.2 division vertical deflection on the CRT.
8. Adjust the cavity tuning slugs (IF ADJ 1, IF ADJ 2 and IF ADJ 3) for maximum gain and flatness. (See Figure 5-8a).
9. Change $8553 \mathrm{~B} / 8552 \mathrm{~A} / 141 \mathrm{~T}$ analyzer SCAN WIDTH PER DIVISION to 0.5 MHz . Repeat cavity tuning adjustments (see Figure 5-8b).
10. Change $8553 \mathrm{~B} / 8552 \mathrm{~A} / 141 \mathrm{~T}$ analyzer SCAN WIDTH PER DIVISION to 0.2 MHz . Repeat cavity tuning adjustments (see Figure 5-8c).
$\qquad$ $\pm 0.4 \mathrm{div}$

## 5-29. 2050 MHz Bandpass Filter (cont'd)




C

Figure 5-8. CRT Display, $1 \mathrm{MHz}, 0.5 \mathrm{MHz}$ and 0.2 MHz Per Division

## ADJUSTMENTS

## $5-30.1500 \mathrm{MHz}$ Notch Filter and Low Pass Filter

REFERENCE: Service Sheets 2 and 3.
DESCRIPTION: The notch filter is checked by removing the filter from its normal location, connecting it to the RF INPUT and comparing the resultant CRT display of a comb signal both through the filter and without the filter. The low pass filter is checked by comparing the CRT display with the filter in its normal position against the display with the filter replaced by a feedthru connector. An optional method would be to use a higher frequency analyzer such as the $8555 \mathrm{~A} / 8552 / 140$ and a swept signal source.


Figure 5-9. 1500 MHz Notch Filter and Low Pass Filter Test Setup

## EQUIPMENT:



## WARNING

If the following procedure is attempted without the modified Display Section Cover, dangerous potentials (up to 7000 Vdc ) will be exposed. Exercise extreme caution.

PROCEDURE:

1. Install a cover over the display section with a cutout above the analyzer plug-ins.

## ADJUSTMENTS

## 5-30. 1500 MHz Notch Filter and Low Pass Filter (cont'd)

2. Remove 1500 MHz Notch Filter FL2 from analyzer and connect HP 11592-60003 cable between A8.J1 and W5 cable to third converter.
3. Connect equipment as shown in Figure 5-9 and make the following control settings:
Spectrum Analyzer:FREQUENCY1200 MHzFINE TUNECentered
BANDWIDTH ..... 300 kHz
SCAN WIDTH PER DIVISION ..... 100 MHz
INPUT ATTENUATION ..... 10 dB
BASE LINE CLIPPER ..... 10 o'clock
SCAN TIME PER DIVISION ..... i. 0.1 SECONDSLOG/LINEARLOG
LOG REF LEVEL ..... $-10 \mathrm{dBm}$
VIDEO FILTER ..... 10 kHz
SCAN MODE ..... INT
SCAN TRIGGER ..... LINE



C


Figure 5-10. CRT Display Without and With Notch Filter and Low Pass Filter

## ADJUSTMENTS

## 5-30. 1500 MHz Notch Filter and Low Pass Filter (cont'd)

## Comb Generator:

$$
\begin{aligned}
& \text { COMB FREQUENCY - MC . . . . . . . . . . . . . . . . . . . . . . . . } 100 \mathrm{MC} \\
& \text { OUTPUT AMPLITUDE }
\end{aligned}
$$

4. Observe CRT for a display similar to Figure 5-10a. With a grease pencil mark level of comb signals on CRT.
5. Install Notch Filter FL2 between HP 1250-0827 and HP 11592-60001 cable to comb generator.
6. Observe CRT display level of comb signals (see Figure $5-10 \mathrm{~b}$ ). The 1500 MHz comb signal (three graticule line) should be at least 20 dB below signal level observed in step 4 above.

20 dB
7. Remove HP 11592-60003 cable and reinstall notch filter FL2.
8. Remove 1500 MHz Low Pass Filter FL1 and replace with OSM Jack-to-Jack Adapter HP 1250-1158.
9. Observe CRT for a display similar to Figure $5-10 \mathrm{c}$. With a grease pencil mark level of comb signals, harmonic mixing products and image responses.
10. Install low pass filter and observe CRT for a display similar to Figure 5-10d. Harmonic mixing products and image responses should be below -70 dBm .
$\qquad$

## ADJUSTMENTS

## 5-31. 0-1250 MHz Scan and Marker Shift

REFERENCE: Service Sheet 9 .
DESCRIPTION: The $0-1250 \mathrm{MHz}$ scan alignment adjusts the center frequency around which the analyzer tunes in the full scan mode. Perform the YIG Oscillator adjustment, paragraph 5-28, prior to adjusting the 0 -1250 MHz center frequency.


Figure 5-11. 0-1250 MHz Scan and Marker Shift Adjustment Test Setup

## EQUIPMENT:

Comb Generator . . . . . . . . . . . . . . . . . . . . . . . . . . HP 8406A
Modified Display Section Cover

## WARNING

If the following procedure is attempted without the modified Display Section cover, dangerous potentials (up to 7000 Vdc ) will be exposed. Exercise extreme caution.

## PROCEDURE:

1. Install a cover over the display section with a cutout above the anlayzer plug-ins.
2. Remove plug-ins from display section and remove top cover from RF Section. Install plug-ins in display section.
3. Apply power to analyzer and allow at least two hours for stabilization. With the test setup as shown in Figure 5-11, make the following control settings:

## Spectrum Analyzer:

FREQUENCY
600 MHz

## ADJUSTMENTS

$5-31 . \quad 0-1250 \mathrm{MHz}$ Scan and Marker Shift (cont'd)
FINE TUNE Centered
TUNING STABILIZER ..... OFF
BANDWIDTH ..... 30 kHz
SCAN WIDTH ..... $0-1250 \mathrm{MHz}$
SCAN WIDTH PER DIVISION ..... 1 MHz
INPUT ATTENUATION ..... 0 dB
BASE LINE CLIPPER ..... 8 o'clock
SCAN TIME PER DIVISION 20 MILLISECONDS
LOG/LINEAR ..... LOG
LOG REF LEVEL ..... $-10 \mathrm{dBm}$
VIDEO FILTER ..... OFF
SCAN MODE ..... INT
SCAN TRIGGER ..... LINE
Comb Generator:
COMB FREQUENCY - MC ..... 100 MC
OUTPUT AMPLITLDE ..... 3 o'clock
4. Adjust FULL SCAN CENTER ADJ A5R22 to align LO feedthru signal on the left hand graticule line. Observe the 500 MHz comb marker at the -1 graticule line. The comb marker should be within $\pm 0.2$ division of the -1 graticule line.

$$
-0.2
$$

$\qquad$ +0.2 div
5. Observe the 1000 MHz comb marker at the +3 graticule line. The comb marker should be within $\pm 0.4$ division of the +3 graticule line.
$-0.4$ $\qquad$ +0.4 div
6. With the test setup and controls set as in step 3 above, center the frequency marker under the 500 MHz comb signal. Set SCAN WIDTH to PER DIVISION and record the difference between the CENTER FREQUENCY graticule line and the 600 MHz comb marker. The comb marker should be within $\pm 5$ divisions ( 5 MHz ) of the center graticule.
7. If comb marker is not within $\pm 5$ divisions tune FREQUENCY control to center 600 MHz marker on CRT. Set SCAN WIDTH to $0-1250 \mathrm{MHz}$ and adjust A5R9 MARKER SHIFT to center marker under the 600 MHz comb signal. Repeat step 6 above.

## 5-32. Sampler Balance and Sampler Bias

REFERENCE: Service Sheet 6.
DESCRIPTION: The amplifier output of the sampler is checked for zero balance with the YIG oscillator signal disconnected between Circulatior Assembly A13 and Sampler Assembly. SAMPLER BAL ADJ A10A2R4 is adjusted for a zero output level as displayed on an oscilloscope.


Figure 5-12. Sampler Balance Check and Adjustment Test Setup

## EQUIPMENT:

Oscilloscope
HP 180A/1801A/1821A
10:1 Divider Probe HP 10004
Modified Display Section Cover
Tuning Tool

## WARNING

If the following procedure is attempted without the modified Display Section cover, dangerous potentials (up to 7000 Vdc ) will be exposed. Exercise extreme caution.

## PROCEDURE:

1. Install a cover over the Display Section with a cutout above the analyzer plug-ins.
2. Connect equipment shown in test setup, Figure $5-12$, and make the following control settings:

## Spectrum Analyzer:

SCAN WIDTH . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . 0 - 1250 MHz
Oscilloscope:
VOLTS/DIV
INPUT . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . DC coupled
TIME/DIV 5 msec
3. Adjust A10A2R4 SAMPLER BAL ADJ for minimum output. It should be $0 \pm 25 \mathrm{mV}$. If not, perform steps 4 through 7.

## ADJUSTMENTS

## 5-32. Sampler Balance and Sampler Bias (cont'd)

4. Remove and extend RF and IF Sections from Display Section (use Extender Cable Assembly HP 11592-60015). Separate RF and IF Sections and connect with Interconnection Cable Assembly HP 11592-60016.
5. Remove bottom cover from 8554B; remove cover from A10 assembly. Connect oscilloscope probe to A10A2TPD and set VOLTS/DIV to 0.1.
6. Adjust A10A1R14 SAMPLER BIAS for a maximum, noise-free indication on the oscilloscope.
7. Replace A10 cover and RF Section bottom cover. Re-install RF and IF Sections in Display Section and repeat steps 1 through 3.

## 5-33. Full Scan Sampler Output

REFERENCE: Service Sheets 6 and 7.
DESCRIPTION: The full scan sampler signal is checked for VSWR and peak-to-peak voltage level. This check should not be required unless components associated with the phase lock circuitry are changed. To check the sampler signal, remove assembly A4 and connect oscilloscope to A10C3. Trigger the oscilloscope externally by connecting the analyzer SCAN OUT to EXT INPUT on the oscilloscope. The resulting display contains the sampling signals over the full scan range of the analyzer. A detailed examination of the sampling signals can be made by selecting a narrow scan width and tuning over the frequency range of interest.


Figure 5-13. Full Scan Sampler Output Check and Adjustment

## ADJUSTMENTS

## 5-33. Full Scan Sampler Output (cont'd)

EQUIPMENT:
Oscilloscope
HP 180A/1801A/1821A
10:1 Divider Probe
HP 10004A
Modified Display Section Cover

## WARNING

If the following procedure is attempted without the modified Display Section cover, dangerous potentials (up to 7000 Vdc ) will be exposed. Exercise extreme caution.

## PROCEDURE:

1. Install a cover over the Display Section with a cutout above the analyzer plug-ins.
2. With analyzer power off, remove the A4 assembly from the analyzer RF Section.
3. Connect test setup shown in Figure 5-13 and make the following control settings:

## Spectrum Analyzer:

SCAN WIDTH . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . 0 - 1250 MHz
SCAN TIME PER DIVISION . . . . . . . . . . . . . . . . . . . . . . 5 MILLISECONDS
SCAN MODE . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . INT
SCAN TRIGGER . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . AUTO
Oscilloscope:
DISPLAY . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . EXT CAL
VOLTS/DIV .05
4. Observe oscilloscope CRT display for a presentation similiar to Figure 5-14.
5. Tune analyzer to approximately the frequency of the lowest peak-to-peak sampling signal displayed on the oscilloscope (see sample, Figure 5-14a).
6. Set analyzer SCAN WIDTH PER DIVISION to 1 MHz and tune FREQUENCY control for minimum peak-to-peak signal display on oscilloscope. Record peak-to-peak signal level.
7. Tune FREQUENCY control for nearest maximum peak-to-peak signal display on oscilloscope. Record peak-to-peak signal level.
8. Divide voltage obtained in step 6 into voltage obtained in step 7. If results exceed 1.6 check RF cabling, circulator and sampler termination.
9. Figure $5-14 \mathrm{~b}$ illustrates a typical display of the minimum and maximum peak-to-peak voltage levels.
10. If the sampler assembly, cabling or components in the phase lock amplifier are replaced, determine the value of resistor A4R3 by multiplying the voltage level obtained in step 6 by $17 \mathrm{~K} \Omega$ (i.e., $1.8 \times 17 \mathrm{~K} \Omega=$ $30.6 \mathrm{~K} \Omega$ value for A 4 R 3 ).

## 5-33. Full Scan Sampler Output (cont'd)



Figure 5-14. Typical Display of Sampling Signals

## ADJUSTMENTS

## 5-34. Search Oscillator

## REFERENCE: Service Sheet 7.

DESCRIPTION: The sampling signal input to the phase lock equalizer circuit is disconnected at A10C3 and the input to the equalizer is then grounded. When the TUNING STABILIZER is turned ON, the equalizer circuit functions as a free running oscillator. The TUNING STABILIZER is switched OFF and On and A4R10, SEARCH LOOP GAIN ADJ, is adjusted to ensure the search oscillations will start and the signal level reaches full amplitude within one second. (See Lock Range Adjustment, Paragraph 5-35).


Figure 5-15. Search Oscillator Check and Adjustment Test Setup
EQUIPMENT:
Oscilloscope . . . . . . . . . . . . . . . . . . . . . . HP 180A/1801A/1821A
10:1 Divider Probe . . . . . . . . . . . . . . . . . . . . . . . . . HP 10004A
Ground Strap with clips
Soldering iron
Modified Display Section Cover

## WARNING

If the following procedure is attempted without the modified Display Section cover, dangerous potentials (up to 7000 Vdc ) will be exposed. Exercise extreme caution.

## PROCEDURE:

1. Install a cover over the Display Section with a cutout above the analyzer plug-ins.
2. With analyzer power off, unsolder center conductor of cable W9 from A10C3. Connect a jumper between W9 center conductor and chassis ground. Do not ground A10C3.
3. With the test setup connected as shown in Figure 5-15, make the following control settings:

## 5-34. Search Oscillator (cont'd)

Spectrum Analyzer :
FREQUENCY . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . 0 MHz
BANDWIDTH 30 kHz
SCAN WIDTH PER DIVISION
SCAN WIDTH PER DIVISION 200 kHz TUNING STABILIZER On

Oscilloscope:
VOLTS/DIV . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . 02
TIME/DIV 20 MSEC
Channel A Input DC coupled
4. Observe oscilloscope display for a search signal. The signal should be approximately a 1.2 -volt peak-to-peak sine wave with noticeable distortion.
5. Switch TUNING STABILIZER to OFF. Oscillations should stop.
6. Set oscilloscope scan time (TIME/DIV) to 0.1 SEC. Switch analyzer TUNING STABILIZER to On. Oscillations should reach full amplitude in less than one second. Switch TUNING STABILIZER OFF and On several times while noting time required for signal to reach peak value.
7. If time required for oscillations to reach full value exceeds one second, adjust A4R10 SEARCH LOOP GAIN ADJ to increase signal amplitude.
8. Repeat steps 6 and 7 until oscillations reach peak value in less than one second.

## ADJUSTMENTS

## 5-35. Lock Range

REFERENCE: Service Sheet 7.
DESCRIPTION: See description in Paragraph 5-34. In addition to the conditions established for search oscillator check, the offset memory relay is energized to apply the search signal to the third LO. With the first and third local oscillators being swept in opposite directions, the LO feedthru signal is observed and the LOCK RANGE ADJ A4R15 set for the least amount of frequency variation.


Figure 5-16. Lock Range Check and Adjustment Test Setup

## EQUIPMENT:

Soldering Iron
Ground Strap with clips (2)
Modified Display Section Cover

## WARNING

If the following procedure is attempted without the modified Display Section cover, dangerous potentials (up to 7000 Vdc ) will be exposed. Exercise extreme caution.

## PROCEDURE:

1. Install a cover over the Display Section with a cutout above the analyzer plug-ins.
2. With analyzer power off, unsolder center conductor of cable W9 from A10C3. Connect a jumper between W9 center conductor and chassis ground. Do not ground A10C3. Connect a jumper between collector of A4Q6 (case) and chassis ground.
3. With the test setup connected as shown in Figure 5-16, make the the following control settings:

## Spectrum Analyzer:

FREQUENCY
FINE TUNE . . . . . . . . . . . . . . . . . . . . . . . . . . . . . 0 MHz

## ADJUSTMENTS

$5-35$. Lock Range (cont'd)
BANDWIDTH ..... 10 kHz
TUNING STABILIZER ..... On
SCAN WIDTH PER DIVISION
SCAN WIDTH PER DIVISION ..... 200 kHz
BASE LINE CLIPPER ..... 9 o'clock
SCAN TIME PER DIVISION 2 MILLISECONDS
LOG/LINEAR ..... LOG
LOG REF LEVEL ..... 0 dBm
VIDEO FILTER ..... OFF
SCAN MODE ..... INT
SCAN TRIGGER ..... AUTO
4. Set analyzer POWER switch to ON position.
5. Tune FREQUENCY control to center LO feedthru signal on CRT display.
6. Reduce SCAN WIDTH PER DIVISION to 20 kHz and center LO feedthru signal on CRT display.
7. Adjust A4R15 for least amount of frequency variation. Set SCAN WIDTH PER DIVISION to 10 kHz and repeat A4R15 adjustment. Frequency variation should not exceed 20 kHz .

20 kHz
8. Remove ground jumpers and connect cable removed in step 2.

## 5-36. 50 MHz Amplifier Gain

REFERENCE: Service Sheet 5.
DESCRIPTION: With the IF Section vertically calibrated, a -30 dBm signal is applied to the RF INPUT and the AMPL CAL potentiometer R4 is checked for a range of 7 to 10 dB . The input signal is reduced by 3 dB and the AMPL CAL pot is checked for sufficient range to increase signal level to -30 dBm . If not the 50 MHz gain adjustment A 12 A 1 R 5 is adjusted to set signal at -30 dBm .


Figure 5-17. AMPL CAL and 50 MHz Gain Check and Adjustment Test Setup

## EQUIPMENT:

Modified Display Section Cover

## WARNING

If the following procedure is attempted without the modified Display Section cover, dangerous potentials (up to 7000 Vdc ) will be exposed. Exercise extreme caution.

PROCEDURE:

1. Install a cover over the Display Section with a cutout above the analyzer plug-ins.
2. With the test setup connected as shown in Figuer 5-17, make the following control settings:

## Spectrum Analyzer:

FREQUENCY . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . 30 MHz
FINE TUNE Centered
BANDWIDTH 100 kHz
SCAN WIDTH PER DIVISION
SCAN WIDTH PER DIVISION 2 MHz
ATTENUATION . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . 10 dB
SCAN TIME PER DIVISION . . . . . . . . . . . . . . . . . . . . . . 5 MILLISECONDS
BASE LINE CLIPPER . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . . Max ccw

## ADJUSTMENTS

## 5-36. 50 MHz Amplifier Gain (cont'd)


3. Vary AMPL CAL potentiometer for minimum and maximum vertical deflection of the 30 MHz CAL OUTPUT signal. Signal level should be adjustable over a 7 to 10 dB range.

7 $\qquad$ 10 dB
4. Adjust AMPL CAL potentiometer for maximum vertical deflection of 30 MHz CAL OUTPUT signal. Signal level should be at least -27 dBm .
5. If signal level (in step 4 above) is below -27 dBm , adjust 50 MHz gain adjustment A12A1R5 for -27 dBm .

## 5-37. Analogic Checks

REFERENCE: IF Section Operating and Service Manual.
DESCRIPTION: Perform the display calibration check tabulated below. If adjustment is required refer to IF Section Operating and Service Manual for adjustment procedure. When performing the display calibration check, if the table indicates the DISPLAY UNCAL light to be "off", it is acceptable for light to be "on" if the light subsequently goes "off", when either the SCAN TIME PER DIVISION or SCAN WIDTH PER DIVISION control is switched one position counterclockwise.

Table 5-3. Display Calibration Conditions

| VIDEO <br> FILTER | SCAN TIME PER DIVISION | BANDWIDTH | SCAN WIDTH PER DIVISION | SCAN WIDTH | DISPLAY UNCAL LIGHT |
| :---: | :---: | :---: | :---: | :---: | :---: |
| OFF | 5 MILLISECONDS | 300 kHz | 200 MHz | PER DIVISION | ON |
| OFF | 5 MILLISECONDS | 300 kHz | 100 MHz | PER DIVISION | OFF |
| OFF | 5 MILLISECONDS | 100 kHz | 100 MHz | PER DIVISION | ON |
| OFF | 5 MILLISECONDS | 100 kHz | 20 MHz | PER DIVISION | OFF |
| OFF | 5 MILLISECONDS | 20 kHz | 20 MHz | PER DIVISION | ON |
| OFF | 5 MILLISECONDS | 30 kHz | 2 MHz | PER DIVISION | OFF |
| OFF | 5 MILLISECONDS | 10 kHz | 2 MHz | PER DIVISION | ON |
| OFF | 5 MILLISECONDS | 10 kHz | 0.2 MHz | PER DIVISION | OFF |
| OFF | 5 MILLISECONDS | 3 kHz | 0.2 MHz | PER DIVISION | ON |
| OFF | 5 MILLISECONDS | 3 kHz | 20 kHz | PER DIVISION | OFF |
| OFF | 5 MILLISECONDS | 1 kHz | 20 kHz | PER DIVISION | ON |
| OFF | 5 MILLISECONDS | 1 kHz | 2 kHz | PER DIVISION | OFF |
| OFF | 5 MILLISECONDS | 0.3 kHz | 2 kHz | PER DIVISION | ON |
| OFF | 50 MILLISECONDS | 0.3 kHz | 2 kHz | PER DIVISION | OFF |
| OFF | 50 MILLISECONDS | 0.1 kHz | 2 kHz | PER DIVISION | ON |
| OFF | 0.2 SECOND | 0.1 kHz | 2 kHz | PER DIVISION | OFF |
| 100 Hz | 5 SECONDS | 300 kHz | 200 MHz | PER DIVISION | OFF |
| 100 Hz | 5 SECONDS | 100 kHz | 200 MHz | PER DIVISION | ON |
| 100 Hz | 5 SECONDS | 100 kHz | 50 MHz | PER DIVISION | OFF |
| 100 Hz | 5 SECONDS | 30 kHz | 50 MHz | PER DIVISION | ON |
| 100 Hz | 5 SECONDS | 30 kHz | 20 MHz | PER DIVISION | OFF |
| 100 Hz | 5 SECONDS | 10 kHz | 20 MHz | PER DIVISION | ON |
| 100 Hz | 5 SECONDS | 10 kHz | 5 MHz | PER DIVISION | OFF |
| 100 Hz | 5 SECONDS | 3 kHz | 5 MHz | PER DIVISION | ON |
| 100 Hz | 5 SECONDS | 3 kHz | 1 MHz | PER DIVISION | OFF |
| 100 Hz | 5 SECONDS | 1 kHz | 1 MHz | PER DIVISION | ON |
| 100 Hz | 5 SECONDS | 1 kHz | 0.2 MHz | PER DIVISION | OFF |
| 100 Hz | 5 SECONDS | 0.3 kHz | 0.2 MHz | PER DIVISION | ON |
| 100 Hz | 5 SECONDS | 0.3 kHz | 50 MHz | PER DIVISION | OFF |
| 100 Hz | 5 SECONDS | 0.1 kHz | 50 MHz | PER DIVISION | ON |
| 100 Hz | 5 SECONDS | 0.1 kHz | 10 MHz | PER DIVISION | OFF |
| 100 Hz | 2 SECONDS | - | - | FULL | ON |
| 100 Hz | 5 SECONDS | - | - | FULL | OFF |
| 100 Hz | 5 MILLISECONDS | Any | Any | ZERO | OFF |
| OFF | 5 MILLISECONDS | - | - | FULL | ON |
| OFF | 10 MILLISECONDS | - | - | FULL | OFF |

## SECTION VI REPLACEABLE PARTS

## 6-1. INTRODUCTION

6-2. This section contains information for ordering parts. Table 6-1 is a list of exchange assemblies and Table 6-2 lists abbreviations used in the parts list and throughout the manual. Table 6-3 lists all replaceable parts in reference designator order. Table 6-4 contains names and addresses that correspond to the manufacturer's code numbers.

## 6-3. EXCHANGE ASSEMBLIES

6-4. Table 6-1 lists assemblies within the instrument that may be replaced on an exchange basis, thus affording considerable cost savings. Exchange, factory-repaired and tested assemblies are available only on a trade-in basis, therefore the defective assemblies must be returned for credit. For this reason, assemblies required for spare parts stock must be ordered by the new assembly part number.

## 6-5. ABBREVIATIONS

6-6. Table 6-2 gives a list of abbreviations used in the parts list, schematics, and throughout the manual. In some cases, two forms of the abbreviation are given, one all capital letters, and one partial or no capitals. This occurs because the abbreviations in the parts list are always all capitals. However, in the schematics and other parts of the manual, other abbreviation forms are used with both lower case and upper case letters.

## 6-7. REPLACEABLE PARTS LIST

$6-8$. Table $6-3$ is the list of replaceable parts and is organized as follows:
a. Electrical assemblies and their components in alpha-numerical order by reference designation.
b. Chassis-mounted parts in alpha-numeric order by reference designation.
c. Miscellaneous parts.
d. Illustrated parts breakdown.

The information given for each part consists of the following:
a. The Hewlett-Packard part number.
b. The total quantity (Qty) in the instrument.
c. The description of the part.
d. The typical manufacturer of the part in a five-digit code.
e. Manufacturer code number for the part.

6-9. The total quantity for each part is given only once - at the first appearance of the part number in the list.

## 6-10. ORDERING INSTRUCTIONS

6-11. To order a part listed in the replaceable parts table, quote the Hewlett-Packard part number, indicate quantity required, and address the order to the nearest Hewlett-Packard office.

6-12. To order a part that is not listed in the replaceable parts table, include the instrument model number, instrument serial number, the description and function of the part, and the number of parts required. Address the order to the nearest Hewlett-Packard office.

Table 6-1. Part Numbers for Exchange Assemblies

| Reference <br> Designation | Description | Part Number |  |
| :---: | :---: | :---: | :---: |
|  |  | Exchange Assy | New Assy |
| A8 | First and Second Converter Assy | $08554-60051$ | $08554-60012$ |

Table 6-2. Reference Designations and Abbreviations (1 of 2)

## REFERENCE DESIGNATIONS



| E | . . . . miscellaneous electrical part |
| :---: | :---: |
| F | fuse |
| FL | filter |
| H | hardware |
| HY | circulator |
| J | electrical connector (stationary portion) jack |
| K | . relay |
| L | coil; inductor |
| M | meter |
| MP | miscellaneous |
|  | mechanical part |


| P | electrical connector (movable portion): plug |
| :---: | :---: |
| Q | . . transistor: SCR; triode thyristor |
| R | . resistor |
| RT | thermistor |
| S | . switch |
| T | transformer |
| TB | terminal board |
| TC | thermocouple |
| TP | test point |

## ABBREVIATIONS



| COEF . . . . . . coefficient | EDP . . . . electronic data | INT . . . . . . . . internal |
| :---: | :---: | :---: |
| COM . . . . . . . . common | processing | kg . . . . . . . . kilogram |
| COMP . . . . composition | ELECT . . . . electrolytic | kHz . . . . . . . kilohertz |
| COMPL . . . . . . complete | ENCAP . . . encapsulated | k ${ }^{\text {a }}$. . . . . . . . . kilohm |
| CONN . . . . . . connector | EXT . . . . . . . . external | kV . . . . . . . . . kilovolt |
| CP . . . . . cadmium plate | F . . . . . . . . . . . farad | lb . . . . . . . . p pound |
| CRT . . cathode-ray tube | FET . . . . . . field-effect | LC . . . . . . inductance- |
| CTL .... complementary | F/F . . . . . . . . flip-flop | capacitance <br> LED . . light-emitting diode |
| CW . . . . continuous wave | FH . . . . . . . . flat head | LF . . . . . low frequency |
| cw . . . . . . . clockwise | FIL H . . . . fillister head | LG . . . . . . . . . . . long |
| cm . . . . . . . . centimeter | FM. . frequency modulation | LH . . . . . . . . . left hand |
| D/A . . . digital-to-analog | FP . . . . . . . front panel | LIM . . . . . . . . . . limit |
| dB . . . . . . . . . decibel | FREQ . . . . . frequency | LIN . . . linear taper (used |
| dBm .... decibel referred | FXD . . . . . . . . . . fixed | in parts list) |
| to 1 mW | g . . . . . . . . . . gram | lin . . . . . . . . . . linear |
| dc . . . . . . direct current | GE . . . . . . . germanium | LK WASH . . lock washer |
| deg . . degree (temperature | GHz . . . . . . . . gigahertz | LO . . . low; local oscillator |
| interval or differ- | GL . . . . . . . . . . . . . glass | LOG . . . logrithmic taper |
| ence) | GRD . . . . . . ground (ed) | (used in parts list) |
| . . . . degree (plane | H . . . . . . . . . . . henry | log . . . . . . . logrithm(ic) |
| C angle) | h . . . . . . . . . . . hour | LPF . . . . low pass filter |
| C . . . . . degree Celsius | HET . . . . . . heterodyne | LV . . . . . . . low voltage |
| - (centigrade) | HEX . . . . . . . hexagonal | m . . . . . meter (distance) |
| ${ }_{0} \mathrm{~F}$. . . degree Fahrenheit | HD . . . . . . . . . . . head | mA . . . . . . milliampere |
| K . . . . . . degree Kelvin | HDW . . . . . . . hardware | MAX . . . . . maximum |
| DEPC . . deposited carbon | HF . . . . . . high frequency | $\mathrm{M} \Omega$. . . . . . . . megohm |
| DET . . . . . . . detector | HG . . . . . . . . . mercury | MEG . . . meg (10 ${ }^{6}$ ) (used |
| diam . . . . . . . . diameter | HI . . . . . . . . . . . high | in parts list) |
| DIA . . . diameter (used in | HP . . . . Hewlett-Packard | MET FLM . . . metal film |
| parts list) | HPF . . . . . high pass filter | MET OX . . metallic oxide |
| DIFF AMPL . . differential amplifier | HR . . . . . . . hour (used in parts list) | MF . . . medium frequency microfarad (used in |
| div . . . . . . . . . division | HV . . . . . . . high voltage | parts list) |
| DPDT . . . . double-pole, | Hz . . . . . . . . . . hertz | MFR . . . . . manufacturer |
| double-throw | IC .... integrated circuit | mg . . . . . . . milligram |
| DR . . . . . . . . . . drive | ID . . . . . inside diameter | $\mathrm{MHz} \mathrm{}. \mathrm{}. \mathrm{}. \mathrm{}. \mathrm{}. \mathrm{}. \mathrm{}$. |
| DSB . . . . double sideband | IF . . . . . . intermediate | mH . . . . . . . millihenry |
| DTL . . . . diode transistor | frequency | mho . . . . . . . . . . mho |
| logic | IMPG ..... impregnated | MIN . . . . . . . minimum |
| DVM . . . digital voltmeter | in . . . . . . . . . . . . . inch | min . . . . minute (time) |
| ECL . . . . emitter coupled | INCD ..... incandescent | . . . . . minute (plane |
| logic | INCL . . . . . . include(s) | angle) |
| EMF . . electromotive force | INP . . . . . . . . input | MINAT . . . . . minature |
|  | INS ........ insulation | mm . . . . . . . millimeter |


| U ..... integrated circuit; microcircuit |  |
| :---: | :---: |
| V | electron tube |
| VR | . . voltage regulator; breakdown diode |
| W | cable; transmission path; wire |
| X | socket |
| Y | crystal unit (piezoelectric or quartz) |
| Z | tuned cavity; tuned circuit |

Table 6-2. Reference Designations and Abbreviations (2 of 2)

| MOD . . . . . . modulator |  |
| :---: | :---: |
| MOM | momentary |
| MOS |  |
|  |  |
| ms . . . . . . milhisecond |  |
| MTG . . . . . . mounting |  |
| MTR ... meter (indicating device) |  |
| mV . . . . . . . . millivolt |  |
| mVac . . . . . millivolt, ac |  |
| mVdc . . . . . millivolt, dc |  |
| mVpk . . . . millivolt, peak |  |
| mVp-p . . . millivolt, peak-to-peak |  |
| mVrms . . . . millivolt, rms mW . . . . . . . . milliwatt |  |
|  |  |
| MUX . . . . . multiplex |  |
| MY . . . . . . . . . . mylar |  |
| $\mu \mathrm{A}$. . . . . . microampere |  |
| $\mu \mathrm{F}$. . . . . . microfarad |  |
| $\mu \mathrm{H}$. . . . . . . microhenry |  |
| $\mu_{\text {mho }}$. . . . . . micromho |  |
| $\mu \mathrm{s}$. . . . . . . . microsecond $\mu \mathrm{V}$. . . . . . . . microvolt |  |
|  |  |
| $\mu \mathrm{Vac} . .$. . . . . microvolt, ac |  |
| $\mu \mathrm{Vdc}$. . . . . microvolt, de |  |
| $\mu \mathrm{Vpk}$. . . microvolt, peak |  |
| $\mu \mathrm{Vp-p} \underset{\text { to-peak }}{\ldots}$. microvolt, peak- |  |
| $\mu \mathrm{Vrms} . . .$. microvolt, rms |  |
| $\mu \mathrm{W} . . . . . . . . . ~ m i c r o w a t t$ |  |
| nA . . . . . . . nanoampere |  |
| NC . . . . . no connection |  |
| N/C . . . . normally closed |  |
| NE |  |
| NEG |  |
| nF . . . . . . . nanofarad |  |
| N1 PL . . . . . nickel plate |  |
| N/O . . . . normally open |  |
| NOM . . . . . . . nominal |  |
| NORM . . . . . . normal |  |
| NPN |  |
|  |  |
| NPO | . negative-positive |
|  | zero (zero temperature coefficient) |
| NRFR | not recommended |
|  | for field replacement |
| NSR | . . not separately |
| ns | anosecond |
| nW | . nanowatt |
| OBD | order by descrip- |
|  | tion |




## NOTE

All abbreviations in the parts list will be in upper-case.
MULTIPLIERS

| Abbreviation | Prefix | Multiple |
| :---: | :--- | :---: |
| T | tera | $10^{12}$ |
| G | giga | $10^{9}$ |
| M | mega | $10^{6}$ |
| k | kilo | $10^{3}$ |
| da | deka | 10 |
| d | deci | $10^{-1}$ |
| c | centi | $10^{-2}$ |
| m | milli | $10^{-3}$ |
| $\mu$ | micro | $10^{-6}$ |
| n | nano | $10^{-9}$ |
| p | pico | $10^{-12}$ |
| f | femto | $10^{-15}$ |
| a | atto | $10^{-18}$ |

Table 6-3. Replaceable Parts

| Reference Designation | HP Part Number | Oty | Description | Mfr Code | Mfr Part Number |
| :---: | :---: | :---: | :---: | :---: | :---: |
| ${ }^{4}$ | 28555-60009 |  | shitch assy:banowidth | 28480 | 08555-60009 |
|  |  | 1 |  | 2680 <br> O2723 <br> 28480 <br> 888 |  |
| ${ }_{\text {AlO }}{ }^{2}$ | $1854-0042$ $1854-042$ |  | TRSTR:S1 NPN | ${ }_{28480}^{28880}$ | $1884-0040$ 1854-0040 |
|  | $1853-0018$ $1854-0354$ | $\frac{1}{2}$ |  | 28480 28480 | $1853-0018$ 1854 18054 |
| ${ }_{\text {al }}^{1} 1$ |  | 13 |  | 28480 28489 | -0757-3340 |
| ${ }_{4}^{4182}$ | ( $\begin{aligned} & \text { c757-0401 } \\ & 0757-0462\end{aligned}$ | 3 |  | 28480 <br> 2848 C | C757-0401 0757 0 |
| ${ }_{\text {atr }}{ }_{\text {atr }}$ | - 6 \%988-4334 | 3 |  | 28480 | - 06988.4534 |
|  | c698-4534 $0698-4521$ | 4 |  | 28480 2848 C | cose-4534 $0698-4521$ |
| A1R ARB AR | - 6 -698-4521 |  |  | 28480 <br> 28480 | O698-4521 $\substack{0698-4521}$ |
|  | -6998-4534 |  | R:FXD MET FLM 309 K OHM $1 \% 1 / 8 \mathrm{~W}$ | 28480 | ${ }_{0698-4534}$ |
|  | $0698-4521$ $0757-6420$ |  |  | 28480 28480 | $0688-4522$ 0 $0757-6420$ |
|  | 3100-2.777 | 1 |  | 2888 C <br> 28480 | - $3100-2677$ |
|  |  | $\frac{1}{2}$ | Stiter | 28480 28880 | (8554-600024 |
| A ${ }_{\text {A2P2 }}$ | $1853-0020$ $1853-0020$ | 10 | (tRSTR:S1 PNP | 28480 28480 | $1853-0020$ $1853-0020$ |
| A284 | 1854-0071 | 18 | TRSTR:SI NPN | 28480 | 1854-0071 |
|  | 0698-6204 $0757-0442$ | 1 |  | ${ }^{28480}$ | 0698-6204 |
|  | -8757-0280 |  |  | 28480 <br> 28480 | - $\begin{aligned} & 0757-0442 \\ & 0757-c 280\end{aligned}$ |
| ( ${ }^{\text {a }}$ |  | 3 1 |  | ${ }^{2848 \mathrm{C}}$ | c757-6424 |
| ARR A2R7 |  | 1 |  |  | -06983223 |
|  | 2098-4002 $0757-049$ | $\stackrel{1}{1}$ |  | 28880 <br> 28480 | 0698-4002 $0757-0449$ |
|  |  | 1 |  | ${ }_{2848 \mathrm{C}}^{2888}$ | 0698-3158 |
| ${ }_{\text {a }}$ |  | $\frac{1}{1}$ |  | ciel |  |
| A 21212 $A_{2} 213$ | 069883260 $0.698-3260$ |  |  | 28480 28480 | O698-3260 $\mathbf{0 6 9 8 - 3 2 6 0}$ |
| A2R14 | $0698-3260$ $0698-3260$ |  |  | ${ }^{28480}$ | 06988-3260 |
|  |  | 1 |  | $\substack { 28886 \\ \begin{subarray}{c}{2846{ 2 8 8 8 6 \\ \begin{subarray} { c } { 2 8 4 6 } } \\{\hline 1846} \end{subarray}$ | - 0698883238080 |
| A2R17 $A_{2218} 18$ | 009883200 $0698-3260$ |  |  | 28880 28480 | Of983-3260 $0698-3260$ |
| A2219 AR20 | O6968-3250 n698-3260 |  |  | ${ }_{2}^{28480}$ | ${ }^{069883260}$ |
|  |  | 2 |  |  | -0698-3260 |
| A2R22 $A 2 R 23$ | ( 569880.0977 | 3 |  | 2888 C <br> 2848 C <br> 188 | - 06988 -0077 |
| ${ }_{\text {AR }}^{\text {AR24 }}$ | O698-3162 |  | RRFXD MET FLM 46.4 K OHM 18 1/8W | ${ }^{2848 \mathrm{C}}$ | 0698-3162 |
|  |  |  | SHICHERTAAY | 28880 <br> 28480 <br> 2848 |  |
| A3R1 | 069884400 $3130-193$ | 1 |  | 28880 <br> 2848 C | - $\begin{aligned} & \text { O698-3400 } \\ & 3130-0193\end{aligned}$ |
| ${ }_{\text {A }}^{\text {A }}$ 3A1 |  | 1 |  | 2848 C 28480 | 09505-c533 |
|  |  | ${ }_{3}^{1}$ | (e) | 28480 5688 5689 |  |
|  | c180-0116 c160-0939 |  |  | 56285 ${ }_{2848 \mathrm{c}}$ | ${ }_{\substack{150685 \times 503582-0 r}}^{160-0939}$ |
|  |  | 1 |  | cher 56289 | $1500684 \times 9035 A Z$-OVS |
| ${ }^{\text {A }} 4.6$ | 0180-0116 |  | C:FXD ELECT 6.8 UF 108 35VCM | 56289 | 15006859903562-ors |
|  | ( $\begin{aligned} & 0180-1735 \\ & 0180-1743\end{aligned}$ | 1 | C:FXO ELECT 0.22 UF $10 \times 3$ SVOCM | $\underset{\substack{2848 \mathrm{c} \\ 56285}}{\text { cen }}$ | O180-1735 $1500104 \times 935$ |
|  | O1800-0174 $0180-1745$ | 1 |  | 568889 <br> $\begin{array}{l}58880\end{array}$ <br> 8680 |  |
| A44, 4.11 | $0180-1745$ $0170-0059$ |  |  |  |  |
|  | 1901-020 ${ }^{\text {190, }}$ | 11 | Cilooe:silicon 100MA/VV | 56289 07263 07263 |  |
|  | (1901-2025 |  | Didotesilicon 100MA/IV | -07263 |  |
| ${ }_{4}{ }^{4} \times 1{ }^{\text {a }}$ | C490-0399 | 1 | RELAY:REED ASSY, 1200 Ohm 12 VOC | 28480 | 0490-0399 |
| A401 A 402 | 1855-0098 |  | TSIR:SI fen | 28480 <br> 28480 <br> 2880 | $1855-0098$ 18540071 |
| ${ }_{\text {A }}^{\text {A } 403}$ | $1854-3071$ $1853-020$ |  | ISRR: | ${ }^{28880}$ | ${ }^{1854540971}$ |
|  | $1853-020$ $1854-0071$ |  | TSRR:SI PNPRELECTEO FROM 2 N37021 | 28480 28480 | $1853-6020$ $1854-0071$ |
| A406 | $1853-0001$ $1854-0071$ | 1 | TSTR:SI PNPP SELECTED FROM 2N1132) |  | $1853-0001$ $1854-0071$ |
| ${ }_{\text {A }}^{\text {A } 4007}$ | - 1854.061 |  |  | 28480 <br> 28480 <br> 1 | $1854-0971$ $1854-0071$ |
| A409 A4010 | 185440071 $1854-071$ |  |  | 284880 28480 | $1854-0071$ $1854-c 271$ |
| $\begin{aligned} & A 4011 \\ & A 41 \\ & A 4 R 2 \\ & A 4 R \\ & A 4 R 3 \\ & A 4 R 3 \end{aligned}$ |  |  |  | $\begin{aligned} & 2848 \mathrm{C} \\ & 2880 \\ & 2886 \mathrm{C} \\ & 2888 \mathrm{C} \\ & \hline 2480 \end{aligned}$ |  |

Table 6-3. Replaceable Parts

| Reference Designation | HP Part Number | Oty | Description | Mfr Code | Mfr Part Number |
| :---: | :---: | :---: | :---: | :---: | :---: |
| A4R4 | 0757-0123 | 3 | R:FXD MET FLM 34.8K OHM 15 1/8W | 28480 | 0757-c123 |
| A4R5 | 0757-0346 |  | R:FXO MET FLM 10 OHM 19 1/8W | 28480 | 6 757-6346 |
| A4R6 | 0757-0458 | 1 | R:FXD MET FLM 51.1K OHM 18 1/8W | 28480 | 0757-C458 |
| A4R 7 | 0757-0465 | 2 | R:FXD MET FLM 100 K OHM $181 / 8 \mathrm{H}$ | 2848 C | 0757-0465 |
| A 4R8 | c757-0443 |  | R:FXD MET FLM L1.OK OHM 16 1/8W | 2848C | c757-0443 |
| A4R9 | 0757-0123 |  | R:FXD MET FLM 34.8K OHM 1\% $1 / 8 \mathrm{~W}$ | 2848 C | 0757-C123 |
| A4RIO | 2100-1759 | 2 | R:VAR WW 2 K OHM 5\% TYPE $V$ IW | 28480 | 2100-1759 |
| A4R11 | 0698-3156 | 4 | R:FXD MET FLM 14.7K OHM 18 1/8W | 2848 C | 3698-3156 |
| A4R12 | 0757-0442 |  | R:FXD MET FLM 10.0K OHM 1\% $1 / 8 \mathrm{M}$ | 28480 | 0757-0442 |
| A4R13 | 0757-0416 | 3 | R:FXD MET FLM 511 OHM 18 $1 / 8 \mathrm{~N}$ | 28480 | 0757-0416 |
| A 4 R 14 | 0757-0438 |  | R:FXD MET FLM S.11K OHM 16 1/8w | 2848 C | 0757-6438 |
| A4R15 | 210C-1761 | 1 | R:VAR WW IOK OHM 52 TYPE V IN | $28480^{\circ}$ | 2100-1761 |
| A4R16 | 0757-0441 | 3 | R:FXD MET FLM 8.25 K OHM 1\% 1/8H | 28480 | 0757-0441 |
| A4R17 | c757-0441 |  | R:FXD MET FIM 8.25K OHM 18 1/8\% | 28480 | 0757-c441 |
| A4R18 | 0757-0438 |  | R:FXD MET FLM 5.11K OHM $1 \% 1 / 8 \mathrm{H}$ | 2848 C | 0757-0438 |
| A4R19 | 0757-0462 |  | R:FXD MET FLM 75.0K OHM $121 / 8 \mathrm{H}$ | 28480 | -757-0462 |
| A4R20 | 0757-0401 |  | R:FXD MET FLM 100 OHM 1: $1 / 8 \mathrm{H}$ | 2848 C | 0757-0401 |
| A4R21 | 0698-3260 |  | R:FXD MET FLM 464 K OHM $151 / 8 \mathrm{~N}$ | 28480 | 0698-3260 |
| A4R22 | 0698-3155 | 4 | R:FXD MET FLM 4.64 K OHM 19 1/8W | 28480 | 0698-3155 |
| A4R23 | 0698-3454 | 1 | R:FXD MET FLM 215 K OHM 18 1/8W | 28480 | 0698-3454 |
| A4R24 | 0757-0443 |  | R:FXO MET FLM 11.CK OHM 12 1/8W | 2848 C | 0757-0443 |
| A4R25 | 0698-3156 |  | R:FXD MET FLM 14.7K OHM 18 1/8H | 28486 | 0698-3156 |
| A4R26 | 0698-3450 | 1 | R:FXD MET FLM 42.2 K OHM $1 \times 1 / 8 \mathrm{H}$ | 28480 | 0698-3450 |
| A4R27 A 428 | $0757-0465$ $0757-0442$ |  | R:FXD MET FLM 100 K OHM $121 / 8 \mathrm{~N}$ | 28480 | 0757-6465 |
| A4R28 | 0757-0442 |  | R:FXD MET FLM 10.0 K OHM $18 \mathrm{~K} 1 / 8 \mathrm{~W}$ | 28480 | 0757-0442 |
| A4R29 | 0698-3162 |  | R:FXD MET FiM 46.4 K OHM $131 / 8 \mathrm{~W}$ | 28480 | 0698-3162 |
| A4R30 | 0757-0444 | 4 | R:FXO MET FLM 12.1K OHM 1\% $1 / 8 \mathrm{~N}$ | 2848 C | 0757-0444 |
| A 4831 | 0698-3156 |  | R:FXD MET FLM 14.7 K OHM 18 1/8w | 28480 | 0698-3156 |
| A 401 A5 | $1826-0013$ $08554-60070$ | 3 | IC:LINEAR | 28480 | 1826-0013 |
| A5 | 08554-60070 | 1 | BOARD ASSY:MARKER GENERATOR | 2848 C | 08554-6C070 |
| A 5C1 | 0180-0197 |  | C:FXD ELECT 2.2 UF 10x 20VOCW | 56289 | 1500225×9020A2-DYS |
| ${ }^{\text {A SC }} 2$ | 0180-0197 |  | C:FXD ELECT 2.2 UF 108 ZOVDCH | 56285 | 1500225×9020A2-0YS |
| ${ }^{\text {A SCR }} 1$ | 1901-0025 |  | DIODE: SILICON IOOMA/IV | 07263 | FO 2387 |
| A5CR2 A5CR3 | 1901-0025 |  | DIODE:SILICON IOOMA/IV | 07263 | FD 2387 |
| A5CR3 | 1901-0025 |  | DICDE:SILICON LOOMA/IV | 07263 | FD 2387 |
| ASCR 4 | 1901-0025 |  | DIODE:SILICON 100ma/IV | 07263 | FD 2387 |
| ASCR 5 | 1910-0016 | 1 | OIODE:GE 60 WIV | 28480 | 1910-0016 |
| A5CR6 | 1901-0025 |  | DICDE:SILICON LOOMA/IV | 07263 | FD 2387 |
| ASCR7 |  |  | NOT ASSIGNED |  |  |
| A5CR8 | 1901-0025 |  | DIODE:SILICON LOOMA/IV | 07263 | Fo 2387 |
| A 501 | 1854-0071 |  | TSTR:SI NPNISELECTED FROM 2 N3704) | 28480 | 1854-0071 |
| A 502 | 1854-0071 |  | TSTR:SI NPN(SELECTED FROM 2N3704) | 28480 | 1854-C071 |
| ${ }^{4} 503$ | 1854-0071 |  | TSTR:SI NPN(SELECTED FROM 2N3704) | 28480 | 1854-C071 |
| ${ }^{\text {A }} 504$ | 1854-0071 |  | TSTR:SI NPN(SELECTED FROM 2N3704) | 28480 | 1854-0071 |
| 4595 | 1854-0053 | 1 | TSTR:SI NPN | 80131 | 2N2218 |
| A5R1 | 0757-0346 |  | R:FXD MET FIM 10 OHM 15 1/8w | 2848C | 0757-0346 |
| A5R2 | 0757-0346 |  | R:FXD MET FLM 10 OHM 1\% $1 / 8 \mathrm{~W}$ | 28480 | 0757-C346 |
| A5R3 | 0698-3449 |  | R:FXD MET FLM 28.7K OHM 1\% 1/8w | 28480 | 0698-3449 |
| A SR 4 A | 0757-0317 | 2 | R:FXD MET FLM 1.33K OHM $151 / 8 \mathrm{~W}$ factory selected part | 28480 | 0757-C317 |
| A 5 R 5 | 0698-3154 | 2 | R:FXD MET FLM 4.22K OHM 12 1/8W | 28480 | 0698-3154 |
| A5R6 | 0757-0288 | 2 | R:FXD MET FLM 9.09K OHM it 1/8H | 28480 | 0757-0288 |
| A5R7 | 0698-3457 | 2 | R:FXD MET FLM 316K OHM $181 / 8 \mathrm{H}$ | 28480 | 0698-3457 |
| A5R8 | 0757-0443 |  | R:FXD MET FLM 11.0K OHM 1\% $1 / 8 \mathrm{~W}$ | 28480 | 0757-C443 |
| ASR9 | 2100-1757 | 1 | R:VAR WH 500 OHM 59 TYPE $V$ IH | 28480 | 2100-1757 |
| A5R 10 | 0757-0139 | 1 | R:FXD MET FLM 1.1 MEGOHM $2 \% 1 / 2 \mathrm{~W}$ | 2848 C | 0757-0139 |
| A5R11 | 0698-3457 |  | R:FXD MET FLM 316K OHM 12 l 1/8M | 28480 | 0698-3457 |
| ASR12 | 0757-0123 |  | R:FXD MET FLM 34.8K OHM 1\% 1/8W | 28480 | 0757-0123 |
| A SR13 A 514 | 0683-3055 | 1 | R:FXD COMP 3 MEGOHM $531 / 4 \mathrm{~N}$ | 01121 | CB 3055 |
| A5R14 | 0698-3438 | 1 | R:FXD MET FLM 147 OHM 12 1/8w | 28480 | 0698-3438 |
| A5R15 | 0757-0447 | 1 | R:FXD MET FLM 16.2K OHM $121 / 8 \mathrm{H}$ | 28480 | 0757-0447 |
| A5R16 | 0757-0199 |  | R:FXD MET FLM 21.5 K OHM 18 1/8\% | 28480 | 0757-0199 |
| 45817 45818 | 0757-0289 | 1 | R:FXD MET FLM 13.3K OHM $151 / 8 \mathrm{H}$ | 28480 | 0757-0289 |
| A5R18 | 0757-0444 |  | R:FXD MET FLM 12.1K OHM 1\% $1 / 8 \mathrm{~B}$ | 28480 | 0757-0444 |
| A5R19 | 0698-3156 |  | R:FXD MET FLM 14.7K OHM 1\% $1 / 8 \mathrm{H}$ | 2848 C | 0698-3156 |
| A58. 20 | 0757-0438 |  | R:FXO MET FLM 5.11 K OHM $121 / 8 \mathrm{H}$ | 28480 | 0757-0438 |
| A5R21 | 0698-3151 | 2 | R:FXD MET FLM 2.87 K OHM 1\% 1/8H | 28480 | 0698-3151 |
| A5R22 | 2100-1758 | 1 | R:VAR WH 1 K OHM 58 TYPE $V$ IW | 28480 | 2100-1758 |
| A5R23 | 0757-0439 | 3 | R:FXD MET FLM 6.8IK OHM $151 / 8 \mathrm{H}$ | 28480 | 0757-0439 |
| A5R24 |  |  | NOT ASSIGNED |  |  |
| A5R28 A 529 |  |  | NOT ASSIGNED |  |  |
| A5R29 | 0764-0015 | 1 | R:FXD MET FLM 560 DHM 5\% 2 W | 28480 | 0764-0015 |
| A 5R30 | 0757-0442 |  | R:FXD MET FLM 10.0 K OHM $121 / 8 \%$ | 28480 | 0757-0442 |
| A5R31 | 0757-0280 |  | R:FXO MET FLM 1 K OHM $18 \mathrm{~L} 1 / 8 \mathrm{H}$ | 2848 C | 0757-0280 |
| A501 | 1826-0013 |  | IC:IINEAR | 28480 | 1826-0013 |

Table 6-3. Replaceable Parts

| Reference Designation | HP Part Number | Oty | Description | Mfr Code | Mfr Part Number |
| :---: | :---: | :---: | :---: | :---: | :---: |
| $\triangle 5 \vee 81$ | 1902-3139 | 1 | DITODE:BREAKDOWN 8.25V 5\% | 04713 | S110939-158 |
| A6 | 08554-60001 | 1 | YIG PCWER SUPPLY ASSY | 2848 C | c8554-6C00 1 |
| A 6 Cl 1 | 0180-0197 |  | C:FXO ELECT 2.2 UF 10\% 20VDCW | 56289 | 150022599020A2-OYS |
| $\triangle$ AC2 | 0180-0374 | 3 | C:FXD TANT. IC UF 10 \% 20 VOCW | 56289 | $1500106 \times 902082$-ors |
| A GC 3 | 2186-0229 | 1 | C:FXD ELECT 33 UF IO\% ICVOCW | 2848 C | 0180-0229 |
| ${ }^{\text {A C C }} 4$ | 2180-C374 |  | C:FXD TANT. 10 UF $10 \% 2 C V O C W$ | 56289 | $1500106 \times 902082-0 Y S$ |
| ${ }^{46 C 5}$ | 0160-2162 | 1 | C:FXD MY 0.C22 UF $10 \% 200 \mathrm{VDCH}$ | 56289 | 192P22392-PTS |
| A6C <br> A 67 |  |  | NOT ASSIGNED ${ }^{\text {C:FXD ELECT } 220 ~ U F ~} 202$ IoVDCH |  |  |
| A $6 C 5$ A 6 \% | 2186-0159 $0150-0023$ | 1 | C:FXD ELECT 220 UF 20\% 10VOCH $C$ : FXD CER 2000 PF $20 \%$ locovocw | 28480 56289 | 0180-0159 20C295A2-CDH |
| A CCs | c180-010 | 1 | C:FXO ELECT 4.7 UF Le\% 35VOCH | 56289 | 1500475*903582-DYS |
| AGClo | c140-. 3198 | 1 | C:FXD MICA 200 PF 5\% | 72136 | RDM15F201J3C |
| AGCR1 | 1902-6033 | 1 | DIODE: BREAKDOWN 6.2V | 04713 | 1 N823 |
| $\mathrm{ACCR}^{4}$ | 1902-0048 | 1 | DICDE: BREAKOOWN 6.8IV 5\% | 04713 | S110939-134 |
| A6CR3 | 1901-C025 |  | DIOOE:SILICON L 100 ma IV | 07263 | FO 2387 |
| A601 | 1854-C221 | 1 | TSTR:SI NPNIREPL. BY 2N4044) | 28480 | 1854-0221 |
| $\triangle 602$ | 1854-0071 |  | TSTR:SI NPN(SELECTED FROM 2N3704) | 28480 | 1854-0071 |
| A603 | 1854-C071 |  | TSTR:SI NPN(SELECTED FROM 2N3734) | 28480 | 1854-0071 |
| ${ }^{4604}$ | 1854-0071 |  | TSTR:SI NPN(SELECTED FROM 2N3704) | 28480 | 1854-0071 |
| A605 | 1853-0020 |  | TSTR:SI PNP(SELECTEO FROM 2N3702) | 28480 | 1853-0020 |
| A606 | 1853-6020 |  | TSTR:SI PNP(SELECTED FRCM 2N3702) | 28480 | 1853-c020 |
| A607 | 1853-0020 |  | TSTR:SI PNP(SELECTED FROM 2N3702) | 28480 | 1853-0020 |
| A608 | 1854-0063 | 1 | TSTR:SI NPN | 80131 | 2N3055 |
| A6R 1 | 0757-0416 |  | R:FXD MET FLM 511 OHM 1\% $1 / 8 \mathrm{~W}$ | 28480 | 0757-C416 |
| A6R2 | 2100-1759 |  | R:VAR WW 2K OHM 5\% TYPE V 1 W | 28480 | 2100-1759 |
| A68 3 | 0757-c317 |  | R:FXO MET FLM 1.33K OHM $121 / 8 \mathrm{~N}$ | 28480 | 0757-0317 |
| ASR 4 | 0757-6467 | 2 | R:FXD MET FLM 121K OHM $181 / 8 \mathrm{~W}$ | 28480 | 0757-0467 |
| A6R 5 | 0698-c 384 |  | R:FXD MET FLM 2.15 K OHM $1 \% 1 / 8 \mathrm{H}$ | 28486 | 0698-0084 |
| A6R6 | 0698-3132 |  | R:FXO FLM 261 OHM 1\% $1 / 8 \mathrm{H}$ | 2848C | 0698-3132 |
| 4687 | 2100-1754 | 1 | R:VAR WW 50 OHM 58 TYPE V IW | 28480 | 2100-1754 |
| A6R8 | 0757-0415 | 1 | R:FXD MET FLM 475 OHM 1\% $1 / 8 \mathrm{NW}$ | 28480 | 0757-0415 |
| A6R9 | 0698-3159 | 1 | R:FXD MET FLM 26.1 K OHM 18 1/8W | 2848C | 0698-3159 |
| AGRIC | 0757-0280 |  | R:FXD MET FLM IK OHM 18 1/8H | 28480 | 0757-C280 |
| A6R11 | 0698-3442 | 1 | R:FXD MET FLM 237 OHM 1\% 1/8W | 2848 C | 0698-3442 |
| A6R12 | 0757-0444 |  | R:FXD MET FLM 12.1K OHM $1 \% 1 / 8 \mathrm{~N}$ | 2848 C | 0757-0444 |
| A6R13 | 0757-028C |  | R:FXO MET FLM 1 K OHM $1 \% 1 / 8 \mathrm{H}$ | 28480 | 0757-0280 |
| A6R14 | 0757-0279 | 2 | R:FXD MET FLM 3.16K OHM 1\% 1/8W | 28480 | 0757-C279 |
| AGR15 | 0683-5145 | 1 | R:FXO COMP 510K OHM 5\% 1/4w | 01121 | CB 5145 |
| AGR16 | 0698-5465 | 1 | R:FXO FLM 4.725K OHM $12 \mathrm{l} / 8 \mathrm{BH}$ | 28480 | 0698-5465 |
| A6R17 | 0757-0467 |  | R:FXD MET FLM 121K OHM $181 / 8 \mathrm{~W}$ | 28480 | 0757-0467 |
| A6R18 | 2100-1760 | 1 | R:VAR WW 5K OHM 5\% TYPE V IW | 28480 | 2100-1760 |
| AGR19 | 0698-3157 |  | R:FXD MET FLM 19.6K OHM is 1/8K | 28480 | 0698-3157 |
| A6R20 | 0757-0441 |  | R:FXD MET FLM 8.25 K OHM 1\% $1 / 8 \mathrm{~K}$ | 28480 | 0757-0441 |
| A6R 21 | 0698-0084 |  | R:FXD MET FLM 2.15 K OHM $151 / 8 \mathrm{H}$ | 28480 | 0698-0084 |
| A6R22 | c757-0420 |  | R:FXO MET FLM 750 OHM $12 \mathrm{1/8w}$ | 28480 | 0757-0420 |
| A6R23 | 0698-3437 |  | R:FXD MET FLM 133 OHM 15 1/8W | 28480 | 0698-3437 |
| $46 R 24$ | 0757-0466 | 1 | R:FXD MET FLM 110K OHM 12 1/8W | 28486 | 0757-0466 |
| A6R25 | C698-3154 |  | R:FXD MET FLM 4.22K OHM 18 1/8W | 28480 | 0698-3154 |
| A6R26 | 0757-029C | 1 | R:FXD MET FLM 6.19K OHM 1\% $1 / 8 \mathrm{~N}$ | 28480 | 0757-0290 |
| A6R27 | 0757-0279 |  | R:FXD MET FLM 3.16K OHM 1\% 1/8H | 28480 | 0757-0279 |
| A6R28 | 0683-9145 | 1 | R:FXO COMP gIok ohm $521 / 4 \mathrm{H}$ | 01121 | CB 9145 |
| A6R 29 | C757-0288 |  | R:FXD MET FLM 9.09K OHM is 1/8\% | 28480 | 0757-0288 |
| 46839 | 0698-3445 | 1 | R:FXD MET FLM 348 OHM 12 1/8w | 28480 | 0698-3445 |
| A6R31 AGR 32 | 0757-0401 |  | NOT ASSIGNED R:FXD MET FLM 100 OHM I\% 1/8W | 28480 | 0757-0401 |
| A6R 33 | 0757-0444 |  | R:FXO MET FLM 12.1K OHM 1\% 1/8w | 28480 | 0757-0444 |
| A6R34 | 0698-3160 | 2 | R:FXD MET R:FXD MET FLM 31.6K OHM | 28480 28480 | 0757-C444 $0698-3160$ |
| A6R34 |  |  | FACTORY SELECTED PART |  | -698-3160 |
| AGR35 | 0757-0418 | 1 | R:FXO MET FLM 619 OHM 1\% $1 / 8 \mathrm{BW}$ | 28480 | 0757-0418 |
| A6R36 | 0757-0439 |  | R:FXD MET FLM 6.81K OHM 19 1/8\% | 28480 | 0757-C439 |
| A6R37 | 0757-c439 |  | R:FXD MET FLM 6.8LK DHM I2 1/84 | 28480 | 0757-0439 |
| A6R38 | 0698-3160 |  | R:FXD MET FLM 31.6K OHM 1\% 1/8\% | 28480 | 0698-3160 |
| A6838 |  |  | FACTORY SELECTED PART |  |  |
| A6R39 | 0698-3151 |  | R:FXD MET FLM 2.87K OHM 18 1/8w | 28480 | 0698-3151 |
| A6U1 | 1826-0013 |  | IC:LINEAR | 28480 | 1826-0013 |
| A 7 | 08554-60074 | 1 | Yig oscillator assy | 2848C | 08554-60074 |
| A 7 | 08554-80075 | 1 | REBUILT 08554-60074, REQUIRES EXCHANGE | 28480 | 08554-60075 |
| A7AT1 | 08554-60058 | 1 | ATTENUATOR:20B | 2848 C | 08554-80058 |
| A8 | 08554-60012 | 1 | FIRST ANO SECOND CONVERTER ASSY | 2848 C | 08554-60012 |
| A8 | C8554-60051 | 1 | FIRST/SECONO CONVERTER EXCHANGE ASSY | 28480 | 08554-60051 |
| $A B C 1$ | 0160-2437 |  | C:FXO CER 5000 PF +80-20\% 200VOCK | 72982 | 2425-000-x5v-502P |
| $A 8 C 2$ | 0160-2437 |  | C:FXO CER 5000 PF $+80-20 \% 200 \mathrm{VOCH}$ | 72982 | 2425-c00-X5V-502P |
| $\triangle 8 C 3$ | 0140-008C | 1 | C:FXD MICA 15 PF 1085 SOVOCH | 00853 | TYPE H100 Elo |
| ABC4 | 014c-c069 | 1 | C:FXD MICA 550 PF IOX SOOVDCW | 00853 | TYPE M ICOE 10 |
| $A B C 5$ | c160-c345 | 2 | C:FXD CER FEED-THRU 1000 PF 500 VOCW | 01121 | F828-102W |

Table 6-3. Replaceable Parts

| Reference Designation | HP Part Number | Oty | Description | Mfr Code | Mfr Part Number |
| :---: | :---: | :---: | :---: | :---: | :---: |
| ABCR 1 | 1901-0535 | 1 | DICDE:HYBRIO HOT CARRIER | 28480 | 1901-C535 |
| A8J1 | 1250-0829 |  | CONNECTOR:RF 50-OHM SCREW ON TYPE | 98291 | 50-045-4610 |
| A8J2 | 1250-0829 |  | CONNECTIR:RF 50-OHM SCREW ON TYPE | 98291 | 50-045-4610 |
| ABMP 1 ABMP 2 | 08554-00016 | 1 | COVER:FIRST \& SECOND CONVERTER NOT ASSIGNED | 28480 | 08554-00016 |
| $48 \mathrm{MP3}$ | 08554-00005 | 1 | COVER:550 MHz IF | 28480 | 08554-00005 |
| $\mathrm{A}^{\text {AMP }} 4$ | 08554-20019 | 1 | BOOY:FIRST \& SECOND CONVERTER | 2848 C | 08554-20019 |
| A8MP5 | 08555-00033 | , | INPUT-GUTPUT LOOP | 28480 | 08555-0.033 |
| A8MP6 A8MP | $08555-20040$ $0855-20041$ | 1 | CAP:CUTER ELEMENT | 28480 | 08555-20040 |
| A8MP7 | 08555-20041 | 1 | CAP:INNER ELEMENT | 28480 | 08555-20041 |
| A8MP8 | 08555-20042 | 1 | CAP:DIELECTRIC | 28480 | 08555-20042 |
| A8R1 | 0757-042C |  | R:FXD MET FLM 750 OHM 1\% 1/8w | 28480 | 0757-0420 |
| ABAI | $08554-60042$ | 1 | FIRST MIXER ASSY | 2848C | 08554-60042 |
| A8AICRI ABAICR2 | 1901-0385 | 2 | DIODE:SI, MATCHED PAIR, HOT CARRIER DIODE:SI, MATCHED PAIR, HOT CARRIER | 28480 28480 | 1901-0385 |
| A8A1U1 | 1250-1157 | 1 | CONNECTOR:RF SMA SERIES | 16179 | OSM 220 |
| A8Alu2 | 1250-0828 | 4 | CONNECTOR:RF SC-OHM SCREW ON TYPE | 98291 | 50-643-4610 |
| A8A1MPI | 08554-20015 | 1 | COVER:INPUT MIXER | 28480 | 08554-20015 |
| A8A2 | 5086-7082 |  | FILTER: LOW-PASS, 5 GHz | 28480 | 6960-0060 |
| AbA3 | 08554-60021 | 1 | SECONO Local osćillator assy | 28480 | 08554-60021 |
| A8A 3C1 | 0160-4052 | 1 | C: FXD PORC. 0.6 PF 500 VDCW <br> factory selected part; typilal value given | 28480 | 0160-4052 |
| ABA3MP1 | 08554-00012 | 1 | COUPLING:SECOND LOCAL OSC LOOP | 28480 | 08554-00012 |
| A84301 A8A302 | $1854-0292$ $1854-0292$ | 3 | TSTR:SI NPN | 28480 | 1854-C292 |
| A84302 | 1854-0292 |  | TSTR:SI NPN | 28480 | 1854-C292 |
| A8A3R1 | 0757-0424 |  | R:FXO MET FLM 1.1OK OHM 12 1/8W | 28480 | 0757-C424 |
| A8A3R2 | 0757-0346 |  | R:FXD MET FLM 10 OHM 12 l 1/8H | 28480 | 0757-0346 |
| A8A3R3 | $0757-0424$ $08554-60081$ | 1 | R:FXD MET FLM 1.1OK OHM 12 $1 / 8 \mathrm{~W}$ BOARD ASSY:THIRO CONVERTER | 28488 | 0757-0424 |
|  |  |  | BCARD ASSY:THIRO CONVERTER | 28480 | 08554-60881 |
| A9C1 | 0160-0345 |  | C:FXD CER FEED-TMRU 1000 PF 500VOCH | 01121 | F828-102w |
| $49 C 2$ $49 C 3$ | 0160-2437 |  | C:FXD CER 5000 PF $+80-20 \pm 200 \mathrm{VDCH}$ | 72982 | 2425-000-x5V-502P |
| A9C3 | 0160-2437 |  | C EFXD CER 5000 PF +80-20\% 200VOCW | 72982 | 2425-000-x5V-502P |
| A9C4 <br> 4965 | 0160-2437 |  | C:FXD CER 5000 PF $+80-208$ 200VOCW | 72982 | 2425-000-X5V-502P |
| A 4 Cs | 0160-2437 |  | C:FXD CER 5000 PF $+80-20 \pm 200 \mathrm{VDCW}$ | 72982 | 2425-000-x5v-502P |
| A9C6 A9C7 | $0160-2437$ $3030-0382$ |  | C:FXD CER 5000 PF +80-202 200VDCH | 72982 | 2425-000-X5V-502P |
| A9C7 | 3030-0382 | 3 | SCREM:SET(LOCKING) | 72962 | 350063 |
| A9C8 | 3030-0382 |  | SCREW:SETILOCKING) | 72962 | 850063 |
| A9C9 | 3030-0382 |  | SCREM: SET(LOCKING) | 72962 | 850063 |
| A9J1 | 1250-0828 |  | CONNECTOR:RF 50-OHM SCREW ON TYPE | 98291 | 50-043-4610 |
| 09.12 | 1250-0828 |  | CONNECTOR:RF 50-OHM SCREW ON TYPE | 98291 | 50-043-4610 |
| A9J3 | 1250-0828 |  | CONNECTOR:RF 50-OHM SCREW ON TYPE | 98291 | 50-043-4610 |
| A9L1 | 9100-2839 | 3 | INDUC TOR: FXO | 28480 | 9100-2839 |
| A9L2 | 9100-2839 |  | INDUCTOR:FXD | 28480 | 9100-2839 |
| A9L3 | 9100-2839 |  | INDUCTOR:FXO | 28480 | 9100-2839 |
| A9MP 1 | 08554-00008 | 1 | COVER:500MHZ OSCILLATOR | 28480 | 08554-00008 |
| A9MP2 | 0855400009 | 1 | COVER:MIXER-AMPLIFIER | 28480 | 28554-00009 |
| A9MP3 | 08554-00015 | 1 | COUPLING:LO DUTPUT | 28480 | 08554-06015 |
| A9MP4 | 08554-20032 | 1 | HOUSING: THIRD CONVERTER | 28480 | 08554-20032 |
| A9MP5 | 08554-20035 | 1 | SUPPORT:FILTER COIL | 28480 | 08554-20035 |
| A9MP6 | 08554-20038 | 1 | CONTACT:FILTER | 28480 | 08554-20038 |
| A9MP 7 | 08554-60017 | 1 | COUPLING:500MHZ | 28480 | 08554-60017 |
| A9R1 | 0698-7200 | 1 | R:FXD FLM 31.6 OHM $221 / 8 \mathrm{M}$ | 28480 | 0698-7200 |
| A9AL | 08554-60009 | 1 | AMPLIFIER ASSY:550MHZ NOT ASSIGNED | 28480 | 08554-60009 |
| A9A1C2 | 0160-2248 | 1 | C: FXD CER 4.3 PF 500VDCW | 28480 | 0160-2248 |
| A9A1C3 | 0150-0093 | 1 | C:FXD CER 0.01 UF +80-209 100VDCW | 72982 | 801-K800011 |
| A9A1C4 | 0160-2266 |  | C:FXD CER 24 PF 52500 VOCW | 72982 | 301-000-COGO-240J |
| A9A1C5 | 0160-2247 | 2 | C:FXD CER 3.9 PF 500VDCH | 72982 | 301-NPO-3.9 PF |
| A9A1J1 | 1250-1220 | 2 | CONNECTOR: RF SO OHM SCREM-ON TYPE | 98291 | 50-051-0109 |
| A9A1MPI | 2190-0326 | 1 | WASHER:FLAT O.115* 10 | 00000 | 080 |
| A9ALO1 | 1853-0020 |  | TSTR:SI PNPISELECTED FROM 2N3702) | 28480 | 1853-0020 |
| A9A102 | 1854-0292 |  | TSTR:SI NPN | 28480 | 1854-0292 |
| A9A1R1 | 0698-3155 |  | R:FXD MET FLM 4.64K OHM 1\% 1/8W | 28480 | 0698-3155 |
| A9A1R2 | 0757-0443 |  | R:FXD MET FLM 11.0K OHM 15 1/8w | 28480 | 0757-0443 |
| A9A1R3 | 0698-3155 |  | R:FXO MET FLM 4.64 K OHM 1\% 1/8W | 28485 | 0698-3155 |
| ASAIR4 | 0757-0280 |  | R:FXD MET FLM 1 K OHM $1 \% 1 / 8 \mathrm{~W}$ | 28486 | 0757-6280 |
| A9A1R5 | 0757-0416 |  | R:FXD MET FLM 511 OHM $121 / 8 \mathrm{H}$ | 28480 | 0757-0416 |
| A9A2 | 08554-60080 | 1 | LO DRIVE ASSY:500 MHZ | 2848C | 08554-66080 |
| A9A2C1 | 0180-0374 |  | $C: F X D$ TANT. 10 UF 10220 VDCH | 56289 | 1500106×902082-DYS |
| A9A2C2 | c18c-0197 |  | C:FXO ELECT 2.2 UF 102 20VDCH | 56289 | 1500225x9020A2-DYS |
| A9A2C3 | 0170-0066 | 1 | C:FXD MY 0.027 UF $10 x 200 \mathrm{VDCH}$ | 56289 | 192P27392-PTS |
| A9A2C4 | 0180-0197 |  | C:FXO ELECT 2.2 UF $10 \pm 20 \mathrm{VDCW}$ | 56289 | 1500225x9620A2-ors |
| A9A2CR1 | 1901-0025 |  | DIODE:SILICON $100 \mathrm{ma/IV}$ | C7263 | FD 2387 |
| a9azal | 1853-0020 |  | TSTR:SI PNP(SELECTED FROM 2N3702) | 28480 | 1853-c020 |

Table 6-3. Replaceable Parts


Table 6-3. Replaceable Parts

| Reference Designation | HP Part Number | Oty | Description | Mfr Code | Mfr Part Number |
| :---: | :---: | :---: | :---: | :---: | :---: |
| AlO | 08554-60020 | 1 | Phase lock amplifier assy | 2848¢ | 08554-60020 |
| A 10 Cl | 0160-2437 | 13 | C:FXO CER 5000 PF +80-20\% 200VOCW | 72982 | 2425-000-X5V-502P |
| A 10 C 2 | 0160-2437 |  | C:FXD CER 50CO PF $+80-20 \% 200 \mathrm{VDCW}$ | 72982 | 2425-000-x5v-502P |
| A10C3 | 0160-2152 | 1 | C:FXD CER 10 PF 2C\% 5COVOCW | 2848 C | c160-2152 |
| A10C4 | 0160-2437 |  | C:FXD CER 50C0 PF $+80-20 \%$ 2COVDCW | 72982 | 2425-000-x5v-502P |
| A10C5 | 0160-2437 |  | C:FXO CER $5003 \mathrm{PF}+8 \mathrm{C}-20 \% 200 \mathrm{VDCW}$ | 72982 | 2425-600-x5V-502P |
| Al0MP1 | 08554-00610 | 1 | COVER:LOCK BOX | 2848 C | C8554-00310 |
| Alomp 2 | 08554-20C34 | 1 | HOUSING: PHASE LOCK | 2848 C | 08554-20034 |
| A1041 | 08554-60018 | 1 | SAMPLER PULSE GENERATOR ASSY | 28486 | 08554-60018 |
| Aloalci | 0180-0197 | 12 | C:FXD ELECT 2.2 UF 10 O 20VOCW | 56289 | 150022599020A2-DYS |
| Aloalcz | 0180-0197 |  | C:FXD ELECT 2.2 UF $10 \%$ 20VOCH | 56289 | 1500225×9020A2-DYS |
| A 1041 Cl 3 | 0180-0197 |  | C:FXD ELECT 2.2 UF 108 2OVDCW | 56289 | 1500225×9020A2-DYS |
| Al0alc 4 | 0160-0161 | 1 | C:FXD MY 0.01 UF 108200 VDCH | 56289 | $192910392-\mathrm{PTS}$ |
| Al0a1C5 | 0180-0197 |  | C:FXO ELECT 2.2 UF 102 E 20VOCW | 56289 | 1500225X9020A2-DYS |
| Alcalcb | 0160-2204 | 1 | C:FXO MICA 1COPF 5\% | 72136 | ROM15FIO1J3C |
| Aloalc ${ }^{\text {a }}$ | 0180-0197 |  | C:FXD ELECT 2.2 UF 10\% 20VDCW | 56289 | 15C0225×902042-DYS |
| Al0alC8 | 0160-2201 | 1 | C:FXD MICA 51 PF 5\% | 72136 | ROM15E510JIC |
| Al0AlCR1 | 1901-0081 |  | OIODE:SILICON 50 VOLTS WORKING | 07263 | FD1415 |
| Aloalle | $1901-C 081$ $9140-6129$ |  | OIDDE:SILICON 50 VULTS WORKING COIL:FXD RF 220 UH | 07263 2848 C | F01415 $9140-0129$ |
| Aloall 3 | 9100-1612 | 1 | CO1L:FXD RF 0.33 UH 20\% | 28480 | 9160-1612 |
| Al0a101 | 1854-0042 |  | TSTR:SI NPN | 28480 | 1854-0042 |
| A10A102 | 1854-0042 |  | TSTR:SI NPN | 2848 C | 1854-0042 |
| A104103 | 1853-0018 | 1 | TSTR:SI PNP(SELECTED FROM 2N42601 | 28480 | 1853-0018 |
| Al0a104 | 1854-0354 | 1 | TSTR:SI NPN | 28480 | 1854-0354 |
| Al0AlR1 | 0757-0346 |  | R:FXD MET FLM 10 OHM $151 / 8 \mathrm{H}$ | 2848 C | 0757-c346 |
| A10A1R2 | 0757-0346 |  | R:FXO MET FLM 10 OHM $1 \mathrm{~F} 1 / 8 \mathrm{~W}$ | 28480 | c757-c346 |
| A1041R3 | 0698-3132 | 2 | R:FXD FLM 261 DHM 19 1/8W | 28480 | 0698-3132 |
| Al0A1R4 | 0698-3153 | 5 | R:FXD MET FLM 3.83 K OHM $121 / 8 \mathrm{~W}$ | 2848C | 0698-3153 |
| Aloalrs | 0698-3153 |  | R:FXO MET FLM 3.83K OHM $121 / 8 \mathrm{~W}$ | 28480 | 0698-3153 |
| AloAlRg | 0757-0346 |  | R:FXO MET FLM 10 OHM 1\% $1 / 8 \mathrm{~W}$ | 2848 C | 0757-c346 |
| Al0A1R7 | 0757-0280 | 9 | R:FXD MET FLM 1 K OMM 12 L 1/8W | 2848 C | 0757-0280 |
| A10A1R8 | 0698-3153 |  | R:FXD MET FLM 3.83 K CHM 1\% $1 / 8 \mathrm{H}$ | 28480 | 0698-3153 |
| Al0A1R9 | 0698-3153 |  | R:FXD MET FLM 3.83K OHM 18 L 1/8W | 28480 | 0698-3153 |
| Aloalric | 0757-0346 |  | R:FXD MET FLM 10 OHM 17 1/8N | 2848 C | 0757-c346 |
| AlOAIRIl | 0698-3153 |  | R:FXD MET FLM 3.83K OHM $151 / 8 \mathrm{~N}$ | 28480 | 0658-3153 |
| Al0alR12 | 0698-3408 | 1 | R:FXO MET FLM 2.15 K OHM $17 \mathrm{~L} 1 / 2 \mathrm{~W}$ | 28480 | 0698-3408 |
| Al0alR13 | 0757-0346 |  | R:FXD MET FLM 10 OHM $181 / 8 \mathrm{H}$ | 28480 | 0757-C346 |
| A10A1R14 | 2100-1774 | 2 | R:VAR WH 2K OHM 5\% TYPE H IW | 28480 | 2100-1774 |
| Aloalxyl | 1200-0770 | 1 | SOCKET:CRYSTAL | 91506 | 8000-AG-26 |
| Al0aly | 0410-0013 | 1 | CRYSTAL: OUARTZ 1 M ${ }^{\text {C }}$ | 2848 C | 0410-0013 |
| Al0az | 08554-60004 | 1 | PHASE LOCK AMPLIFIER ASSY | 2848 C | 08554-60004 |
| A10A2Cl | 0180-0197 |  | C:FXD ELECT 2.2 UF 10 S 20VOCH | 56289 | 1500225×9020A2-0YS |
| A10A2C2 | 0180-0197 |  | C:FXD ELECT 2.2 UF $10 \% 20 \mathrm{VDCW}$ | 56285 | 1500225x502042-0YS |
| AlOA2C3 | 0160-3153 | 2 | C:FXD GLASS C.01 UF $20 \% 50 \mathrm{VOCH}$ | 14674 | Crkol bt 103M |
| A10A2C4 | 0160-3153 |  | C:FXD GLASS C.01 UF 20\% 50VDCW | 14674 | Crkal 8 t 103 m |
| A10A2C5 | 0160-2259 |  | C:FXD CER 12 PF 58500 VDCW | 72982 | 301-c00-COGO-120J |
| A1042C6 | 0160-0153 | 1 | C:FXD MY 0.001 UF 10\% 200VDCH | 56289 | 192P10292-PTS |
| A1042C7 | 0160-2257 | 1 | C:FXD CER 10 PF 5\% 500 VDCW | 72982 | $3 \mathrm{CI}-\mathrm{COO}-\mathrm{COHO}-10 \mathrm{CJ}$ |
| A10A201 | 1855-0050 |  | TSTR:SI FET DUAL | 28480 | 1855-0050 |
| A104292 | 1853-0020 | 10 | ISTR:SI PNPISELECTED FROM 2N3702) | 2848 C | 1853-0020 |
| A104203 | 1853-0020 |  | TSTR:SI PNP(SELECTED FROM 2N3702) | $2848{ }^{\text {C }}$ | 1853-0020 |
| A104204 | 1854-0071 | 18 | TSTR:SI NPNISELECTED FROM 2N37041 | 28480 | 1854-0071 |
| A10ARRI | 0698-3430 | , | R:FXD MET FLM 21.5 OHM $12 \mathrm{1/8W}$ | 28480 | 0658-343C |
| AloazR2 | 0698-3430 |  | R:FXD MET FLM 21.5 OHM 1\% $1 / 8 \mathrm{~W}$ | 28480 | 0698-3430 |
| Al0 A2R 3 | 0757-0488 | 2 | R:FXD MET FLM 909K OHM 18 1/8W | 28480 | 0757-0488 |
| A10A2R4 | 2100-2650 | 1 | R:VAR FLM 200 K OHM 107 LIN $1 / 2 \mathrm{~W}$ | 28480 | 2100-2650 |
| Al0azrs | 0757-0488 |  | R:FXD MET FLM 9C9K OHM 15 1/8H | 28480 | 0757-6488 |
| A10A2R6 | 0757-0199 | 3 | R:FXD MEI FLM 21.5 K OHM $181 / 8 \mathrm{H}$ | 2848 C | 0757-6199 |
| Al0A2R7 | 0698-0082 | 1 | R:FXD MET FLM 464 OHM $181 / 8 \mathrm{H}$ | 28480 | 0698-0082 |
| A 10A2R8 | 0757-0443 | 5 | R:FXD MET FLM 11.OK OHM 17 1/8w | 28480 | 0757-0443 |
| Al0A2R9 AlOALR10 | $0757-0199$ $0698-0084$ |  | R:FXO MET FLM 21.5K OHM 1\% 1/8W | 2848 C | 0757-6199 |
| Aloazr 10 Aloazril | 0698-0084 $0698-0084$ | 4 | R:FXD MET FLM 2.15 K OHM 18 1/8W R:FXD MET FIM 2 | 28480 | 0698-0084 |
| AloazR12 | 0698-0084 |  | R:FXD MET FLM 2.15K OHM $1 \% ~ 1 / 8 \mathrm{~W}$ NOT ASSIGNEO | 28480 | 0698-0084 |
| A10A2R13 | 0757-0442 | 6 | R:FXD MET FLM 1C.OK OHM $181 / 8 \mathrm{~W}$ | 28480 | 0757-0442 |
| A 10 A 2 R 14 | 0698-3157 | 2 | R:FXD MET FLM 19.6 K OHM 18 1/8w | 2848 C | C698-3157 |
| A10A2R15 | 0698-3437 |  | R:FXD MET FLM 133 OHM 1\%, $1 / 8 \mathrm{~W}$ | 28480 | 0698-3437 |
| Al0a2R16 | 0698-3161 |  | R:FXD MET FLM 38.3K OHM $151 / 8 \mathrm{~W}$ | 2848 C | c698-3161 |
| Aloazr 17 | 0757-0438 | 4 | R:FXD MET FLM 5.11K OHM 18 1/8w | 2848 C | 0757-0438 |
| Al0a3 | 5086-7042 | 1 | SAMPLER ASSY: $2-3.3 \mathrm{GHZ}$ | 2848 C | 5086-7042 |
| All |  |  | NOT ASSIGNED |  |  |
| A12 | 08554-60048 | 1 | AMPLIFIER ASSY: 50 MHZ | 28480 | 08554-60048 |
| A12C1 | 0160-2437 |  | C:FXD CER 5000 PF $+80-208200 \mathrm{VDCW}$ | 72982 | 2425-000-x5v-5C2P |
| A12C2 | 0160-2437 |  | C:FXD CER 5000 PF +80-20\% 2 COVOCW | 72982 | 2425-000-x5v-502P |

Table 6-3. Replaceable Parts


Table 6-3. Replaceable Parts

| Reference Designation | HP Part Number | Oty | Description | $\mathrm{Mfr}$ Code | Mfr Part Number |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  | $\begin{aligned} & 08554-60071 \\ & 11593 A \\ & 11593 A \end{aligned}$ | $\frac{1}{2}$ |  | $\begin{gathered} 28800 \\ 280 \\ 28980 \end{gathered}$ |  |
|  | 0180-0155 | $\frac{1}{1}$ |  <br>  DISPLAY UNCAL |  | $1500225 \times 0020 A 2-0 Y S$ $300107 G 025002-0 S M$ <br> CM8-1099 |
| ost | ${ }^{1450-0153}$ | 1 | LaMphlolerfecr $1-1$ SERIES | 08717 | 10258 |
|  |  | $\stackrel{1}{1}$ |  |  |  |
| ${ }_{5}$ | -88554-600999 | $\frac{1}{1}$ |  | ${ }_{\substack{28880 \\ 28489}}^{208}$ | - 085545406049 |
|  |  | 1 | Boorip convector | 边 22860 |  |
|  |  | $\stackrel{1}{1}$ |  |  |  |
| $\begin{aligned} & J 1 M P 6 \\ & J 1 M P 7 \\ & J 1 M P 8 \end{aligned}$ |  | 1 | HOLDER:CAPACITOR BOARD BDARD:CAPACITOR C:FXD CER O. 1 UF $20 \geq 100 \mathrm{VDCH}$ |  |  |
| jilleg |  | $\frac{1}{1}$ | - | coick | cose |
| лıpp11 | 08554-20083 | 1 | cable: capactior to attenatorini). | ${ }^{2488}$ | 08554-20083 |
| $\begin{aligned} & j_{2} \\ & \substack{3 \\ y_{1} \\ p 1 \\ \hline} \end{aligned}$ | $9100-1642$ | 1 |  <br>  | 28480 | 9100-1642 |
| $\begin{aligned} & p_{2} p_{1} p_{3} p_{1} \\ & k_{2} \end{aligned}$ |  | ${ }_{1}^{1}$ | CONNECTOR:MALE 24 CONTACTS <br> SHIELD:CONNECTOR $\quad$ CONNECTORER AND P 41 MALE CONTACT NOT ASSIGNED NOT ASSIGNED | $\begin{gathered} 288480 \\ \substack{2868 \\ 7468} \end{gathered}$ | $1251-0055$ $08555-00002$ DDM-43H2-P |
| ¢ |  | ${ }_{1}^{1}$ | $\cdots$ Revar |  |  |
| ( ${ }_{\text {R }}^{8 .}$ | $2100-261$ | 1 | NOT ASSIGNED R:VAR CERMET $2 K$ OHM 10Z LIN $2 W$ NOT ASSIGNED | 28480 | 100- |
| ¢88 | 0812-2788 | 1 |  | ${ }_{\substack{28480 \\ 2880}}^{\substack{\text { a }}}$ | ${ }^{0811-2788}$ |
| cis |  | $\frac{1}{1}$ |  | coick |  |
| ${ }_{\text {R11 }}$ |  | $\stackrel{1}{1}$ | TERHINAION:RF So ont screw on tre | ¢ | $\substack{\text { coo-001-0101 } \\ 3101-1500}$ |
| - | -8855t-20082 | ${ }_{1}^{1}$ |  | $\substack{28880 \\ 28880}$ | -89554-20082 |
| ${ }_{5}$ | 08554-20050 | 1 |  |  | 08554-20050 |
| ${ }^{4} \mathbf{4}$ |  | 1 | (taste | cis |  |
| ${ }^{4} 7$ | -0855460035 | i | cheme |  | - |
| ${ }^{\text {x }}$ |  | $\stackrel{1}{1}$ |  | ( 284880 |  |
|  |  | 1 |  |  |  |
|  | 08554-20071 | 1 | Cable assyperisst lo | ${ }_{2848 \mathrm{c}}$ |  |
| ${ }_{\text {cher }}$ | 08 | 1 |  | 28880 | 08554-60055 |
| ${ }_{1}{ }_{15}$ | 0855460046 | 1 | (incle | 28880 | 08554-60046 |
|  | ( |  |  | $\underset{\substack{\text { 95354 } \\ 95354}}{ }$ | $91-6915-1500-00$ $91-6915-15 c 000$ |
|  |  | $\begin{aligned} & 2 \\ & \frac{2}{3} \\ & 1 \end{aligned}$ |  |  |  |

Table 6-3. Replaceable Parts

| Reference Designation | HP Part Number | Oty | Description | Mfr Code | Mfr Part Number |
| :---: | :---: | :---: | :---: | :---: | :---: |
|  | 08554-00022 $08554-00023$ 08554-00024 08554-00032 08554-00034 08554-20013 0370-0102 0370-0114 0370-0116 08555-00C1 08554-00003 08553-0014 |  | COVER:BOHTOM <br> PANEL:REAR <br> REAR HALL:PC BOX <br> FRONT WALL:PC BOX <br> BRACKET:CIRCULATOR <br> MOUNT:YIG OSCILLATOR <br> KNOB:RED BAR C.125" SHAFT 0.500 NDIA (SCAN <br> (SCAN WIDTH) <br> KNOA:RED W/ARROW 5/8" OD 1/8* SHAFT <br> (fine tune) <br> KNOB: BLACK ROUND(FREQUENCY) <br> DIAL/KNOB(BANDWIDTH) <br> DIAL-KNOB ASSY:SCAN WIDTH <br> dial: Knob assycinput attenuation) | 28480 28860 2886 28880 28880 28860 28480 2886 2880 28480 28480 2880 2880 28480 28480 | 08554-00022 $08554-00023$ $08554-00024$ 08554-00025 08554-00032 <br> 08554-0C034 $08554-20013$ $0370-0102$ <br> 0370-0114 <br> 0370-0116 <br> 08555-00011 <br> 08554-00003 <br> 08553-0014 |



Table 6-3. Replaceable Parts


Table 6-3. Replaceable Parts

| $\begin{aligned} & \text { MFR } \\ & \text { NO. } \end{aligned}$ | manufacturer name | ADDRESS | $\begin{aligned} & \text { 2ID } \\ & \text { CODE } \end{aligned}$ |
| :---: | :---: | :---: | :---: |
| 00000 | U.S.A. COMmon | ANY SUPPLIER OF U.S.A. |  |
| 00853 | SANGAMO Electric co.pickens div. | PICKENS, S.C. | 29671 |
| 01121 | ALLEN RRAOLEY CO. | MILHAUKEE, WIS. | 53204 |
| 02660 | AMPHENOL CORP. | Broadvieh, ILL. | 60153 |
| 02735 | RCA SOLID STATE \& RECEIVING TUBE DIV. | SOMERVILLE, N.d. | 08876 |
| 04009 | ARROW, HART \& HEGEMAN ELECT. CO. | HARTFORD, CONN. | 08106 |
| 04713 | MOTOROLA SEMICONOUC TOR PROD. INC. | PHOENIX, ARIZ. | 85008 |
| 07263 | FAIRCHILO CAMERA E INST. CORP. SEMICONDUCTOR DIV. | MOUNTAIN VIEN, CALIF. | 94040 |
| 08717 | SLIAN CO. The | SUN VALLEY, CALIF. | 91352 |
| 14674 | CORNING GLASS WORKS | CORNING, N.Y. | 14830 |
| 16179 | OMNI-SPECTRA INC. | FARMINGTON, MICH. | 48024 |
| 28480 | HEWLETT-PACKARD CO. CORPORATE HO | Your nearest hp office |  |
| 56289 | SPRAGUE ELECTRIC CO. | N. AOAMS, MASS. | 01247 |
| 70276 | ALLEN MFG. CO. | HARRTFORD, CONN. | 06101 |
| 71041 | BOSTON GEAR WORKS DIV N. AMERICAN ROCKWELL CORP. | QUINCY, MASS. | 02171 |
| 71468 71744 | ITT CANNON ELECT. INC. | LOS ANGELES, CALIF. | 90031 |
| 71744 72136 | CHICAGO MINIATURE LAMP HORKS ELECTRO MOTIVE MFG. CO. INC. | CHICAGO, ILL, | 60640 |
| 72962 | ELASTIC STOP NUT DIY. AMERACE ESNA CORP. | UNION, N.J. | 07083 |
| 72982 | ERIE TECHNOLOGICAL Prod. Inc. | ERIE, PA. | 16512 |
| 73734 | FEDERAL SCREW PROD. INC. | CHICAGO, ILL. | 60618 |
| 78189 | SHAKEPROOF DIV. ILLINOIS TOOL WORKS | ELGIN, ILL. | 60120 |
| 80131 | ELECTRONIC INDUSTRIES ASSOCIATION | WASHINGTON D.C. | 20006 |
| 82110 | GUDEBROD BROS. SILK CO. INC. | PHILADELPHIA, PA. | 19107 |
| 91506 | AUGAT INC. | ATTLEBORO, MASS. | 02703 |
| 95354 | METHODE MFG. CO. | ROLLING MEAOONS, ILL. | 60008 |
| 98291 | SEALECTRO CORP. | MAMARONECK, N.Y. | 10544 |

## SECTION VII MANUAL CHANGES

## 7-1. INTRODUCTION

$7-2$. This section normally contains information for adapting this manual to instruments for which the content does not apply directly. Since this
manual does apply directly to instruments having serial numbers listed on the title page, no change information is given here. Refer to INSTRUMENTS COVERED BY MANUAL in Section I for additional important information about serial number coverage.

# SECTION VIII SERVICE 

## 8-1. INTRODUCTION

8-2. This section provides instructions for troubleshooting and repairing the Hewlett-Packard Model 8554B Spectrum Analyzer RF Section.

## 8-3. THEORY OF OPERATION

8-4. Theory of operation appears on the foldout pages opposite the block diagram on the Service Sheets. The block diagram on Service Sheet 1 is keyed to the remaining service sheets so that the reader may quickly locate the schematic and theory concerning any specific circuit.

## 8-5. RECOIMMENDED TEST EQUIPMENT

8 -6. Test equipment and test equipment accessories required to maintain the RF Section are listed in Tables 1-4 and 1-5. Equipment other than that listed may be used if it meets the listed minimum specifications.

8-7. The HP 11592A Service Kit is an accessory item available from Hewlett-Packard for use in maintaining both the RF and IF Sections of the Spectrum Analyzer. Some maintenance can be performed without this kit by removing the top covers from both the RF Section and the Display Section. However, this procedure exposes dangerous potentials in the Display Section chassis and should not be used unless absolutely necessary.

8-8. All maintenance can and should be performed with the analyzer plug-ins installed on the extender cables provided in the service kit, or using a Display Section cover that has the area over the RF Section cut out. The kit can be obtained by contacting the nearest Hewlett-Packard Sales and Service Office. An extra cover can be obtained from Hewlett-Packard and modified.

## 8-9. TROUBLESHOOTING

8-10. The System Test and Troubleshooting Procedure (Table $8-3$ ) is designed to isolate trouble to the circuit board or assembly level. It should be used in conjunction with the top and bottom internal views on the first fold-out in this manual. It should also be used in conjunction with the block diagram on Service Sheet 1.

8-11. Circuit level troubleshooting and analysis is provided on the foldout page opposite each schematic. After the cause of a trouble has been isolated and corrected, check the troubleshooting information associated with that circuit for any adjustments that may have to be performed.

## 8-12. GENERAL SERVICE INFORMATION

## 8-13. Part Location Aids

8-14. The locations of chassis-mounted parts and major assemblies are shown in Figure 8-5. The locations of individual components mounted on printed circuit boards or other assemblies are shown on the appropriate schematic diagram page or on the page opposite it. The part reference designator is the assembly designator plus the part designator. (Example: A6R9 is R9 on the A6 assembly). For specific component description and ordering information refer to the parts list in Section VI.

## 8-15. Factory Selected Components

8-16. Some component values are selected at the time of final checkout at the factory (see Table 5-2). Usually these values are not extremely critical; they are selected to provide optimum compatibility with associated components. These components are identified on individual schematics by an asterisk (*). The recommended procedure for replacing a factory-selected part is as follows:
a. Try the original value, then perform the calibration test specified for the circuit in the performance and adjustment sections of this manual.
b. If calibration cannot be accomplished, try the typical value shown in the parts list, if different, and repeat the test.
c. If the test results are still not satisfactory, substitute various values within the tolerances specified in Table 5-2 until the desired result is obtained.

## 8-17. Diagram Notes

8-18. Table 8 -2, Schematic Diagram Notes, provides information relative to symbols and


Figure 8-1. 8554B RF Section with Circuit Board Extended
measurements units shown in schematic diagrams.

## 8-19. Servicing Aids on Printed Circuit Boards

$8-20$. The servicing aids include test points, transistor and integrated circuit designations, adjustment callouts and assembly stock numbers.

## 821. REPAIR

## 8-22. Etched Circuits

8-23. The etched circuit boards in the RF Section are of the plated-through type consisting of metallic conductors bonded to both sides of insulating material. The metallic conductors are extended through the component mounting holes by a plating process. Soldering can be done from either side of the board with equally good results. Table 8-1 lists recommendations and precautions pertinent to etched circuit repair work.
a. Avoid unnecessary component substitution; it can result in damage to the circuit board and/or adjacent components.
b. Do not use a high-power soldering iron on etched circuit boards. Excessive heat may lift a conductor or damage the board.
c. Use a suction device (Table 8-1) or wooden toothpick to remove solder from component mounting holes. DO NOT USE A SHARP METAL OBJECT SUCH AS AN AWL OR TWIST DRILL FOR THIS PURPOSE. SHARP OBJECTS MAY DAMAGE THE PLATED-THROUGH CONDUCTOR.
d. After soldering, remove excess flux from the soldered areas and apply a protective coating to prevent contamination and corrosion. See Table 8-1 for recommendation.


Figure 8-2. Examples of Diode and Transistor Marking Methods
Table 8-1. Etched Circuit Soldering Equipment

| ITEM | USE | SPECIFICATION | ITEM RECOMMENDED |
| :---: | :---: | :---: | :---: |
| Soldering tool | Soldering, unsoldering | Wattage rating: 37-50; Tip Temp: 750-800 ${ }^{\circ}$ | Ungar \#766 handle w/*Ungar \#1237 heating unit |
| Soldering Tip | Soldering, unsoldering | *Shape: pointed | *Ungar \# PL111 |
| De-soldering <br> Aid | To remove molten solder from connection | Suction device | Soldapullt by Edsyn Co., Arieta, California |
| Resin (flux) Solvent | Remove excess flux from soldered area before application of protective coating | Must not dissolve etched circuit base board | Freon; Acetone; Lacquer Thinner |
| Solder | Component replacement Circuit board repair Wiring | Resin (flux) core, high tin content (60/40 tin/lead), 18 gauge (SWG) preferred |  |
| Protective Coating | Contamination, corrosion protection | Good electrical insulation, corrosionprevention properties | Silicone Resin such as GE DRI-FILM**88 |
| *For working on circuit Boards: for general purpose work, use Ungar No. 4037 Heating Unit ( $47^{1 / 2}-56{ }^{1 / 2} \mathbf{W}$ ) tip temperature of 850-900 degrees) and Ungar No. PL1 13 1/8' chisel tip. <br> **General Electric Co.. Silicone Products Dept., Waterford, New York, U.S.A. |  |  |  |

## 8-24. Etched Conductor Repair

8-25. A broken or burned section of conductor can be repaired by bridging the damaged section with a length of tinned copper wire. Allow adequate overlay and remove any varnish from etched conductor before soldering wire into place.

## 8-26. Component Replacement

8-27. Remove defective component from board.

## NOTE

Although not recommended, axial lead components, such as resistors and tubular capacitors, can be replaced without unsoldering. Clip leads near body of defective component, remove component and straighten leads left in board. Wrap leads of replacement component one turn around original leads. Solder wrapped connection, and clip off excess lead.

8 -28. If component was unsoldered, remove solder from mounting holes, and position component as orginal was positioned. DO NOT FORCE LEADS INTO MOUNTING HOLES; sharp lead ends may damage plated-through conductor.

8-29. Transistor Replacement. Transistors are packaged in many physical forms. This sometimes results in confusion as to which lead is the collector, which is the emitter, and which is the base. Figure 8-2 shows typical epoxy and metal case transistors and the means of identifying the leads.

8-30. To replace a transistor, proceed as follows:
a. Do not apply excessive heat; see Table 8-1 for recommended soldering tools.
b. If possible, use long-nose pliers between transistor and hot soldering tools.
c. When installing replacement transistor, ensure sufficient lead length to dissipate soldering heat by using about the same length of exposed lead as useful for orginal transistor.
d. Integrated circuit replacement instructions are the same as those for transistors.

8 -31. Some transistors are mounted on heat sinks for good heat dissipation. This requires good thermal contact with mounting surfaces. To assure good thermal contact for a replacement transistor, coat both sides of the insulator with Dow Corning No. 5 silicone compound or equivalent before fastening the transistor to the chassis. Dow Corning

No. 5 compound is available in 8 oz. tubes from Hewlett-Packard; order HP Part No. 8500-0059.

8-32. Diode Replacement. Solid state diodes have many different physical forms. This sometimes results in confusion as to which lead or connection is the cathode (negative) and which lead is the anode (positive), since not all diodes are marked with the standard symbols. Figure $8-2$ shows examples of some diode marking methods. If doubt exists as to polarity, an ohmmeter may be used to determine the proper connection. It is necessary to know the polarity of the ohms lead with respect to the common lead for the ohmmeter used. (For the HP Model 410B Vacuum Tube Voltmeter, the ohms lead is negative with respect to the common; for the HP Model 412A DC Vacuum Tube Voltmeter, the ohms lead is positive with respect to the common). When the ohmmeter indicates the least diode resistance, the cathode of the diode is connected to the ohmmeter lead which is negative with respect ot the other lead.

## NOTE

Replacement instructions are the same as those listed for transistor replacement.

## 8-33. OPERATIONAL AMPLIFIERS

## 8-34. Circuits and Symbols

8-35. Operational amplifiers are widely used as summing amplifiers, offset amplifiers, buffers and level detectors in regulated power supplies. The particular function is determined by external circuit connections.

8-36. Figure 8-3 shows a typical operational amplifier. Circuit $A$ is a non-inverting buffer amplifier with a gain of 1 . Circuit B is a non-inverting amplifier with gain determined by the resistance of R1 and R2. Circuit C is an inverting amplifier with gain determined by R2 and R1. Circuit D shows typical circuit connections and parameters. It is assumed that the amplifier has high gain, low output impedance and high input impedance.

## 8-37. Troubleshooting

8-38. An operational amplifier can be characterized as an ideal voltage amplifier having low output impedance, high input impedance, and very high gain. Also the output voltage is proportional to the difference in the voltages applied to the two input terminals. In use, the amplifier drives the input voltage difference close to zero.

8 -39. When troubleshooting an operational amplifier, measure the voltages at the two inputs with no

## OPERATIONAL AMPLIFIER

A


B


INPUT IMPEDANCE: VERY HIGH
OUTPUT IMPEDANCE: VERY LOW

(1)
$V_{0}=V_{1}\left(1+\frac{R 1}{R 2}\right)-V_{2}\left(\frac{R 1}{R 2}\right)$
(2)
2) IF $\mathrm{V}_{2}=0\left(\frac{1}{\nabla}\right)$, THEN $\mathrm{V}_{0}=\mathrm{V}_{1}\left(1+\frac{\mathrm{R} 1}{\mathrm{R} 2}\right)$
(3)

$$
\text { IF } V_{1}=0(\nabla), \text { THEN } \quad V_{0}=-V_{2}\left(\frac{R 1}{R 2}\right)
$$

## D



IF 'A Ein $^{\prime}$ IS LARGE,$V_{F}=V_{1}$

INPUT IMPEDANCE = R2
OUTPUT IMPEDANCE: VERY LOW

Figure 8-3. Operational Amplifier Equivalent Circuit
signal applied; the difference between these voltages should be less than 10 mV . A difference voltage much greater than 10 mV indicates trouble in the amplifier or its external circuitry. Usually this difference will be several volts and one of the inputs will be very close to an applied circuit operating voltage (for example, $+20 \mathrm{~V},-12 \mathrm{~V}$ ).

8-40. Next, check the amplifier's output voltage. It will probably also be close to one of the applied circuit potentials: ground, $+20 \mathrm{~V},-12 \mathrm{~V}$, etc. Check to see that the output conforms to the inputs. For example, if the inverting input is positive, the output should be negative; if the non-inverting input is
positive, the output should be positive. If the output conforms to the inputs, check the amplifier's external circuitry. If the amplifier's output does not conform to its inputs, it is probably defective.

## 8-41. DIAL CALIBRATION PROCEDURE

$8-42$. To restring the frequency dial, follow the procedure outlined in Figure 8-4. After the dial is restrung, or after the frequency tuning pot R6 is replaced, perform the following adjustments:
a. Turn FREQUENCY full counterclockwise. The dial pointer should indicate $3 / 4$ to $11 / 4$ small divisions to the left of 0 MHz .
b. Turn FREQUENCY full clockwise. The dial pointer should indicate at least $3 / 4$ of a small difision to the right of 1250 MHz .
c. If necessary, loosen the set screws on the gear shaft of the FREQUENCY pot and re-position
the gear slightly by turning the FREQUENCY knob while the pot is at either stop. Then retighten the set screws and repeat steps $a$ and $b$.
e. Perform the frequency calibration adjustments specified in Section V.


1. Remove top cover.
a. Tune to low end of scale.
b. Remove front panel assembly from side panels.
c. Remove scale assembly.
d. Remove tuning knobs.
e. Remove 2 screws which hold gearbox to panel assembly.
f. Remove left pully at left end of pointer slot.
2. To replace string on RIGHT side of pointer:
a. Remove pointer from slot, detach old string.
b. Access to fixed end of string is through the hole in the front gearbox plate. Line up dial drum with this hole so that old string may be withdrawn.
c. Pass a new piece of dial string (about $15^{1 / 2 ")}$ through the hole and double knot the fixed end. Clip off excess string and draw the knot into the hole.
d. Reset the tuning shaft fully ccw.
e. Pass the free end of the string into the right end of the pointer slot. Tie it to the pointer spring where it is attached to the pointer.
f. Replace pointer in slot.
g. Replace gearbox screws.
h. Turn shaft fully cw .
i. Loosen fixing screw at opposite end of string and adjust string tension so that pointer is stretched $3 / 16^{\prime \prime}$ when string is on pulleys.
j. Reassemble, using reverse of procedure in 1.
3. To replace string on LEFT side of pointer:
a. Remove pointer from slot and remove old string.
b. Tie approximately 12 " of dial string (use double knot) to the pointer spring and replace pointer in slot.
c. Replace gearbox screws.
d. Turn shaft fully cw .
e. Place dial string on pulleys.
f. Wrap string around dial drum, and tie under screwhead, while maintaining about $3 / 16$ " stretch on pointer spring.
g. Reassemble, using reverse of procedure in 1.
4. Check calibration; adjust by moving the 29 tooth gear on the tuning pot shaft. Perform dial calibration procedure.

## SCHEMATIC DIAGRAM NOTES

R,L,C Resistance is in ohms, inductance is in microhenries, capacitance is in picofarads unless otherwise noted.
P/O Part Of.
Asterisk denotes a factory-selected value. Value shown is typical. Capacitors may be omitted or resistors jumpered.
O
Panel control.

* Screwdriver adjustment.


Encloses front panel designations.
$\qquad$ Circuit assembly borderline.

-     -         -             -                 - Other assembly borderline.

Heavy line with arrows indicates path and direction of main signal.
Heavy dashed line with arrows indicates path and direction of main feedback.

Wiper moves toward CW with clockwise rotation of control as viewed from shaft or knob.

Numbers in stars on circuit assemblies show locations of test points.
Lettered test point; no measurement provided.
Encloses wire color code. Code used (MIL-STD-681) is the same as the resistor color code. First number identifies the base color, second number the wider stripe, and the third number identifies the narrower stripe; e.g. (947) denotes white base, yellow wide stripe, violet narrow stripe.

Tunable resonator (or cavity).
Loop coupling to coaxial path.
E-plane coupling by aperture ( H -plane coupling indicated by H in symbol): $\mathrm{E}(\mathrm{H})$ indicates that the physical plane of the aperture is perpendicular to the transverse component of the major $E(H)$ lines.

Circulator (isolator): arrowhead indicates direction of power flow is from any arm to next adjacent arm but not to any other arm. Power does not flow between arms separated by resistor symbol.

Short circuit. (Not a fault).

Letter = off page connection.
Number = Service Sheet number for off-page connection.

1 Block numbers reference between text and schematic.

Table 8-2. Schematic Diagram Notes (2 of 3)

## SWITCH DESIGNATIONS



Table 8-2. Schematic Diagram Notes (3 of 3)


Table 8-3. System Test and Troubleshooting Procedure

| TEST | FAULT | PROCEDURE |
| :---: | :---: | :---: |
| 1. Set POWER switch to ON. Power lamp on, fan operates. Proceed to test 2. | Light not on and/or fan inoperative. | Check Display Section. |
| 2. Rotate INPUT ATTENUATION control and observe LOG REF LEVEL index lights. <br> Lights operate properly. Proceed to test 3. | Some, but not all lights illuminate. <br> Some, but not all lights illuminate. | Check the -12.6 volt supply from Display Section. If voltage is present see Service Sheet 4. If voltage is not present, check the Display Section power supply. <br> Check light bulbs and see Service Sheet 4. |
| 3. Set Analyzer controls as follows: SCAN TIME <br> PER DIVISION . . . 5 ms SCAN MODE . . . . . . . INT SCAN TRIGGER . . . AUTO and observe SCANNING light. <br> Light operates normally. Proceed to test 4. | SCANNING light does not illuminate. | See System Test and Troubleshooting Procedure in 8552 Operating and Service Manual. Check power supply circuits. |
| 4. Adjust Display Section for a baseline trace. <br> Baseline trace is normal. Proceed to test 5. | Trace does not appear. | See System Test and Troubleshooting Procedure in 8552 Operating and Service Manual. Check scan amplifier, scan generator and horizontal deflection amplifier. |
| 5. Set analyzer controls as follows: <br> SCAN WIDTH . . $0-1250 \mathrm{MHz}$ SCAN TIME <br> PER DIVISION . . . 2 ms LOG REF LEVEL . . -10 dBm <br> Connect CAL OUTPUT to RF INPUT. | Signal does not appear on Display Section CRT. | See System Test and Troubleshooting Procedure in 8552 Operating and Service Manual. Check calibration oscillator. If calibration oscillator is operating properly, go to test 6. |
| 6. Set analyzer controls as follows: BANDWIDTH . . . . . 10 kHz FINE TUNE . . . . . Centered SCAN WIDTH . PER DIVISION SCAN WIDTH <br> PER DIVISION.$\dot{20}$ 0 kHz 0 dB INPUT ATTENUATION $\quad$ OUB TUNING STABILIZER BASELINE CLIPPER . . . ccw LOG REF LEVEL |  |  |

Table 8-3. System Test and Troubleshooting Procedure (cont'd)

| TEST | FAULT | PROCEDURE |
| :---: | :---: | :---: |
| Connect a $50 \mathrm{MHz}-33 \mathrm{dBm}$ signal from the signal generator to the W7 jack (located between green coaxial cable W7A and clear coaxial cable W7B) using the 11592-60001 cable. Tune the signal generator slightly around 50 MHz until the signal is centered on the CRT display. With the AMPL CAL centered the signal should read $-30 \mathrm{dBm} \pm 2 \mathrm{~dB}$. If signal is correct, reconnect W7A and proceed to test 7. | Signal incorrect or missing. | See System Test and Troubleshooting Procedure in 8552 Operating and Service Manual. |
| In steps $7 a$ and $7 e$ it is necessary to simulate the input impedance of the circuit following the point tested to insure accuracy of the meter readings. Use the HP 11563A 50 ohm tee, two HP 1250-0780 BNC jack to type N Plug Adapters, the HP 11593A termination, the HP 11592-60001 subminiature to BNC cable, and the HP 1250-0827 Jack-to-jack adapter. |  |  |
| 7. Perform following sub-tests until a malfunction has been found and corrected, then repeat test 5 . |  |  |
| 7a. Set signal generator for 400 MHz at -10 dBm and connect to RF INPUT through a BNC tee; also connect Channel A input of tunable RF voltmeter to BNC tee. Connect Channel B input of voltmeter to output of AT1, 3 dB pad (use 50 ohm tee and adapter cable). <br> Set INPUT ATTENUATION to 10 dB. Adjust signal generator for -10 dBm on Channel A of voltmeter. Channel B should read about - 14.5 dBm . Rotate INPUT ATTENUATION and check for correct attenuation. <br> Reconnect AT1 and proceed to test 7 b . | Signal is missing or level is incorrect. | See Service Sheet 2. Check A3, FL1, and AT1. |
| 7b. Remove the output connector from the notch filter FL2 and connect the tunable RF Voltmeter Channel A input to the notch filter using a 50 ohm dummy load. With the signal generator connected as in 7a and the analyzer tuned for maximum (in ZERO scan) the tunable RF Voltmeter should indicate about -28.5 dBm . If correct reading is obtained, proceed to test 7 f . | Signal is missing. | Proceed to test 7c. |

Table 8-3. System Test and Troubleshooting Procedure (cont'd)

| TEST | FAULT | PROCEDURE |
| :--- | :--- | :--- |
| $\begin{array}{l}\text { 7c. Connect the frequency counter to } \\ \text { FIRST LO OUTPUT on front panel. } \\ \text { With the analyzer operating in ZERO } \\ \text { scan mode the first LO output should } \\ \text { be 2050 MHz above the frequency in- } \\ \text { dicated on the CENTER FREQUENCY } \\ \text { scale. Check at 100, 600 and 900 MHz. } \\ \text { If correct readings are observed, pro- } \\ \text { ceed to test 7d. }\end{array}$ | $\begin{array}{l}\text { Signal is missing } \\ \text { or incorrect. }\end{array}$ | $\begin{array}{l}\text { Check the input voltages to A7 (the } \\ \text { YIG oscillator assembly). If proper dc } \\ \text { levels are not present check A6 (the }\end{array}$ |
| YIG power supply) and the FRE- |  |  |\(\left.] \begin{array}{l}QUENCY control circuits. If proper <br>

dc levels are present at the A7 assem- <br>
bly, repeat the test with the counter <br>
connected directly to the output of <br>
the YIG oscillator. If correct readings <br>
are obtained, replace circulator A13. <br>
If not, replace the A7 assembly and <br>
repeat 7b. See Service Sheet 4.\end{array}\right\}\)

Table 8-3. System Test and Troubleshooting Procedure (cont'd)

| TEST | FAULT | PROCEDURE |
| :---: | :---: | :---: |
| 8. Set analyzer controls and connections same as test 5 . Rotate BASE LINE CLIPPER to 10 o'clock. The display should be similar to that shown in the Procedure column. Return BASE LINE CLIPPER to full ccw. <br> Vary VERTICAL POSITION control to center baseline trace on bottom CRT graticule. <br> Signal amplitude is unimportant in this test. Proceed to test 9 . | Sweep does not extend to full width of graticule. <br> Signals not all present or improperly spaced. <br> Baseline trace does not vary. | See System Test and Troubleshooting Procedure in 8552 Operating and Service Manual. Check Scan Generator and Deflection Amplifier assy's. <br> Same as above. Also refer to Service Sheet 3. YIG power supply may be defective. <br> See System Test and Troubleshooting Procedure in 8552 Operating and Service Manual. Check vertical deflection circuit. |
| 9. Set LOG REF LEVEL maximum ccw. Set SCAN TIME PER DIVISION to 10 SECONDS and adjust focus and astigmatism. Adjust trace align to center trace on bottom CRT graticule. Return SCAN TIME to $3 \mathrm{~ms} / \mathrm{div}$. Proceed to test 10 . | Focus and astigmatism inoperative or trace will not align. | Refer to Display Section Manual and repair as required. |
| 10. Turn the FREQUENCY control and observe the marker. Marker should move as FREQUENCY is tuned. Proceed to test 11. | Marker is missing. | Refer to Display Section Manual and repair as required. |
| 11. Tune the FREQUENCY control to move the marker exactly under the 30 MHz signal. The signal will null when the marker is tuned to the exact frequency of the signal. <br> Set analyzer controls as follows: BANDWIDTH . . . . . . . 10 kHz SCAN WIDTH . . . PER DIVISION SCAN WIDTH PER DIVISION 2 MHz LOG REF LEVEL . . . . -10 dBm <br> Center 30 MHz signal on the CRT. Signal should be similar to that shown in the procedure column. Proceed to test 12 . | 30 MHz signal does not appear on Display CRT. | Check calibration and alignment of the analyzer. |

Table 8-3. System Test and Troubleshooting Procedure (cont'd)

| TEST | FAULT | PROCEDURE |
| :--- | :--- | :--- |
| 12. Reduce SCAN WIDTH PER <br> DIVISION to 10 kHz and recenter the <br> display with the FINE TUNE control. <br> Signal should be similar to that shown <br> in the Procedure column. Proceed to <br> test 13. | Signal is unstable. <br> FINE TUNE does <br> not, <br> position. | Refer to Service Sheets 6 and 7 and <br> repair APC or reference signal circuits. <br> Refer to Service Sheet 5 and check <br> third LO circuit and third LO drive <br> circuits. |

Table 8-4. Assembly and Component Locations

| ASSEMBLY |  | SCHEMATIC | PHOTO |
| :---: | :---: | :---: | :---: |
| A1 | Bandwidth Switch | Service Sheet 10, 11 | Figure 8-37, 8-40 |
| A2 | Scan Width Switch | Service Sheet 4, 7, 8, 10, 11 | Figure 8-33, 8-41 |
| A3 | Input Attenuation Switch | Service Sheet 2 | Figure 8-8 |
| A4 | Search Loop and Phase Lock Memory Amplifier | Service Sheet 7 | Figure 8-31 |
| A5 | Marker Generator | Service Sheet 9 | Figure 8-35 |
| A6 | YIG Power Supply | Service Sheet 4 | Figure 8-16 |
| A7 | YIG Oscillator (First LO) | Service Sheet 4 | Figure 8-5 |
| A8 | First and Second Converter | Service Sheet 3 | Figure 8-10 through 8-12 |
| A9 | Third Converter | Service Sheet 5 | Figure 8-18 through 8-23 |
|  | Reference Oscillator and Phase Lock Amplifier | Service Sheet 6 | Figure 8-27 |
| A11 | Not Assigned |  |  |
| A12 | 50 MHz Amplifier | Service Sheet 5 | Figure 8-24 |
| A13 | Circulator | Service Sheet 4 | Figure 8-5 |
|  | COMPONENT | SCHEMATIC | PHOTO |
| C1 <br> C2 <br> C3 <br> DS1 <br> FL1 <br> FL2 <br> J1 <br> J2 <br> J3 <br> L1 <br> P1 <br> P2 <br> P3 <br> R1, 2 <br> R3 | DISPLAY UNCAL <br> RF INPUT <br> FIRST LO OUTPUT <br> THIRD LO OUTPUT <br> Not assigned <br> Not assigned <br> FINE TUNE | Service Sheet 2 <br> Service Sheet 5 <br> Service Sheet 3 <br> Service Sheet 10 <br> Service Sheet 2 <br> Service Sheet 3 <br> Service Sheet 2 <br> Service Sheet 6 <br> Service Sheet 5 <br> Service Sheet 4 <br> Service Sheet 9,10 <br> Service Sheet $2,5,8,9,10,11$ <br> Service Sheet 5 | Figure 8-8 <br> Figure 8-5 <br> Figure 8-11 <br> Figure 8-5 <br> Figure 8-5 <br> Figure 8-5 <br> Figures $8-5 \& 8-8$ <br> Figure 8-5 <br> Figure 8-5 <br> Figure 8-5 <br> Figure 8-5 <br> Figure 8-5 <br> Figure 8-5 |

Table 8-4. Assembly and Component Locations (cont'd)

| COMPONENT | SCHEMATIC | PHOTO |
| :---: | :---: | :---: |
| R4 AMPL CAL | Service Sheet 10 | Figure 8-5 |
| R5 Not Assigned |  |  |
| R6 FREQUENCY | Service Sheet 4 | Figure 8-5 |
| R7, $8 \quad$ Not Assigned |  |  |
| R9 | Service Sheet 4 | Figure 8-5 |
| R10, 11 | Service Sheet 3 | Figure 8-11 |
| S1 TUNING STABILIZER | Service Sheet 7 | Figure 8-5 |
| W1 | Service Sheet 2 | Figure 8-5 \& 8-8 |
| W2 | Service Sheet 2 | Figure 8-5 |
| W3 | Service Sheet 2, 3 | Figure 8-5 |
| W4 | Service Sheet 3, 4 | Figure 8-5 |
| W5 | Service Sheet 3,5 | Figure 8-5 |
| W6 | Service Sheet 4 | Figure 8-5 |
| W7 | Service Sheet 5 | Figure 8-5 |
| W8 | Service Sheet 4, 6 | Figure 8-5 |
| W9 | Service Sheet 6 | Figure 8-5 |
| W10, 11 | Service Sheet 4 | Figure 8-5 |
| W12, 13, 14 | Service Sheet 5 | Figure 8-5 |
| W15 | Service Sheet 4, 8 | Figure 8-5 |
| AT1 | Service Sheet 2 | Figure 8-5 |
| AT2 | Service Sheet 6 | Figure 8-5 |
| AT3 | Service Sheet 5 | Figure 8-5 |



INTERNAL VIEWS



Figure 8.5. Assembly and Adjustment Locolions


Figure 8-6. Simplified Analyzer Block Diagram

## SERVICE SHEET 1

## SIMIPLIFIED ANALYZER BLOCK DIAGRAM

The Spectrum Analyzer is basically a superheterodyne receiver which may be manually tuned to a fixed frequency or swept tuned through a selected span of frequencies. The "receiver" output is applied to a calibrated CRT resulting in a visual amplitude-vs-frequency display.

A simplified block diagram of the analyzer is illustrated in Figure 8-6. The sawtooth generator provides the time base for both CRT horizontal deflection and receiver local oscillator tuning. Shaping circuits modify the sawtooth ramp, which tunes the local oscillators, to ensure that frequencies displayed on the CRT are separated linearly with respect to the time base of the CRT display.

The detected output of the receiver is applied to the vertical deflection plates. When a signal is received, a vertical deflection proportional to the amplitude of the signal is displayed.

The analyzer utilizes quadruple conversion and the block in Figure 8-6 shows the local oscillators and mixers and the required IF amplifiers. The final stages contain selectable bandpass filters which determine the analyzer bandwidth.

Before detection, the signal is amplified in circuitry which provides (at operator option) a display which is either linearly or logarithmically proportional in amplitude to the input RF signal. The signal is detected and amplified to drive the vertical plates of the CRT. The vertical deflection of the CRT beam is thus proportional to the input signal level. Since the horizontal plates of the CRT are driven by the same voltage that drives the local oscillators, the horizontal deflection of the CRT is proportional to frequency.

## RF SECTION BLOCK DIAGRAM

## Input Circuits (Service Sheet 2)

The RF input is coupled into the analyzer through a capacitor to protect diodes in the first mixer from damage when signals containing dc components are to be analyzed. Signals containing dc components as high as 50 volts may be directly coupled to the analyzer without harming the first mixer diodes.

The input attenuator provides 0 to 50 dB of attenuation to the input RF signal to expand the signal handling capability of the analyzer. Ganged with the attenuator control, but not a part of the attenuation circuit, are wafers to provide power to index lamps adjacent to the LOG REF LEVEL control (in the IF Section) and to aid in programming the step gain of the linear scale factor amplifier (in the IF Section) when the analyzer is operated in the LINEAR mode.

The Low Pass Filter provides FL1 rejection to signals above 1500 MHz to prevent responses from signals outside the passband of the analyzer.

The 3 dB attenuator provides isolation to optimize flatness of the passband from 100 kHz to 1250 MHz .

## First and Second Converter (Service Sheet 3)

The first mixer is driven by the output from the first local oscillator (a YIG controlled oscillator) which may be swept tuned from 2050 to 3300 MHz , portions of this range, or fixed tuned. The output of the first mixer ( 2050 MHz ) is applied to the second mixer through a three-cavity 2050 MHz filter and a thin-film low pass filter.

The second mixer is driven by two signals: a fixed second local oscillator frequency of 1500 MHz and the first IF frequency of 2050 MHz . The output of the second mixer is an IF frequency of 550 MHz which is applied through a notch

## SERVICE SHEET 1 (cont'd)

filter (which offers maximum rejection to the 1500 MHz second local oscillator frequency) to the third mixer and IF amplifier assembly.
First LO and Control Circuits (Service Sheet 4)
A YIG sphere is a resonant element whose resonant frequency changes with a change in the magnetic field which surrounds it. In the 8554 B the magnetic field surrounding the YIG is closely controlled to provide a swept tuning range of 2050 MHz to 3300 MHz , any portion of this frequency range (down to 5 MHz segments) or a fixed frequency.

The output of the YIG oscillator-amplifier is applied to the first mixer through a circulator. The circulator also routes reflected power from the first mixer to a sampler for use when the analyzer is operated in stabilized (phase locked) modes.

The YIG frequency is directly controlled by the output of the tune amplifier circuit which is driven by the scan ramp, selected portions of the scan ramp or a dc level from the FREQUENCY control. The tune amplifier also provides an output for use in the marker generator when the analyzer is operated in the 0 1250 MHz scan mode.

The sweep amplifier shapes the scan ramp to provide the proper ramp to drive the tune amplifier. It also processes the phase lock error signal when the analyzer is operated in the stabilized mode.

## Third Converter (Service Sheet 5)

The 550 MHz IF signal is amplified approximately 11 dB before being applied through the 550 MHz bandpass filter to the third mixer.

The third mixer is a double balanced mixer using a matched diode quad for mixing.
The third local oscillator is a voltage controlled oscillator which is swept in narrow-scan stabilized modes. When the analyzer is operated in wide scan modes ( $0.5 \mathrm{MHz} / \mathrm{DIV}$ or more) the third local oscillator frequency is fixed at a nominal 500 MHz , determined by the $\pm 500 \mathrm{kHz}$ range of the FINE TUNE control. When the analyzer is operated in narrow-scan mc ? ( $200 \mathrm{kHz} / \mathrm{DIV}$ or less) the third local oscillator is swept five times the SCAN WIDTH PER DIVISION setting on both sides of the center frequency; the center frequency is determined by the amount of first local oscillator shift required to achieve phase lock, and the setting of the FINE TUNE control. As an example, if the analyzer is operated at $200 \mathrm{kHz} / \mathrm{DIV}$ in stabilized mode and the first local oscillator is down shifted 100 kHz to achieve phase lock, the third local oscillator will sweep 2 MHz centered at 499.9 MHz . The offset in third local oscillator frequency is required to retain display accuracy.

The third local oscillator driver amplifier combines the FINE TUNE voltage, the attenuated scan ramp and the offset voltage to control operation of the third local oscillator.

## Phase Lock Reference and Sampler (Service Sheet 6)

When the analyzer is operated in the stabilized (phase locked) mode, the first local oscillator is locked to a stable reference.

In the 8554 B the required reference signal is provided by a crystal controlled 1 MHz pulse generator. The 1 MHz pulse is used to enable a gate in the 2 to 3.3 GHz sampler.

In the 2 to 3.3 GHz sampler the first local oscillator frequency is sampled and the sampler output is used in the search loop and phase lock circuits to find a point in the first local oscillator range that is harmonically related to the 1 MHz reference oscillator. Once a phase lock point is found, the sampler output signal is used to maintain phase lock.

The output of the lock amplifier is used to maintain the first local oscillator phase lock and is also processed to provide an offset to the third local oscillator to compenstate for the initial frequency shift required to attain first local oscillator phase lock.

## SERVICE SHEET 1 (cont'd)

## Phase Lock Memory (Service Sheet 7)

In order to phase lock the first local oscillator a frequency must be found in which the 1 MHz sample pulses and negative-going positive half cycles of the first local oscillator are in time coincidence.

When the same point on the negative-going slope of the first local oscillator positive half cycles is sampled each time the sampler gate is opened, the system is phase locked and the first local oscillator frequency is fixed.

When consecutive sample gates sample the negative-going slope of the first local oscillator positive cycles at a different point on the negative-going slopes, the level of the signals differ and an error voltage is generated to shift the first local oscillator to a phase lock point.

While the first local oscillator is being tuned by the search loop to find a phase lock point, the phase lock offset memory amplifier is tracking the search scan and error signal. Approximately 0.5 second after a phase lock mode has been initiated a relay is energized to remove the output of the phase lock memory amplifier from the offset amplifier. The signal level present at the time the relay contacts open is stored in a capacitor which is between the base of the offset amplifier, an FET amplifier (source follower), and ground. Since the input impedance of a FET is very high the memory capacitor cannot discharge and the stored dc level in the capacitor maintains the FET conduction at the level present when the relay contacts opened. The output from the offset amplifier is used to shift the third local oscillator, in frequency, by an amount equal to the frequency shift required to phase lock the first local oscillator. This prevents signal shift on the CRT display when phase lock is enabled. The scan width and tuning stabilizer switches are used to disable the phase lock capabilities of the analyzer when not selected or in wide scan width modes.

## Scan Width Attenuator (Service Sheet 8)

This portion of the scan width attenuator assembly contains the resistive network and switching required to attenuate the scan ramp for the various SCAN WIDTH PER DIVISION modes.

## Marker Generator (Service Sheet 9)

When the analyzer is operated in the 0 to 1250 MHz mode, the marker generator compares the scan ramp and the dc level from the FREQUENCY control to provide an inverted marker on the display section CRT which represents the center frequency to which the analyzer is tuned (in ZERO and PER DIVISION scan width modes).

The scan width switch disables the marker generator when the analyzer is operated in PER DIVISION or ZERO scan modes.

## IF Section Control Circuits (Service Sheet 10)

The BANDWIDTH switch and the AMPL CAL adjustment both control circuitry in the IF Section.

## Analogic (Service Sheet 11)

The DISPLAY UNCAL lamp is lit whenever SCAN WIDTH, BANDWIDTH, SCAN TIME PER DIVISION, and VIDEO FILTER are set at any combination of positions which does not permit accurate calibration of the analyzer. The switches all have wafers devoted to the analogic circuitry that controls the lamp. The analogic portion of SCAN WIDTH and BANDWIDTH is shown.

## SERVICE

SHEET 1



Figure 8-7. Block Diagram

## SERVICE SHEET 2

It is assumed that one or more of the following conditions exists:

1. The steps specified in Table 4-1, Front Panel Checks, have been performed; that only the marker and first local oscillator signals appear on the Display Section CRT and that there is no RF input to the first mixer with a RF input signal applied to J1 RF INPUT. (Follow steps 1,2 and 3 .)
2. The index lights do not function properly. (Follow step 4 .)
3. Sensitivity as displayed on the CRT in LINEAR mode is incorrect. (Follow step 5 .)

## TROUBLESHOOTING PROCEDURE

Since there are no active components in the circuits to be tested, the 8554B should be disconnected from the IF and Display Sections.

Following the procedures under individual circuit descriptions should aid in isolating the defective component or circuit.

All RF tests are conducted with a $400 \mathrm{MHz},-10 \mathrm{dBm}$ signal from the HP 608 F connected to the analyzer RF INPUT.

EOUIPMENT REQUIRED


## CONTROL SETTINGS

As specified in individual tests.

## 1 INPUT CAPACITOR C1 AND ATTENUATOR ASSEMBLY A3

The input coupling capacitor C1 protects the diodes in the first mixer when signals are applied which contain dc components. The breakdown voltage of the capacitor is in excess of 50 volts. The response of C 1 is essentially flat over the rated passband of the analyzer.

## NOTE

C 1 is located inside of RF INPUT jack J1. To gain access to C1, see J1 disassembly procedure below.
The input attenuator provides 0 to 50 dB of attenuation using five fixed, pad type attenuators. The input attenuator is used to attenuate large input signals to provide better display resolution and minimize distortion products generated in the first mixer. The input attenuator may also be used to identify spurious responses; increasing the input attenuation by 10 dB causes a 10 dB reduction on the display for the input signal, but a much greater reduction in any spurious signals. The flat frequency response and attenuator accuracy contribute to the analyzer's absolute amplitude calibration.

NOTE
Do NOT attempt to disassemble the A3A1 Attenuator. It is NOT field repairable.

TEST PROCEDURE 1
Before removing the analyzer from the Display Section verify failure by using a different signal source.

## SERVICE SHEET 2 (cont'd)

Disconnect the attenuator output W2 and connect it to the HP 8405A using the HP 1250-0827 jack-to-jack adapter, the HP 11592-60001 cable, the UG 274 B/U BNC Tee and the HP 11593A dummy load. With the HP 608 F output ( 400 MHz , -10 dBm ) connected to the analyzer RF INPUT, rotate the INPUT ATTENUATION control through its range. With the INPUT ATTENUATION control set at 0 through 50 dB , the HP 8405A should indicate -10 through -60 dBm respectively. If meter readings are not as specified the input capacitor or attenuator assembly are defective. Reconnect W2.

## 2 1500 MHz LOW PASS FILTER FL1

The 1500 MHz low pass filter FL1, an encapsulated thin-film microcircuit, is relatively flat throughout the passband of the analyzer. The cutoff frequency is 1500 MHz ; the filter is down 3 dB at 1500 MHz and provides more than 70 dB rejection to frequencies between 2.05 and 12 GHz .

## TEST PROCEDURE 2

Connect the HP 8405A to the output of the low pass filter FL1 with the accessories used in step 1 . With the HP 608F connected as in step 1 and the INPUT ATTENUATION control set to 0 dB the HP 8405A should indicate approximately -10.5 dBm . If the correct reading is not obtained the filter is defective.

## 33 dB ATTENUATOR AT1

The 3 dB attenuator provides impedance matching between the low pass filter and the first mixer.

## TEST PROCEDURE 3

Connect the HP 8405A to the output of the 3 dB attenuator with the accessories listed in step 1 . With the HP 608F connected as in step 1 and the INPUT ATTENUATION control set to 0 dB the HP 8405A should indicate approximately -13.5 dBm . If the correct reading is not obtained the 3 dB attenuator is defective.

## 4 INDEX LIGHT SELECTOR SWITCH A3S1-1F

Index light selection switch S1-F1 on the INPUT ATTENUATION control, selects the index lamps associated with the LOG REF LEVEL/LINEAR SENSITIVITY control in the analyzer IF Section. In LOG mode, the selected index lamp is opposite the scale factor on the LOG REF LEVEL control that corresponds to full-scale deflection on the display. In LINEAR mode, the selected index lamp is opposite the LINEAR SENSITIVITY volts per division scale factor. The lamps provide a moveable index point, positioned by the INPUT ATTENUATION control, thus the analyzer's amplitude calibration is maintained for any INPUT ATTENUATION control setting.

## TEST PROCEDURE 4

Connect one lead of the HP 412A to the -12.6 volt source. Rotate the INPUT ATTENUATION control and check for a reading of approximately 91 ohms at connector pins of P3 as follows: 0 dB - pin $33,10 \mathrm{~dB}$ - pin $34,20 \mathrm{~dB}$ - pin 35 , $30 \mathrm{~dB}-\operatorname{pin} 9,40 \mathrm{~dB}-\operatorname{pin} 10,50 \mathrm{~dB}-$ pin 11. If the 91 ohm reading is not obtained at any setting, A3R1 may be defective. If the reading is obtained at some, but not all positions, switch A3S1 or wiring is probably defective.

## SERVICE SHEET 2 (cont'd)

## 5 LINEAR AMPLIFIER COMPENSATION SELECTOR A3S1-1R

S1-1R is a part of an amplifier compensation programming circuit for 10 dB steps of INPUT ATTENUATION control when the analyzer is operated in the LINEAR mode. Refer to the IF Section Operating and Service Manual for a detailed circuit description of the log amplifier.

## TEST PROCEDURE 5

Connect one lead of the HP 412A to the -12.6 volt source. Connect the other lead to P3 pin 8. Meter should indicate continuity on the even-numbered positions of the INPUT ATTENUATION control. Move connection from P3 pin 8 to P3 pin 7. Meter should indicate continuity on the odd-numbered positions of the INPUT ATTENUATION control. If readings are correct, trouble should be in the IF Section. If indications are incorrect, check $\mathrm{S} 1-1 \mathrm{R}$ and 8554 B wiring.

## RF INPUT JACK J1 DISASSEMBLY

1. Disconnect semi-rigid coaxial cable W1 (J1MP11) from the A3 Assembly. Remove J1 by removing nut behind front panel. Remove connector body (J1MP1).
2. Slide the shell, J1MP10, back onto cable J1MP11 (do not bend the cable more than necessary).
3. If necessary, unsolder inner-conductor J1MP4 from blocking capacitor board J1MP7. Remove J1MP2, MP3, and MP4 and disassemble.
4. If necessary, unsolder C1 (J1MP8) from J1MP7. Correct orientation for J1MP8 and MP7 is shown in Figure 8-8.
5. If necessary, unsolder W1 (J1MP11) from J1MP7 and remove.
6. If necessary, unsolder J1MP7 from J1MP6 by removing solder from two holes in bottom of J1MP6.
7. Reassemble J1 by reversing the procedures in steps 1 through 6.


Figure 8-8. RF INPUT Jack J1, Exploded View



Figure 8-9. Input Circuits: A 3 Schematic Diagram

## SERVICE SHEET 3

It is assumed that the input circuit including the attenuators and filter have been proved to be operational and that there is no 550 MHz IF output from the notch filter.

## TROUBLESHOOTING PROCEDURE

When the source of trouble is suspected to be in the first and second converter assembly A8, follow the test procedures listed after individual circuit descriptions until the malfunction is corrected.

## EQUIPMENT REQUIRED



## CONTROL SETTINGS

As specified in individual tests.
FIRST AND SECOND CONVERTER ASSEMBLY (General)
The first and second converter assembly consists of a casting containing four cavities, the first and second mixers, a thin-film low pass filter and the second local oscillator.

## 1 FIRST CONVERTER SUBASSEMBLY

The first converter sub-assembly is part of a replaceable cover plate on the first and second converter assembly. The sub-assembly consists of input connectors (from the first local oscillator and the input RF circuit), a matched diode pair and a stripline circuit. The first converter mixes the first local oscillator signal and the input RF to produce an up-converted IF frequency of 2050 MHz . Conversion loss in the first mixer is typically 10 dB .

## TEST PROCEDURE 1

Disconnect the first local oscillator output from the first mixer. Use the HP $1250-0827$, the HP 1250-0832 the UG $274 \mathrm{~B} / \mathrm{U}$, the HP 11593A and a BNC to BNC cable to connect the first local oscillator output to the HP $5245 \mathrm{~L} / 5254 \mathrm{C}$. The counter should indicate approximately 2050 MHz above the frequency indicated on the slide-rule frequency dial (check at 200,600 and 900 MHz ). If the correct reading is not obtained, see Service Sheet 4.

If the first local oscillator signal is present, reconnect the mixer input and apply a $400 \mathrm{MHz}, 0 \mathrm{dBm}$ signal from the HP 608 F to the analyzer RF INPUT with INPUT ATTENUATION set at 0 dB . Remove one of the small screws next to the first cavity tuning slug (CAUTION - do NOT remove both screws) and couple the $5245 \mathrm{~L} / 5254 \mathrm{C}$ to the first cavity with a stub probe made of $\# 22$ AWG insulated wire. With the analyzer in ZERO scan mode, tune the analyzer and the 5254 C for a reading on the 5245 L meter at the first IF frequency of 2050 MHz . The $5245 \mathrm{~L} / 5254 \mathrm{C}$ should indicate $2050 \mathrm{MHz} \pm 5 \mathrm{MHz}$. If the correct signal is present

## SERVICE SHEET 3 (cont'd)

proceed to test 3 . If the signal is not present first verify presence of the YIG signal (see Service Sheet 4),then repair or replace the first mixer subassembly.

## 2 BANDPASS AND LOW PASS FILTERS

The output of the first mixer is coupled through two tuned bandpass cavities to the low pass filter.

The low pass filter A8A2 couples the signal from the second tuned cavity to the third tuned cavity. The filter is a thin-film device with a cutoff frequency of 5 GHz ; it provides 70 dB rejection from 5.0 through 20 GHz .

## TEST PROCEDURE 2

The low-pass filter is tested during test 3

## 3 SECOND OSCILLATOR AND MIXER

The second mixer is a single diode (CR1) located between the third and fourth tuned cavities. The 550 MHz output of the mixer is coupled through the top of the casting to the 1500 MHz notch (rejection) filter. The second local oscillator is a transistor pair, physically located in the wall of the fourth tuned cavity, which oscillates at a fixed frequency of 1500 MHz .

## TEST PROCEDURE 3

Remove two of the three screws from the plate which covers the 550 MHz output from the second mixer and turn the cover to expose the output coupling.

With the 5254 C set for 1.5 GHz connect the $5245 \mathrm{~L} / 5254 \mathrm{C}$ to the output of the second mixer. The counter should indicate $1500 \mathrm{MHz} \pm 100 \mathrm{kHz}$. If the correct reading is obtained, proceed to the next paragraph. If not, first verify the presence of +10 and -10 volts to the feedthru capacitors at back end of assembly, then remove casting and casting bottom cover to test components. If the diode is defective, replace it and repeat the test. The oscillator components are available as a preassembled unit including all transistors, diodes and resistors.


Placement of the components in relation to each other and the the cavity is critical; replacement of individual components should be avoided.

When the 1500 MHz is present as specified in the paragraph above, retain the connection specified to the counter and tune the 5254 C to 500 MHz . With the 608 F connected as in test 1 , carefully tune the analyzer to get an indication of $550 \mathrm{MHz} \pm 3 \mathrm{MHz}$ on the counter. If there is no 550 MHz output, the low pass filter is probably defective.

## NOTE

After repairing any part of the first and second converter assembly, the calibration procedures specified in paragraphs 5-27 and 5-29 should be performed.

## 41500 MHz NOTCH FILTER

The 1500 MHz notch filter is a coaxial tank rejection filter which prevents the second local oscillator output from being processed in following analyzer circuits.

## SERVICE SHEET 3 (cont'd)

## TEST PROCEDURE 4

Disconnect the 550 MHz input from the third mixer and IF amplifier and connect it to the HP 8405A using the HP 1250-0827, the HP 11592-60003, the HP $1250-0832$, the UG $274 \mathrm{~B} / \mathrm{U}$ and the HP 11593A. With the HP 608 F ( 400 MHz -10 dBm ) connected to the analyzer RF INPUT carefully tune the analyzer for a maximum reading on the HP 8405 A . Typical level is -28.5 dBm . If the correct reading is not observed, the notch filter or the coax cable to the third mixer and IF amplifier is probably defective. If the correct reading is obtained, the circuits shown on Service Sheet 3 are functioning properly.

## NOTE

If the notch filter is replaced, it should be aligned in accordance with paragraph 5-30.

FL2


A8 INSIDE BOTTOM


Figure 8-10. First and Second Converter Assembly A8, Component Locations, and 1500 MHz Notch Filter

## A5

See Figure 8-35 for A5 component locations.


Figure 8-11. First and Second Converter Assembly A8, Top View


Figure 8-12. First and Second Converter Assembly A8, Bottom View (A8A2 Cover Removed)

## SERVICE

SHEET 3



Figure 8-13. First and Second Converter: A8 Sichematic Diagram

## SERVICE SHEET 4

It is assumed that the first local oscillator (YIG) input to the first mixer is missing or incorrect, and that input de voltages to the YIG power supply are present and correct.

## TROUBLESHOOTING PROCEDURE

Before performing the test procedure for the YIG power supply it is best to verify failure of the YIG oscillator or circulator by first performing test procedure 4 , then test procedure 3. When trouble has been isolated to the YIG power supply, the circuit board should be removed and reinstalled using the extender board to provide ready access to test points and components.

## EQUIPMENT REQUIRED

Frequency Counter ..... HP 5245L/5254C
Oscilloscope ..... HP 180A/1801A/1821A
Digital Voltmeter ..... HP 3440A/3443A
Volt-ohm-ammeter ..... HP 412A
Service Kit ..... HP 11592A
50 ohm Dummy Load ..... HP 11593A
Jack-to-jack Adapter ..... HP 1250-0827
VHF Signal Generator ..... HP 608F
BNC Tee ..... UG 274 B/U
Miniature to BNC Adapter ..... HP 1250-0832

## CONTROL SETTINGS

As specified in individual tests.

## 1 SWITCHING FUNCTIONS

The scan width switch assembly performs the following functions:
ZERO scan mode.
A2S2-2F couples the dc level from the FREQUENCY control to the tune amplifier.

A2S2-1F provides a ground return for the noise filter A6C7.
A2S1-5R provides no required function.
PER DIVISION scan mode.
A2S2-2F couples the dc level from the FREQUENCY control to the tune amplifier.

A $2 \mathrm{~S} 2-1 \mathrm{~F}$ provides a ground return to $\mathrm{A} 2 \mathrm{~S} 1-5 \mathrm{R}$.
A2S1-5R provides a ground return for the noise filter (A6C7) when the analyzer is operated in narrow scan modes ( 2 kHz to 200 kHz ).
$0-1250 \mathrm{MHz}$ scan mode.
A2S2-2F connects the dc level from the FREQUENCY control to the marker generator and also couples a fixed dc level from the marker generator to the input of the tune amplifier. The dc level from the marker generator centers the sweep at 625 MHz .

A2S2-1F opens the ground return path for the noise filter C7.
A2S1-5R has no function when the analyzer is operated in the $0-1250 \mathrm{MHz}$ mode.

## TEST PROCEDURE 1

Since there are no active components in the switching circuits they may be easily checked using an ohmmeter to make continuity tests with no power applied to the analyzer.

## SERVICE SHEET 4 (cont'd)

## 2 YIG POWER SUPPLY

The YIG power supply consists basically of two operational amplifiers (tune amplifier and sweep amplifier) and a YIG driver. Its purpose is to closely control the current through the YIG tune coil. The functions of the YIG power supply in the various scan modes are as follows:

ZERO scan mode.
In the ZERO scan mode the analyzer operates as a fixed tuned receiver. The YIG driver and YIG frequency is controlled by the tune amplifier which, in turn, is controlled by the FREQUENCY control. The noise filter, A6C7, is provided a ground return since it is required when the YIG is not being swept.

The sweep amplifier does not function when the analyzer is operated in ZERO scan at wide scan widths. At narrow scan widths the sweep amplifier processes the phase lock error signal to maintain phase lock when the analyzer is stabilized.

PER DIVISION scan mode.
At wide scan widths ( 0.5 to 100 MHz per division) the YIG driver and YIG frequency are controlled by both the tune and sweep amplifiers. One input to the tune amplifier is the dc level from the FREQUENCY control to the base of A6Q1A. The other input is the output of the sweep operational amplifier applied to the base of A6Q1B. Together, these inputs control the tune amplifier which in turn controls the YIG oscillator frequency.

The noise filter, A6C7, is not required when the YIG is being swept so the ground return path is removed at wide scan widths. The sweep amplifier, A6U1, processes the attenuated scan ramp to control the width of the YIG sweep about the center frequency selected by the FREQUENCY control. There is no phase lock error signal present during wide scan modes so the second input to the sweep amplifier performs no function. During narrow scan width ( 2 to 200 kHz per division) operating the scan width attenuator grounds the scan ramp input to the sweep operational amplifier and applies the attenuated scan ramp to the third local oscillator control circuits. Operation of the tune amplifier is the same as in wide scan widths except that the second input from the sweep amplifier consists of the phase lock error signal. The phase lock error signal shifts the first local oscillator to compensate for drift detected by the sampler. This is accomplished by controlling the current in the YIG tune coil. High frequency changes (FM deviations) are compensated for by a secondary error signal applied to the YIG FM coil. The noise filter, A 6 C 7 , is again provided a ground return path since the YIG oscillator frequency is not swept.
$0-1250 \mathrm{MHz}$ scan mode.
In the $0-1250 \mathrm{MHz}$ scan mode the tune amplifier is controlled by a dc level from a preset voltage divider on the marker generator board which causes the scan to be centered at 625 MHz and the output of the sweep amplifier which processes the full scan ramp (the scan width attenuator is bypassed). Since the noise filter is not required when the YIG oscillator is swept, the ground return for A 6 C 7 is removed in this mode.

## TEST PROCEDURE 2

To troubleshoot the tune operational amplifier and the YIG driver place the analyzer in the ZERO scan mode at a wide ( 0.5 to 100 MHz ) scan width. Use the

## SERVICE SHEET 4 (cont'd)

HP $3440 \mathrm{~A} / 3443 \mathrm{~A}$ to check for dc levels shown on the schematic.

## NOTE

In each case where two voltages are listed, the top reading is with the FREQUENCY control set to 0 MHz and the lower reading is at 1250 MHz .
To troubleshoot the sweep amplifier follow steps $\mathrm{a}, \mathrm{b}$ and c below.
2-a. Connect the $1801 / 1801 \mathrm{~A} / 1821 \mathrm{~A}$ Channel A input to XA6-10 (Test Point A) and Channel B input to Test Point B and observe the waveform.


Test Point A BAD: Check scan generator in IF Section and scan width attenuator (Service Sheet 8). Test Point A GOOD and B BAD: Check U1 and associated components. Both waveforms GOOD: Proceed to step 2-b.

2-b. Oscilloscope Channel A remains connected as in 2-a above. Connect oscilloscope Channel B to XA6-15 and observe waveforms.

Test Point C GOOD: Proceed to step 2-c. Test Point C BAD: Check signal path from U1 and YIG coil. (YIG coil should measure about 12 ohms from A7 pin 6 to A7 pin 5).

2-c. Disconnect the circulator from the sampler to defeat phase lock. Connect oscilloscope as in step 2-a and observe waveforms.

Test Point B BAD: Check for same signal at XA6-13 with the oscilloscope Channel B probe. If waveform is good at XA6-13, check U1 and associated components. If the waveform is bad at XA6-13 check the phase lock circuitry, Service Sheets 6 and 7. If waveforms shown in $2-\mathrm{a}, 2$-b and 2 -c are correct, assembly is functioning properly.

## 3 YIG OSCILLATOR

The first local oscillator is a transistor oscillator tuned by a Yttrium-Iron-Garnet (YIG) spherical resonator and followed by at two-stage amplifier. The YIG resonator tunes the oscillator from 2050 to 3300 MHz under the control of a magnetic field from a built-in electromagnet. The two stage amplifier buffers the oscillator circuit from the load impedance and increases the output power. The oscillator-amplifier, including the magnetic circuit, is hermetically sealed to ensure long stable life. The tuning coil (see simplified schematic below) tunes the resonator through all, or portions of, its tuning range when the analyzer is operated in the $0-1250 \mathrm{MHz}$ mode or in wide scan widths of the PER DIVISION mode. The FM coil is used to sweep the YIG resonator $\pm 600 \mathrm{kHz}$ about the center frequency selected by the FREQUENCY control when the stabilized mode is initiated, until a lock point is found.

SERVICE SHEET 4 (cont'd)


## TEST PROCEDURE 3

Connect the RF output of the YIG oscillator-amplifier to the $5245 \mathrm{~L} / 5254 \mathrm{C}$. When the input voltages are as shown on the schematic the RF output should be 2050 MHz above the frequency read on the slide-rule frequency dial.

If the frequency shown on the counter is correct, the tune function of the YIG is operating properly. (If the FM coil is suspected of being defective, refer to Service Sheet 7). If there is no RF output or the frequency is different than that specified, the YIG must be replaced; it is not a field repairable assembly.

## 4. CIRCULATOR

The circulator is a three-port device which accepts the output of the YIG oscillator-amplifier and distributes it to the first converter where it is used for mixing. The reflected power from the first mixer is then coupled by the circulator to the 2 to $3.3 \mathrm{GH} z$ sampler where it is compared to a reference signal to produce the phase lock control signal.

## TEST PROCEDURE $\downarrow$

In order to test the circulator the YIG oscillator-amplifier must be functioning properly. To check the circulator disconnect the output ports one at a time and connect them to the $5245 \mathrm{~L} / 5254 \mathrm{C}$. Reading at both ports should be 2050 MHz above the frequency indicated on the slide-rule frequency dial. If correct readings are not obtained, check the coax cable from the YIG oscillator-amplifier and if found good, replace the circulator.


Figure 8-14. Simplified Circuit Diagram of YIG Oscillator


Figure 8-15. YIG Terminal Locations (ATl Coupled to RF Output Nol Shown)





NOTES:

1. DC VOLTAGES SHOWN ARE TAKEN IN ZERO SCAN. TOP READING IS TAKEN AT OFREQUENCY AND OTHER READING IS AT 1250 MHz .
2. A6R34 AND A6R38 MUST BE THE SAME VALUE.
REFERENCE DESIGNATIONS

| A6 | NO PREFIX | A2 |
| :--- | :--- | :--- |
|  |  |  |
| C1-5, 7-10 | L1 | S1,2 |
| CR1-3 | R6,9 |  |
| R1-20, 32-39 | XA6 | A7 |
| Q1-8 | W4,8 |  |
| U1 | $10,11,15$ | AT1 |
|  | A13 | AT1 |
|  |  |  |

Figure 8-17. First LO and Control Circuits:
A2, A6, A 7 and A13 Schematic Diagram

## SERVICE SHEET 5

It is assumed that all inputs are present and correct and that the output is missing or incorrect.

## TROUBLESHOOTING PROCEDURE

When the cause of a malfunction has been traced to the third converter assembly, three of the four circuit boards may be revealed for testing by removing top and bottom covers.

## CAUTION

Only one cover should be removed at any given time; shielding is critical.
Test procedures for each circuit follow the technical description of the circuit.
EQUIPMENT REQUIRED
Oscilloscope . . . . . . . . . . . . . . . . . . HP 180A/1801A/1821A
Signal Generator . . . . . . . . . . . . . . . . . . . . . . . . HP 612A
Vector Voltmeter . . . . . . . . . . . . . . . . . . . . . . HP 8405A
Service Kit . . . . . . . . . . . . . . . . . . . . . . . . . . HP 11592A
CONTROL SETTINGS
Unless otherwise specified in individual tests.


## THIRD CONVERTER ASSEMBLY A9 (General)

Third converter assembly consists of four small printed circuits and a 550 MHz bandpass filter. All are in the same casting but are individually shielded from each other. Inputs consist of the 550 MHz IF signal from the second converter, an attenuated sweep ramp (in narrow scan stabilized modes), a phase lock offset level (in narrow scan stabilized modes) and a dc level from the FINE TUNE control. Outputs are the 50 MHz IF signal and the third local oscillator signal. Since both sides of the circuit boards are not readily accessible, they will be described and tested in order of "easiest test".

## 1500 MHz LOCAL OSCILLATOR DRIVE A9A2

In wide scan modes the only input to the 500 MHz local oscillator drive circuit is the dc level from the FINE TUNE control. In narrow scan stabilized modes the 500 MHz local oscillator drive amplifier combines and inverts the attenuated scan ramp, the phase lock offset and the FINE TUNE dc level.

## TEST PROCEDURE 1

Connect the HP $180 \mathrm{~A} / 1801 \mathrm{~A} / 1821 \mathrm{~A}$ Channel A probe to Test Point A (C4) and the Channel B probe to Test Point B (Q3-c) and observe the waveforms.


## SERVICE SHEET 5 (cont'd)

If both waveforms are GOOD, proceed to step 2 .
If Test Point A is bad, trouble is in the scan circuits.
If Test Point $B$ is bad, check Q1/Q2/Q3 and associated components.

## NOTE

If it is necessary to remove the 500 MHz local oscillator drive circuit for repair extreme caution must be observed. In addition to the five leads going to input feedthru capacitors, and the four mounting screws, the board is soldered to two feedthru capacitors mounted on the local oscillator assembly. These feedthru capacitors are located between the mounting screws at each end of the board.

## 2500 MHz LOCAL OSCILLATOR A9A4

The 500 MHz local oscillator is a two-transistor voltage tuned oscillator. Tuning is accomplished by controlling the collector to base capacitance in Q1 and Q2. In wide scan modes the oscillator is fixed tuned (first local oscillator is swept), and in narrow scan modes it is swept tuned (first local oscillator is fixed tuned).

## TEST PROCEDURE 2

Connect HP 8405A to THIRD LO OUTPUT jack on front panel using 50 ohm load for termination. With the analyzer operating at 0.5 MHz in the SCAN WIDTH PER DIVISION mode, the output should be 500 MHz at approximately +6 dBm .

If the correct level is observed, proceed to step 3 .
If the correct level is not observed, check Q1/Q2 and associated components.

## 3550 MHz IF AMPLIFIER A9A1

The 550 MHz IF amplifier (A9A1) is a common emitter amplifier (Q2) which provides 11 dB of gain. Q1 controls the bias current to Q2.

## TEST PROCEDURE 3

To gain access to the 550 MHz amplifier it is necessary to remove the bottom cover from the third converter assembly. To ensure validity of measurements however, it is necessary to shield the bandpass filter. A small metal plate securely fastened over the bandpass filter compartment will serve this purpose.

## NOTE

A hole in the cover plate is provided over the input to the bandpass filter for convenience in measuring the 550 MHz amplifier output.

With a $450 \mathrm{MHz},-10 \mathrm{dBm}$ signal applied to the RF INPUT from the HP 612A, connect the HP 8405A to Test Point D (input to bandpass filter). With the analyzer tuned to maximum in ZERO scan mode, with INPUT ATTENUATION set to 0 dB , the reading should be approximately -19 dBm .

If correct reading is obtained proceed to step 4 . If not, check Q1/Q2 and associated components.

## SERVICE SHEET 5 (cont'd)

4550 MHz BANDPASS FILTER
The 550 MHz bandpass filter has a bandpass of 10 MHz centered at 550 MHz . Loss in the filter is approximately 3 dB .

## TEST PROCEDURE 4

The bandpass filter is tested along with the $550 / 50 \mathrm{MHz}$ mixer.

## 5 550/50 MHz MIXER A9A3

The $550 / 50 \mathrm{MHz}$ mixer is a doubled balanced mixer using a matched diode quad for the mixing bridge. Conversion loss is approximately 6 dB .

## TEST PROCEDURE 5

With a $550 \mathrm{MHz},-10 \mathrm{dBm}$ signal from the HP 612 A applied to the 550 MHz IF amplfier input, connect the HP 8405A to the third converter assembly output. The HP 8405A should indicate approximately -7 dBm .

If correct reading is obtained proceed to step 6 . If not, check the Bandpass Filter and the Mixer.

## 650 MHz IF AMPLIFIER A 12

The 50 MHz IF amplifier provides approximately 13 dB of gain. The gain of the amplifier is partly controlled by an attenuated scan ramp to provide higher gain, as the input frequency is increased, to compensate for circuit and cable losses at higher frequency.

## TEST PROCEDURE $\mathfrak{j}$

With the CAL OUTPUT connected to the RF INPUT, connect the HP 8405A to the 50 MHz output (green coax near frequency potentiometer) through a Tee connector and observe the reading. With the analyzer set in ZERO scan mode ( 0.5 $\mathrm{MHz} / \mathrm{DIV}$ ) tune for maximum signal on the meter. Meter should indicate approximately -21 dBm .

To verify gain programming of the amplifier, connect the HP 180A/1801A/1821A to Test Point E (A12C1) and observe the waveform.
$5 \mathrm{~ms} / \mathrm{div}$


CONTROL SETTINGS:
SCAN WIDTH: $0-1250 \mathrm{MHz}$


Figure 8-18. Third Converter Assembly A9, Top View Component Locations (A9A1 and A9A3)

## А9 ВОТТОМ



Figure 8-19. Third Converter Assembly A9, Bottom View Component Locations (A9A2)


Figure 8-20. 550 MHz Amplifier A9A1. Component Locations


Figure 8-22. $550 / 50 \mathrm{MHz}$ Mixer A9A3, Component Locatlons


Figure 8-24. 50 MHz Amplifier A12, Component Locations



50 MHZ AMPLIFIER ASSY A12 108554 - 600481


Figure 8-26. Third Converter: A9 and A12 Schematic Diagram


## CONTROL SETTINGS:

Same as basic.

If either or both of the waveforms are not present, the sampler is probably defective.

If waveforms are good, first disconnect the sampler input from the circulator; the waveforms should disappear. Next place the SCAN WIDTH PER DIVISION switch in the 200 kHz position; the waveforms should again disappear. If the waveform disappears when the circulator input is removed but not when the SCAN WIDTH PER DIVISION switch is set to 200 kHz , the phase lock circuit is defective. Proceed to 2-b.

2-b. Connect the HP $180 \mathrm{~A} / 1801 \mathrm{~A} / 1821 \mathrm{~A}$ to Test Point E (Q4-e) and observe the waveform.

If the correct waveform is not present move the oscilloscope probe to Test Point F (Q2-b) and set scope sensitivity to $0.05 \mathrm{~V} /$ div.



Same as basic

The waveform should be the same as shown above. If the waveform is correct at Test Point F but not at Test Point E, check Q2, Q3, Q4 and associated components. If the waveform is not present at Test Points E or F, trouble is probably in the sampler balance circuit, Q1, or associated components.

When the correct waveform appears at Test Point E momentarily remove the input to the sampler from the circulator; the waveform should disappear. Next momentarily place the SCAN WIDTH PER DIVISION switch in the 200 kHz position; the waveform should again disappear. If the waveform disappears when the circulator is disconnected but does not when the SCAN WIDTH PER DIVISION switch is set to 200 kHz , trouble is probablyin the phase lock circuits shown on Service Sheet 7.

1-b. Connect the HP $180 \mathrm{~A} / 1801 \mathrm{~A} / 1821 \mathrm{~A}$ to Test Point B (Q4-b) and observe the waveform.


Waveform BAD: Check Q3 and associated components.
Waveform GOOD: Proceed to 1-c.
1-c. Connect the HP 180A/1801A/1821A to Test Point C and observe waveform.
$0.2 \mu \mathrm{~s} / \mathrm{div}$


CONTROL SETTINGS:
Any

Waveform BAD: Check Q4 and associated components.
Waveform GOOD: Proceed to step 2 .

## 2 PHASE LOCK AMPLIFIER ASSEMBLY A10A2

The phase lock amplifier assembly consists of a balancing circuit, a two stage differential amplifier and an emitter follower output stage. Feedback is employed from the emitter follower output stage to the input differential amplifier to ensure amplifier gain stability. Positive feedback from C5 compensates for amplifier input capacitance to increase gain bandwidth.

TEST PROCEDURE 2
2-a. Connect the HP 180A/1801A/1821A Channel A and B probes to Test Points D and observe the waveforms.

## SERVICE SHEET 6

It is assumed that the analyzer is working properly in wide scan modes but will not phase lock in narrow scan modes. It is also assumed that dc input voltages are present and correct.

## TROUBLESHOOTING PROCEDURE

In order to gain access to the Sampler Pulse Generator (A10A1) and the Phase Lock Amplifier (A10A2) assemblies it is necessary to remove nine screws and a cover plate under the RF Section.

## EQUIPMENT REQUIRED

Oscilloscope . . . . . . . . . . . . . HP 180A/1801A/1821A
Service Kit . . . . . . . . . . . . . . . . . . . . HP 11592A

## CONTROL SETTINGS

Unless otherwise specified in individual tests.


## 1 SAMPLER PULSE GENERATOR ASSEMBLY A10A1

The sampler pulse generator consists of a reference oscillator and a step generator. The reference oscillator is a standard crystal controlled 1 MHz oscillator which provides the reference base for stabilizing the first local oscillator when the analyzer is operated in narrow scan phase locked modes. The step generator converts the 1 MHz output from the reference oscillator to negative-going spikes to operate a sampling gate in the 2 to 3.3 GHz sampler A10A3. CR1 clips the positive half cycles of the 1 MHz signal and Q3 inverts this signal and applies it to Q4. CR2 clips the negative portion of the signal which is again inverted by Q4. The resulting signal is a negative-going spike at a rate of 1 MHz .

## TEST PROCEDURE 1

1-a. Connect the HP $180 \mathrm{~A} / 1801 \mathrm{~A} / 1821 \mathrm{~A}$ to Test point A (Q1-c) and observe the waveform.


## CONTROL SETTINGS:

Any

Waveform BAD: Check Q1, Q2, Y1 and associated components.
Waveform GOOD: Proceed to 1-b. Note that some clipping has occurred on the positive half of the sine wave. If it has not, CR1 is probably defective.


Figure 8-27 Sampler Pulse Generator A10A1, Component Locations


Figure 8-28. Phase Lock Amplifier Assembly A10A2, Component Locations


Figure 8-29. Sampler Assembly A10A3, Connector Identification


## -60020)




Figure 8-30. Phase Lock Reference and Sampler:
A 10 Schematic Diagram

## SERVICE SHEET 7

It is assumed that the analyzer is working properly in wide scan modes but will not phase lock in narrow scan modes. It is also ascumed that dc input voltages are present and correct and that the tests prescribed in Service Sheet 6 have been conducted with satisfactory results.

## TROUBLESHOOTING PROCEDURE

When the cause of a malfunction has been traced to circuits shown on Service Sheet 7 the A4 assembly should be removed from the RF Section and reinstalled on the extender board to provide access to components and test points.

EQUIPMENT REQUIRED

> Oscilloscope
> HP 180A/1801A/1821A
> Service Kit
> HP 11592A

## CONTROL SETTINGS

Unless otherwise specified in individual tests.

| SCAN WIDTH | PER DIVISION |
| :---: | :---: |
| SCAN WIDTH PER DIVISION | 0.5 MHz |
| BANDWIDTH | Hz |
| SCAN TIME PE | LLISECONDS |

## 1 EOUALIZER-LIMITER-SEARCH LOOP

The equalizer-limiter-search loop provides the search and control function required to phase lock the first local occillator when the analyzer is operated in the stabilized mode at narrow scan widthe ( 2 to 200 kHz per division). When the analyzer in switched to a stabilized mode the search loop oscillates at a frequency of approximately 5 Hz . The search signal, approximately $0.6 \mathrm{Vp}-\mathrm{p}$, is applied to the YIG power supply to sweep the first local oucillator $\ddagger 600 \mathrm{kHz}$. Since the lock points are 1 MHz apart ( $\pm 500 \mathrm{kHz}$ ), thin ensures that the first local oscillator will be swept through a phase lock point. When a phase lock point is reached the output from the sampler, through the phase lock amplifier, ends the search function and the search orcillator ceases to oscillate.

After phase lock is accomplished any phase shift in the first focal oscillator frequency, as compared to the reference signal, is "shaped" by the equalizer and applied to the YIG power supply and the FM coil in the YIG oscillator assembly. The YIG power supply compensates for low frequency shift and the FM coil compensates for high frequency shifts.

## TEST PROCEDURE 1

1s. Connect the HP 180A/1801A/1821A Channel A input to Test Point A (XA4-14) and the Channed B input to Test Point B (U1-10) and observe the waveforms.


## CONTROL SETTINGS:

Same as basic.

## NOTE

It may be necessary to vary the FREQUENCY control to obtain the correct waveform.

If neither waveform is present, trouble is probably in the phase lock amplifier. See Service Sheet 6.

If waveform $A$ is present and B is not, check U1 and associated components. The apparent loss in gain between Test Point A and Test Point B is due to the fact that Test Point 1 is grounded when the analyzer is operated in the unstabilized mode.

Momentarily switch the SCAN WIDTH PER DIVISION switch to 200 kHz . Both waveforms should disappear. If they do not, check the limiter, the lock range adjust setting and connections to the YIG power supply.

1-b. Connect the HP $180 \mathrm{~A} / 1801 \mathrm{~A} / 1821 \mathrm{~A}$ Channel A input to Test Point C (XA4-13) and the Channel B input to Test Point B (U1-10) and observe the search signal.

## NOTE

The signal from the circulator to the sampler must be disconnected to disable the phase lock loop.


CONTROL SETTINGS
Same as basic.

If both waveforms are missing check the search loop components. If signal at Test Point C is missing check the lock range adjust.

## SERVICE SHEET 7 (cont'd)

## 2 AUTOMATIC PHASE COMPENSATION CIRCUIT

The automatic phase compensation circuit consists of a phase lock memory offset amplifier and a phase lock offset delay circuit. The input APC signal is grounded through the SCAN WIDTH PER DIVISION switch at settings of 0.5 MHz or greater regardless of the position of the TUNING STABILIZER switch. For scan widths of 200 kHz or less the APC signal is grounded only when the TUNING STABILIZER switch is in the OFF position.

When the SCAN WIDTH PER DIVISION switch is set to 200 kHz or less, the TUNING STABILIZER switch is on, and the SCAN WIDTH is set to PER DIVISION, a ground is removed from XA4 pins 5 and 12. When the ground is removed from XA4-5 the base of Q5 goes more negative and Q5 is cut off as soon as the charge on C10 is overcome (about 0.5 second). When Q5 is cut off it causes the base of Q6 to go positive to cut off Q6. When Q6 cuts off the relay contacts of Q1 open and the dc level present at the contacts remains stored on C11 to maintain conduction of Q1 at the level at which it was conducting when the relay contacts opened.

During the 0.5 second time that relay Q1 is energized after initiation of the phase lock cycle, the APC signal is processed by operational amplifier Q2/Q3/Q4 and applied to C11 which charges to the level of the signal. When relay K1 opens, C11 cannot discharge because of the high impedance of Q1. The output of Q1 is held at a level determined by the level of the charge on C11 and is applied as an offset voltage to the third local oscillator to compensate for frequency shift required to phase lock the first local oscillator.

When the TUNING STABILIZER is OFF the system is no longer phase locked. Relay K1 energizes and allows C11 to discharge. This removes the offset voltage from the third local oscillator control circuits.

## TEST PROCEDURE 2

Connect the HP $180 \mathrm{~A} / 1801 \mathrm{~A} / 1821 \mathrm{~A}$ Channel A input to Test Point 1 and the Channel B input to Test Point D (Q4-c) and observe the waveforms.

## NOTE

The circulator must be disconnected from the sampler to prevent the analyzer from becoming phase locked.


If waveform at Test Point $D$ is missing check Q2/Q3/Q4 and associated components.

## SERVICE SHEET 7 (cont'd)

If correct waveforms are present and the analyzer still will not phase lock when the circulator is reconnected, check Q1 and the third local oscillator control circuits (Service Sheet 5).

## NOTE

If any of the circuits shown on this Service Sheet are repaired, the procedures specified in paragraphs 5-32 through 5-35 should be performed.


Figure 8-31. Search Loop and Phase Lock Memory Amplifier A4, Component Locations

P/O $8554 B$ RF SECTION



Figure 8-32.


REFERENCE DESIGNATIONS

| A4 | NO PREFIX |
| :--- | :--- |
| C1-11 | S1 |
| CR1-3 | XA4 |
| Q1-11 | W9 |
| R1-32 | A2 |
| U1 | S1-3R |
|  | S2-1F |
|  |  |

## A2,A4

Figure 8-32. Search Loop and Phase Lock Memory: A2 and A4 Schematic Diagram

## SERVICE SHEET 8

It is assumed that trouble has been isolated to that portion of Scan Width Attenuator Assembly A2 shown on Service Sheet 8.

## TROUBLESHOOTING PROCEDURE

Since there are no active components in the circuit to be repaired, the 8554B should be disconnected from the IF Section and the Display Section and an ohmmeter used for point-to-point measurements.

## EQUIPIMENT REQUIRED

Volt-Ohm-Ammeter . . . . . . . . . . . . . . . . . . . HP 412A

## TEST PROCEDURE

Since there are no active components in the circuit, a point-to-point resistance check with the HP 412A should quickly isolate defective components.

## A2



Figure 8-33. Scan Width Switch Assembly A2, Component Locations





Figure 8-34. Scan Width Attenuator: A2 Schematic Diagram

## SERVICE SHEET 9

It is assumed that the marker signal is missing or incorrectly placed, analyzer gain is not linear at the high end of the passband, or that the CRT sweep is not centered at 625 MHz when the analyzer is operated in the 0 to 1250 MHz scan mode.

## TROUBLESHOOTING PROCEDURE

When the cause of a malfunction has been traced to circuits shown on Service Sheet 9 the A5 assembly should be removed from the frame and reinstalled on the extender board to provide proper access to components and test points.

EQUIPMENT REQUIRED

> Oscilloscope . . . . . . . . . . . . . HP 180A/1801A/1821A
> Service Kit . . . . . . . . . . . . . . . . . HP HP 11592A
> Digital Voltmeter . . . . . . . . . . . . . HP 3440A/3443A

## CONTROL SETTINGS

Unless otherwise specified in individual tests.


## MARKER GENERATOR

Markers are generated by comparing the scan voltage with the fixed voltage from the FREQUENCY control when the analyzer is operated in the $0-1250 \mathrm{MHz}$ scan mode. When the two voltage levels are the same the marker appears below the display baseline at a point corresponding to the setting of the FREQUENCY control. The scan voltage is applied to the base of Q1 and the fixed dc level from the FREQUENCY control is applied to the base of Q2. When the input voltages to the bases of Q1/Q2 are equal the output at the junction of CR1 and CR3 is minimum. When this condition exists Q3 provides a negative-going marker signal. This marker is coupled through emitter-follower Q4 to the deflection amplifier in the IF Section through the Display Section.

## TEST PROCEDURE 1

1-a. Connect the HP 180A/1801A/1821A Channel A input to Test Point A (XA5-1) and the Channel B input to Test Point B (junction of CR1 and CR3) and observe the waveform.
CONTROL SETTINGS:

## SERVICE SHEET 9 (cont'd)

If the Test Point A waveform is missing, first check for scan ramp at SCAN IN/OUT jack on the IF Section, then check sweep and tune amplifiers in the YIG power supply.

If Test Point A waveform is correct the the Test Point B waveform is not, verify presence of the dc level from the FREQUENCY control, then check Q1/Q2 and associated components.

1-b. With the Channel B probe connected as in 1-a above, connect the Channel A probe to Test Point C (XA5-14) and observe the waveform.


If waveform A is not present check Q3/Q4 and associated components.

## 2 GAIN COMPENSATION CIRCUIT

As the input frequency to the analyzer increases cabling and circuit losses increase also. To compensate for these losses and to make the analyzer flat across the passband the 50 MHz IF amplifier A12, gain is programmed by the sweep ramp. Operational amplifier U1 isolates the IF amplifier from the tuning ramp and directly controls the gain of the $50 \mathrm{MHz} \mathrm{IF} \mathrm{amplifier} \mathrm{A12}$.

## TEST PROCEDURE 2

Connect the 180A/1801A/1821A Channel A input to Test Point A (XA5-1) and the Channel B input to Test Point D (XA5-8) and observe the waveform.


Waveform BAD: Check U1 and associated components.
Waveform GOOD: Proceed to step 3 .

## SERVICE SHEET 9 (cont'd)

$3 \mathbf{0 - 1 2 5 0 ~ M H z ~ S C A N ~ C E N T E R ~ A D J U S T ~}$
The scan center adjust circuit is a voltage divider which is used to center the scan ramp at 625 MHz when the analyzer is operated in the $0-1250 \mathrm{MHz}$ scan mode.

## TEST PROCEDURE 3

Connect the HP 3440A/3443A between ground and Test Point E (XA5-9). Note the initial reading, rotate R 22 through its range, and return to initial setting. Voltage range should be about -2.6 to -3.6 volts.

## 447 MHz LOCAL OSCILLATOR DRIVE

R19 and R20 comprise a voltage divider which provides a dc level to maintain the frequency of the fourth local oscillator in the IF Section at a fixed frequency of 47 MHz . The output dc level at XA5-6 should be +5.2 volts $\pm 0.15$ volt.


Figure 8-35. Marker Generator Assembly A5, Component Locations



Figure 8


Figure 8-36. Marker Generator: A5 Schematic Diagram

## SERVICE SHEET 10

Normally, malfunctions which occur in the switching circuits will be detected and corrected while troubleshooting the IF Section bandwidth circuits.

## TROUBLESHOOTING PROCEDURE

Since these switches function for voltage switching only, all components and wiring can be checked by monitoring voltage levels at the input connectors to the IF Section Bandwidth Circuits.

## EQUIPMENT REQUIRED

> Service Kit . . . . . . . . . . . . . . . . . . . . HP 11592A Digital Voltmeter . . . . . . . . . . . . .

## CONTROL SETTINGS

As required to check dc levels.

## SCAN WIDTH SWITCH ASSEMBLY A2

The switch section shown provides -12.6 volts to the bandwidth switch assembly for use in selecting desired bandwidths when the analyzer is operated in the ZERO or PER DIVISION modes. When the analyzer is operated in the 0 to 1250 MHz mode, the -12.6 volt dc level is applied through the bandwidth switch assembly to disable the bandwidth selection circuits and ensure that the 300 kHz bandwidth is used.

## TEST PROCEDURE 1

See step 2

## 2 BANDWIDTH SWITCH ASSEMBLY A1

This portion of the bandwidth switch assembly provides positive or negative voltages to various IF Section Components to add, bypass, or remove bandwidth shaping elements in the signal path.

## TEST PROCEDURE 2

Use the HP 3440A/3443A Digital Voltmeter to verify switching voltages.
If all voltages are correct the portions of the SCAN WIDTH and BANDWIDTH switches shown are functioning properly.


Figure 8-37. Bandwidth Switch Assembly A 1. Component Locations




Figure 8-38. IF Section Control Circuits: A1 and A2 Schematic Diagram


Figure 8-39. Simplified Analogic Diagram


8-39. Simplified Analogic Diagram

## SERVICE SHEET 11

It is assumed that the DISPLAY UNCAL lamp is operating erratically or not at all.

## TROUBLESHOOTING PROCEDURE

When a malfunction has been isolated to the analogic circuits, the RF and IF Sections should be extended on the extender cable assembly to provide access to the scan width and bandwidth switch assemblies.

## EQUIPMENT REQUIRED

> Service Kit . . . . . . . . . . . . . . . . . . . . Digital Voltmeter 3440A/3443A

## ANALOGIC CIRCUITS

The DISPLAY UNCAL lamp DS1 illuminates when SCAN WIDTH PER DIVISION, BANDWIDTH, SCAN TIME and VIDEO FILTER are set at any combination of positions which does not permit accurate calibration of the analyzer (see Figure 8-39). The DISPLAY UNCAL lamp is illuminated by a simulated signal and has no actual connection to signal processing circuits.

The RF Section Scan Width Switch Assembly A2 and Bandwidth Switch Assembly A1 both have switch wafers devoted exclusively to analogic. (In the IF Section the Scan Time Switch Assembly and Video Filter Switch also have analogic wafers). When SCAN WIDTH is set to PER DIVISION, current is added to the two buss lines ( 956 and 957 wires) by BANDWIDTH and PER DIVISION. In the IF Section this current is summed with the current added by SCAN TIME and VIDEO FILTER. When the current on either buss line is high enough to bias the light driver in the IF Section into conduction, it turns on and lights the DISPLAY UNCAL lamp (see Service Sheet 10 ). When SCAN WIDTH is set to ZERO, the analogic circuit is disabled.

## TEST PROCEDURE

1-a. Connect the HP $3440 \mathrm{~A} / 3443 \mathrm{~A}$ to TP A (956 wire) and set the analyzer controls as follows:

| SCAN WIDTH | PER DIVISION |
| :---: | :---: |
| BANDWIDTH | 10 kHz |
| VIDEO FILTER | OFF |
| SCAN WIDTH PER DIVISION | 20 kHz |
| SCAN TIME PER DIVISION | MILLISECOND |

The voltmeter should read about +580 mVdc - DISPLAY UNCAL lamp off.
Place VIDEO FILTER switch in 10 kHz position. Meter should read about -600 mVdc - DISPLAY UNCAL remains on. Return VIDEO FILTER switch to OFF. Meter reads about +580 mVdc - DISPLAY UNCAL lamp off.

Place SCAN TIME PER DIVISION switch in 0.5 MILLISECOND position. Meter should read about - 2.4 volts - DISPLAY UNCAL on.

If meter readings are correct but DISPLAY UNCAL does not illuminate, check IF Section analogic circuits.

If voltage are incorrect check switches, resistors, wiring, etc.

## SERVICE SHEET 11 (cont'd)

1.b. Connect the HP 3440A/3443A to TP B (9.57 wire) and set the analyzer controls as initially set in test 1 -a. Meter should read about +165 mVdc .

Place VIDEO FILTER switch in the 10 kHz position. Meter should read about + 50 mVdc - DISPLAY UNCAL on.

Place VIDEO FLLTER switch in the 100 Hz position. Meter should read about -40 mVdc - DISPLAY UNCAL on. Return VIDEO FILTER switch to OFF. Meter reads about +165 mVdc - DISPLAY UNCAL off.

Place SCAN TIME PER DIVISION switch to 0.5 MILLISECONDS. Meter should read about - 1.4 volts DISPLAY UNCAL on. Return SCAN TIME PER DIVISION switch to 1 MILLISECOND. DISPLAY UNCAL off - meter reads about +165 mVdc .

Place BANDWIDTH switch to 3 kHz position. Meter reads approximately -58 mVdc - DISPLAY UNCAL on. Return BANDWIDTH switch to 10 kHz position. DISPLAY UNCAL off - meter reads about +165 mVdc.

Lf readings are correct but DISPLAY UNCAL does not illuminate check IF Section analogic circuits.
If readings are incorrect check switches, resistors, wiring, etc.

## NOTE

A further aid to troubleshooting is Table 5-3. Using the table in conjunction with the schematic should aid in localizing cause of malfunction to specific components.


Figure 8-40. Bandwidth Switch Assembly A 1, Component Locations


PO SCAN WIDTH SWITCH ASSY AZ TO8SSA- 60024)




Figure 8-42. Analogic Circuits: A1 and A2 Schernatic Diagram

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[^0]:    *Signal level at input mixer $=$ (signal level at input) - (input RF
    attenuation.
    18552A IF Section: $60 \mathrm{~dB} / 3$ IF bandwidth ratio $<\mathbf{2 0 : 1}$ from 1
    1 kHz to 300 kHz and $<25: 1$ for 300 Hz IF bandwidth.

[^1]:    28552A IF Section has 10 dB /division $\log$ display only.
    $3_{\text {More than }} 55 \mathrm{~dB}$ below at $3 \mathrm{MHz} \pm 100 \mathrm{kHz}$.
    $4 \pm 0.3 \mathrm{MHz}$ with 8552 A IF section.
    $5_{8552}$ IF Section does not provide manual scan.

[^2]:    * Signal level at input mixer - (signal level at INPUT) (INPUT ATTENUATION).

[^3]:    * Vernier accuracy at 0,6 , and 12 dB ; otherwise $\pm 0.25 \mathrm{~dB}$ ( $\pm 2.8 \%$ ).

[^4]:    * More than 55 dB below at $3 \mathrm{MHz} \pm 100 \mathrm{kHz}$.

